

Automatic Control 1

Linear State Feedback Control

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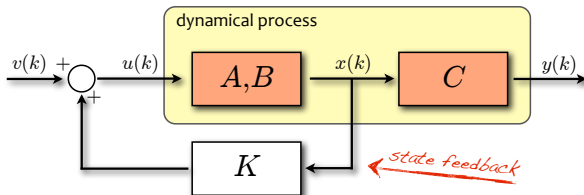
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Stabilization by state feedback

- **Main idea:** design a device that makes the process (A, B, C) asymptotically stable by manipulating the input u to the process



- If measurements of the state vector are available, we can set

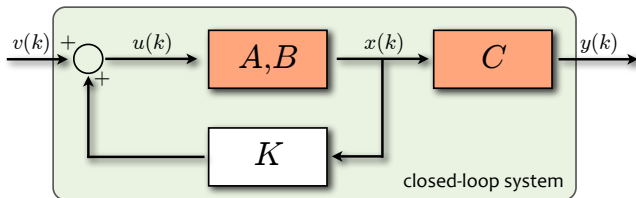
$$u(k) = k_1 x_1(k) + k_2 x_2(k) + \dots + k_n x_n(k) + v(k)$$

- $v(k)$ is an exogenous signal exciting the closed-loop system

Problem

Find a *feedback gain* $K = [k_1 \ k_2 \ \dots \ k_n]$ that makes the closed-loop system asymptotically stable

Stabilization by state feedback



- Let $u(k) = Kx(k) + v(k)$. The overall system is

$$x(k+1) = (A + BK)x(k) + Bv(k)$$

$$y(k) = (C + DK)x(k) + Dv(k)$$

Theorem

(A, B) reachable \Rightarrow the eigenvalues of $(A + BK)$ can be decided **arbitrarily**

Eigenvalue assignment problem

Fact

(A, B) reachable $\Leftrightarrow (A, B)$ is algebraically equivalent to a pair (\tilde{A}, \tilde{B}) in *controllable canonical form*

$$\tilde{A} = \begin{bmatrix} 0 & & & & \\ \vdots & & & & \\ 0 & & I_{n-1} & & \\ -a_0 & -a_1 & \dots & -a_{n-1} & \end{bmatrix}, \quad \tilde{B} = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

The transformation matrix T such that $\tilde{A} = T^{-1}AT$, $\tilde{B} = T^{-1}B$ is

$$T = [B \ AB \ \dots \ A^{n-1}B] \begin{bmatrix} a_1 & a_2 & \dots & a_{n-1} & 1 \\ a_2 & a_3 & \dots & 1 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ a_{n-1} & 1 & 0 & \dots & 0 \\ 1 & 0 & 0 & \dots & 0 \end{bmatrix}$$

where a_1, a_2, \dots, a_{n-1} are the coefficients of the characteristic polynomial

$$p_A(\lambda) = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0 = \det(\lambda I - A)$$

- Let (A,B) reachable and assume $m = 1$ (single input)
- Characteristic polynomials:

$$p_A(\lambda) = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0 \text{ (open-loop eigenvalues)}$$

$$p_d(\lambda) = \lambda^n + d_{n-1}\lambda^{n-1} + \dots + d_1\lambda + d_0 \text{ (desired closed-loop eigenvalues)}$$

- Suppose (A,B) in controllable canonical form

$$A = \begin{bmatrix} 0 & & & \\ \vdots & & I_{n-1} & \\ 0 & & & \\ -a_0 & -a_1 & \dots & -a_{n-1} \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

- As $K = [k_1 \dots k_n]$, we have

$$A + BK = \begin{bmatrix} 0 & & & \\ \vdots & & I_{n-1} & \\ 0 & & & \\ -(a_0 - k_1) & -(a_1 - k_2) & \dots & -(a_{n-1} - k_n) \end{bmatrix}$$

- The characteristic polynomial of $A + BK$ is therefore

$$\lambda^n + (a_{n-1} - k_n)\lambda^{n-1} + \dots + (a_1 - k_2)\lambda + (a_0 - k_1)$$

- To match $p_d(\lambda)$ we impose

$$a_0 - k_1 = d_0, a_1 - k_2 = d_1, \dots, a_{n-1} - k_n = d_{n-1}$$

Procedure

If (A, B) is in controllable canonical form, the feedback gain

$$K = [a_0 - d_0 \quad a_1 - d_1 \quad \dots \quad a_{n-1} - d_{n-1}]$$

makes $p_d(\lambda)$ the characteristic polynomial of $(A + BK)$

- If (A, B) is not in controllable canonical form we must set

$$\begin{aligned}\tilde{K} &= [a_0 - d_0 \quad a_1 - d_1 \quad \dots \quad a_{n-1} - d_{n-1}] \\ K &= \tilde{K}T^{-1} \quad \leftarrow \text{don't invert } T, \text{ solve instead } T'K' = \tilde{K}' \text{ w.r.t. } K' !\end{aligned}$$

where

$$T = R \begin{bmatrix} a_1 & a_2 & \dots & a_{n-1} & 1 \\ a_2 & a_3 & \dots & 1 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ a_{n-1} & 1 & 0 & \dots & 0 \\ 1 & 0 & 0 & \dots & 0 \end{bmatrix}$$

- Explanation: a matrix M and $T^{-1}MT$ have the same eigenvalues

$$\begin{aligned}\det(\lambda I - T^{-1}MT) &= \det(T^{-1}T\lambda - T^{-1}MT) = \det(T^{-1}) \det(\lambda I - M) \det(T) \\ &= \det(\lambda I - M)\end{aligned}$$

- Since $(\tilde{A} + \tilde{B}\tilde{K}) = T^{-1}AT + T^{-1}BKT = T^{-1}(A + BK)T$, it follows that $(\tilde{A} + \tilde{B}\tilde{K})$ and $(A + BK)$ have the same eigenvalues

Ackermann's formula

- Let (A, B) reachable and assume $m = 1$ (single input)
- Characteristic polynomials:

$$p_A(\lambda) = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1\lambda + a_0 \text{ (open-loop eigenvalues)}$$

$$p_d(\lambda) = \lambda^n + d_{n-1}\lambda^{n-1} + \dots + d_1\lambda + d_0 \text{ (desired closed-loop eigenvalues)}$$

- Let $p_d(A) = A^n + d_{n-1}A^{n-1} + \dots + d_1A + d_0I \leftarrow$ (This is $n \times n$ matrix !)

Ackermann's formula

$$K = -[0 \ 0 \ \dots \ 0 \ 1][B \ AB \ \dots \ A^{n-1}B]^{-1}p_d(A)$$

MATLAB

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» K=-acker(A,B,P);
» K=-place(A,B,P);
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where $P = [\lambda_1 \lambda_2 \dots \lambda_n]$ are the desired closed-loop poles

Pole-placement example

- Consider the dynamical system

$$x(k+1) = \begin{bmatrix} 0 & -\frac{1}{4} \\ -3 & 1 \end{bmatrix} x(k) + \begin{bmatrix} 1 \\ -\frac{1}{2} \end{bmatrix} u(k) = Ax(k) + Bu(k)$$

- $\text{rank}R = [B \quad AB] = \begin{bmatrix} 1 & \frac{1}{8} \\ -\frac{1}{2} & -\frac{7}{2} \end{bmatrix} = 2$: the system is reachable

- We want to assign the closed-loop eigenvalues $\frac{1}{2}$ and $\frac{1}{4}$:

$$p_d(\lambda) = (\lambda - \frac{1}{2})(\lambda - \frac{1}{4}) = \lambda^2 - \frac{3}{4}\lambda + \frac{1}{8} = \det(\lambda I - A - BK)$$

$$\begin{aligned} \lambda I - A - BK &= \begin{bmatrix} \lambda & 0 \\ 0 & \lambda \end{bmatrix} - \begin{bmatrix} 0 & -\frac{1}{4} \\ -3 & 1 \end{bmatrix} - \begin{bmatrix} 1 \\ -\frac{1}{2} \end{bmatrix} \begin{bmatrix} k_1 & k_2 \end{bmatrix} \\ &= \begin{bmatrix} \lambda & 0 \\ 0 & \lambda \end{bmatrix} - \begin{bmatrix} 0 & -\frac{1}{4} \\ -3 & 1 \end{bmatrix} - \begin{bmatrix} k_1 & k_2 \\ -\frac{k_1}{2} & -\frac{k_2}{2} \end{bmatrix} \\ &= \begin{bmatrix} \lambda - k_1 & \frac{1}{4} - k_2 \\ 3 + \frac{k_1}{2} & \lambda - 1 + \frac{k_2}{2} \end{bmatrix} \end{aligned}$$

- Therefore:

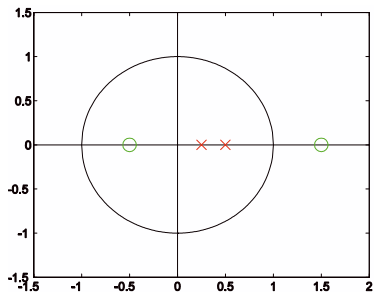
$$\begin{aligned} \det(\lambda I - A - BK) &= \begin{vmatrix} \lambda - k_1 & \frac{1}{4} - k_2 \\ 3 + \frac{k_1}{2} & \lambda - 1 + \frac{k_2}{2} \end{vmatrix} = (\lambda - k_1)\left(\lambda - 1 + \frac{k_2}{2}\right) - \left(3 + \frac{k_1}{2}\right)\left(\frac{1}{4} - k_2\right) \\ &= \lambda^2 + \left(\frac{k_2}{2} - 1 - k_1\right)\lambda + \left(\frac{7k_1}{8} - \frac{3}{4} + 3k_2\right) \end{aligned}$$

- Impose that the coefficients of the polynomials are the same

$$\begin{cases} \frac{k_2}{2} - 1 - k_1 = -\frac{3}{4} \\ \frac{7k_1}{8} - \frac{3}{4} + 3k_2 = \frac{1}{8} \end{cases} \Rightarrow \begin{cases} k_1 = \frac{k_2}{2} - \frac{1}{4} \\ \frac{7k_2}{16} - \frac{7}{32} - \frac{3}{4} + 3k_2 = \frac{1}{8} \end{cases}$$

- Finally, we get

$$\begin{cases} k_2 = \frac{7}{22} \\ k_1 = -\frac{1}{11} \end{cases} \Rightarrow K = \frac{1}{11} \begin{bmatrix} -1 & \frac{7}{2} \end{bmatrix}$$



o = open-loop eigenvalues

x = closed-loop eigenvalues

MATLAB

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» A=[ 0 -1/4 ; -3 1 ];  
» B=[ 1 ; -1/2 ];  
» K=-place(A,B,[ 1/2 1/4 ])  
  
K = -0.0909  0.3182
```

Eigenvalue assignment for unreachable systems

Theorem

If $\text{rank}(R) = n_c < n$ then $n - n_c$ eigenvalues cannot be changed by state feedback

Proof:

- Let T be the change of coordinates transforming (A, B) in canonical reachability decomposition
- Let $K \in \mathbb{R}^{m \times n}$ be a feedback gain, and let $\tilde{K} = KT$ the corresponding gain in transformed coordinates
- As observed earlier, $A + BK$ and $\tilde{A} + \tilde{B}\tilde{K}$ have the same eigenvalues
- Let $\tilde{K} = [K_{uc} \ K_c]$, $K_c \in \mathbb{R}^{m \times n_c}$. Then

$$\tilde{A} + \tilde{B}\tilde{K} = \begin{bmatrix} A_{uc} & 0 \\ A_{21} + B_c K_{uc} & A_c + B_c K_c \end{bmatrix}$$

The eigenvalues of the unreachable part A_{uc} cannot be changed !