

Modeling, Control, and Reachability Analysis of Discrete-Time Hybrid Systems

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DISC Course on
Modeling and Control of Hybrid Systems



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(founded in 1240)

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COHES Group
Control and Optimization of Hybrid and Embedded Systems

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Outline

- What is a “hybrid system” ?
- Models for hybrid systems
- Control of hybrid systems
- Safety analysis of hybrid systems
- Examples (automotive)

Motivating Problem



GOAL:

command gear ratio, gas pedal, and brakes to track a desired speed and minimize consumptions



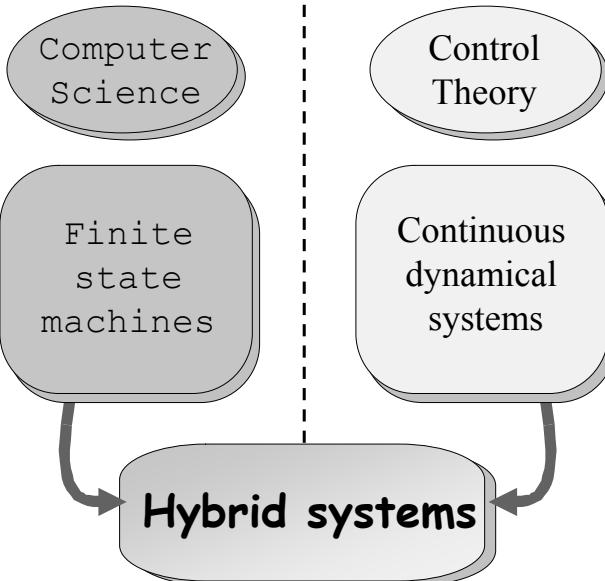
CHALLENGES:

- continuous and discrete inputs
- dynamics depends on gear
- nonlinear torque/speed maps

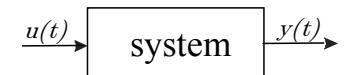
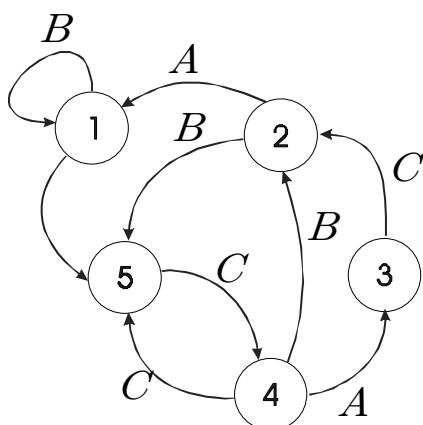
Hybrid Systems

(Witsenhausen, 1966)

$$\begin{aligned} X &= \{1, 2, 3, 4, 5\} \\ U &= \{A, B, C\} \end{aligned}$$

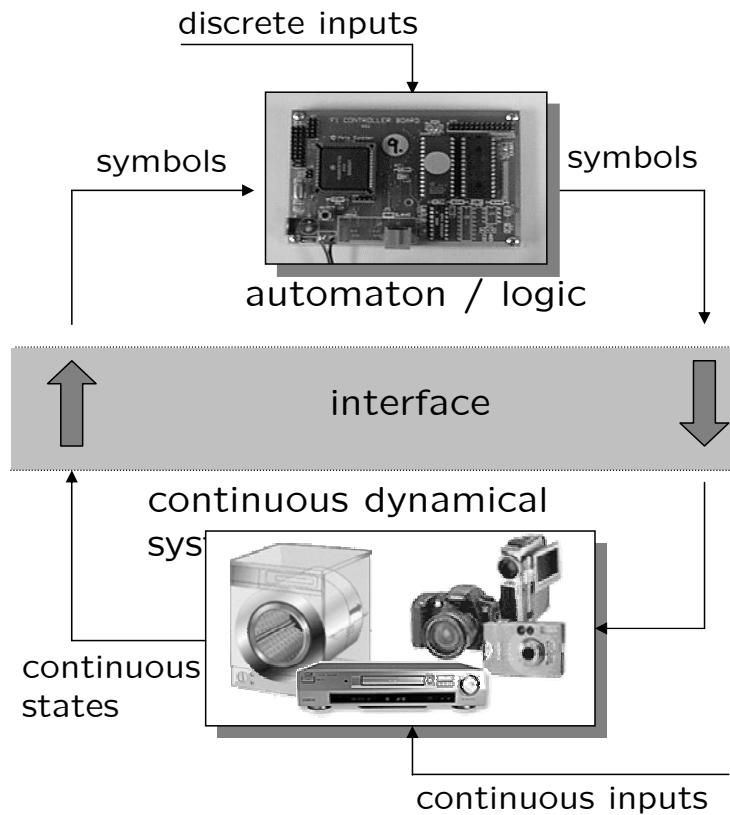


$$\begin{aligned} x \in \mathbf{R}^n, u \in \mathbf{R}^m \\ y \in \mathbf{R}^p \end{aligned}$$



$$\begin{cases} \frac{dx(t)}{dt} = f(x(t), u(t)) \\ y(t) = g(x(t), u(t)) \end{cases}$$

Motivation: Embedded Systems



- Consumer electronics
- Home appliances
- Office automation
- Automobiles
- Industrial plants
- ...

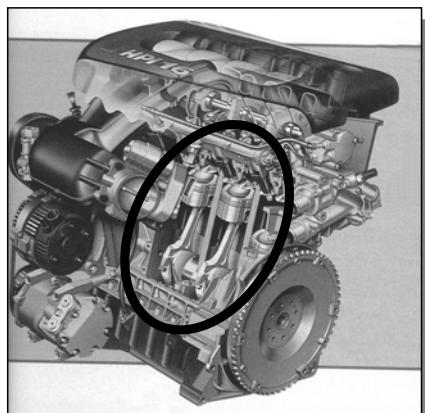
Motivation: “Intrinsically Hybrid”



- Transmission

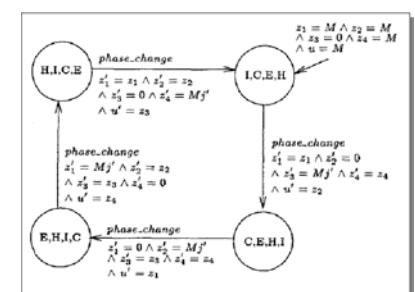
Discrete command
(R,N,1,2,3,4,5)

+ Continuous
dynamical variables
(velocities, torques)



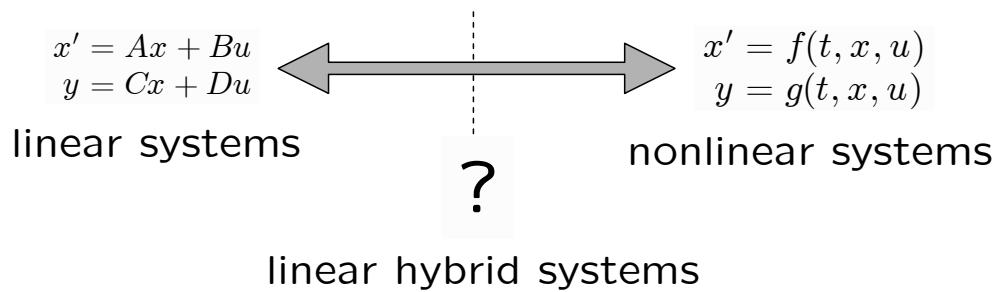
- Four-stroke engines

Automaton,
dependent on
power train motion



Key Requirements for Hybrid Models

- Descriptive enough to capture the behavior of the system
 - continuous dynamics (physical laws)
 - logic components (switches, automata, software code)
 - interconnection between logic and dynamics
- Simple enough for solving *analysis* and *synthesis* problems



Piecewise Affine Systems

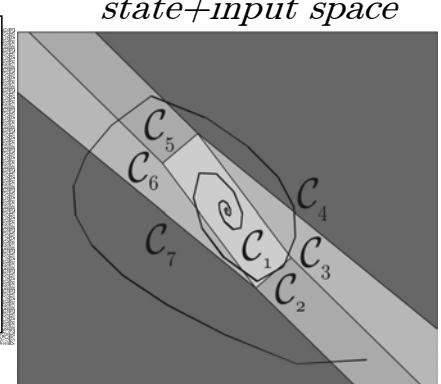
(Sontag 1981)

$$\begin{aligned} x'(k) &= A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\ y(k) &= C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \end{aligned}$$

$$i(k) \text{ s.t. } H_{i(k)}x(k) + J_{i(k)}u(k) \leq K_{i(k)}$$

$$x \in \mathcal{X} \subseteq \mathbb{R}^n, u \in \mathcal{U} \subseteq \mathbb{R}^m, y \in \mathcal{Y} \subseteq \mathbb{R}^p$$

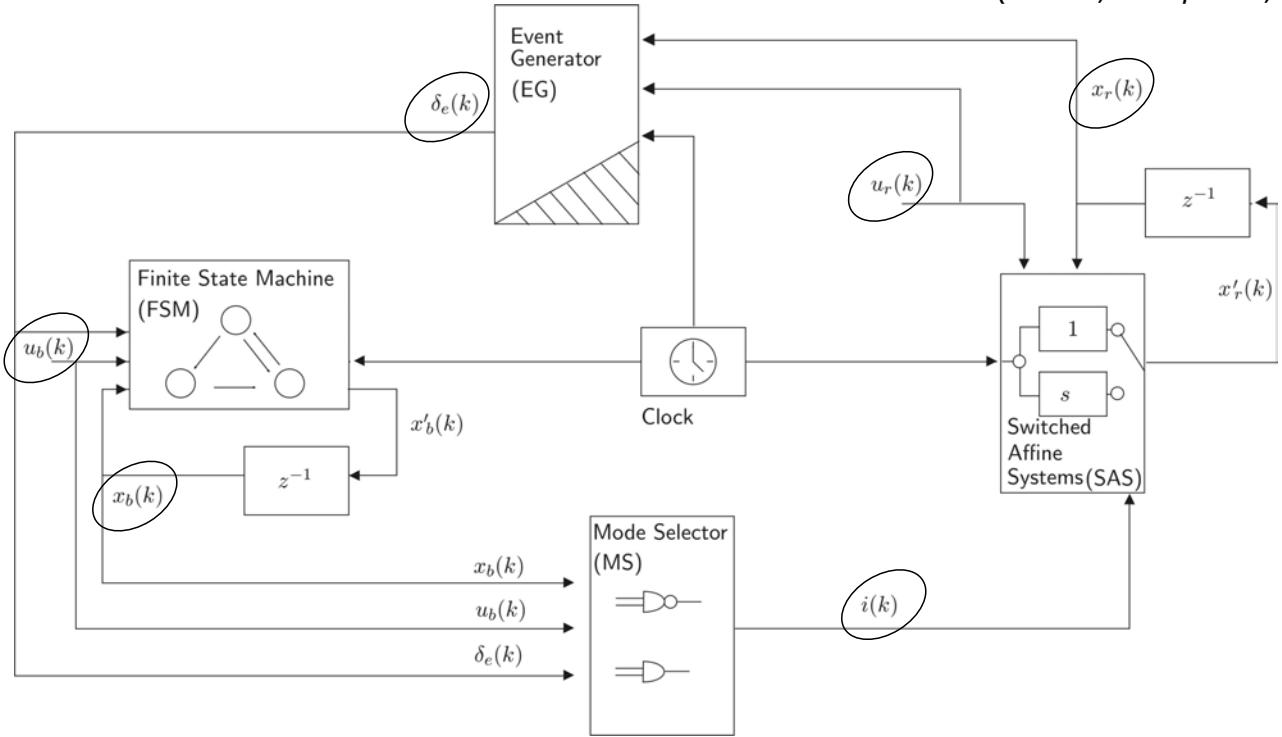
$$i(k) \in \{1, \dots, s\}$$



- Approximates nonlinear dynamics arbitrarily well
- Suitable for stability analysis, reachability analysis (verification), controller synthesis, ...

Discrete Hybrid Automata

(Torrisi, Bemporad, 2003)



$x_r \in \mathbb{R}^{n_r}$ = continuous states

$x_b \in \{0, 1\}^{n_b}$ = binary states

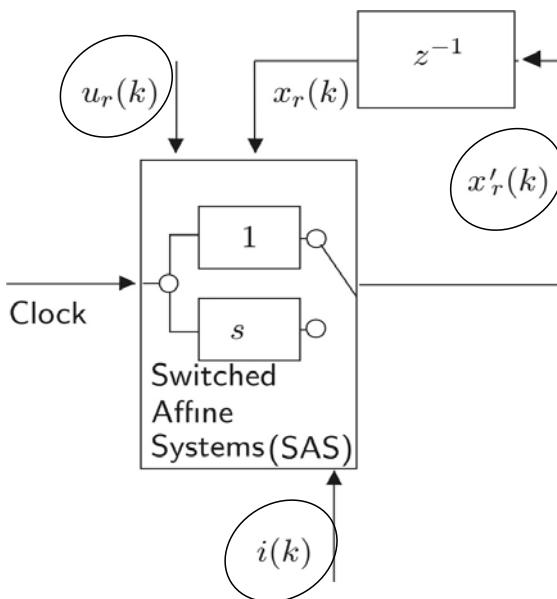
$i(k) \in \{1, \dots, s\}$ = mode

$u_r \in \mathbb{R}^{m_r}$ = continuous inputs

$u_b \in \{0, 1\}^{m_b}$ = binary inputs

$\delta_e \in \{0, 1\}^{n_e}$ = event conditions

Switched Affine Systems



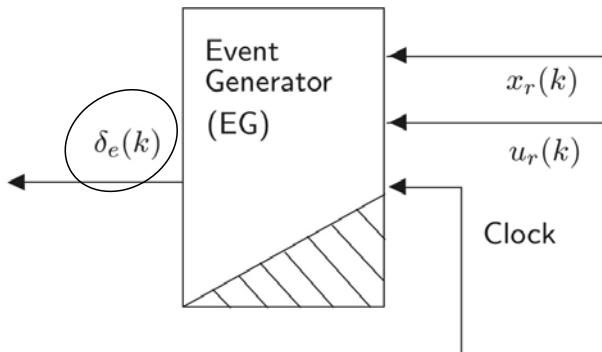
Linear affine dynamics depends upon the mode $i(k)$

$$x'_r(k) = A_{i(k)}x_r(k) + B_{i(k)}u_r(k) + f_{i(k)}$$

$$y_r(k) = C_{i(k)}x_r(k) + D_{i(k)}u_r(k) + g_{i(k)}$$

$x_r \in \mathbb{R}^{n_r}$, $u_r \in \mathbb{R}^{m_r}$, $y_r \in \mathbb{R}^{p_r}$

Event Generator

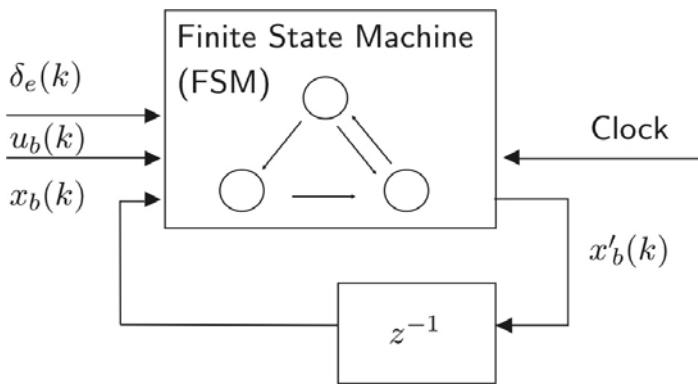


Generates a logic signal according to the satisfaction of a linear affine constraint

$$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_r(k) + K^i u_r(k) + J^i k \leq W^i], \quad \delta_e^i \in \{0, 1\}$$

$$x_r \in \mathbb{R}^{n_r}, \quad u_r \in \mathbb{R}^{m_r}, \quad \delta_e^i \in \{0, 1\}$$

Finite State Machine

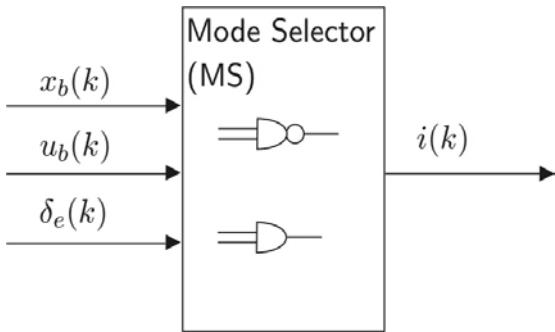


Discrete dynamics evolving according to a Boolean state update function

$$\begin{aligned} x'_b(k) &= f_B(x_b(k), u_b(k), \delta_e(k)) \\ y_b(k) &= g_B(x_b(k), u_b(k), \delta_e(k)) \end{aligned}$$

$$x_b \in \{0, 1\}^{n_b}, \quad y_b \in \{0, 1\}^{p_b}$$

Mode Selector



a Boolean function
selects the active mode
 $i(k)$ of the SAS

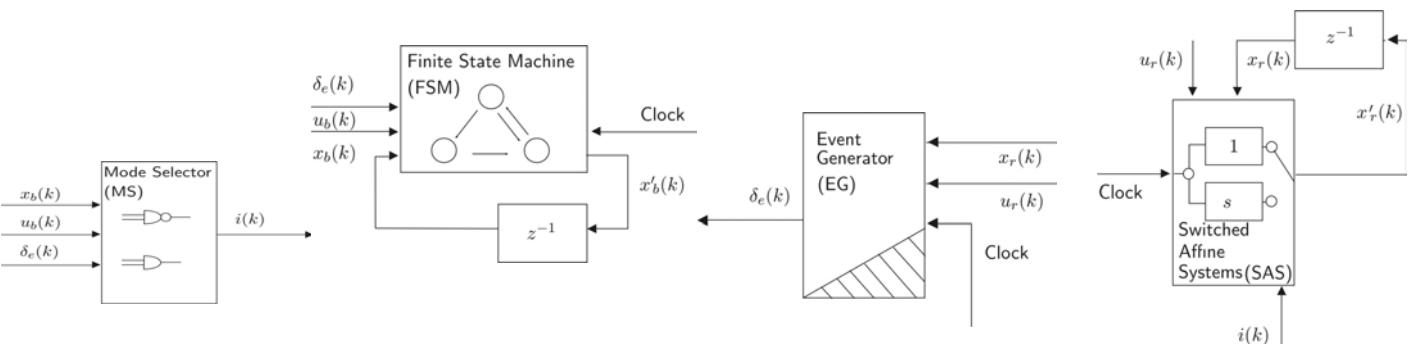
$$i(k) = f_M(x_b(k), u_b(k), \delta_e(k))$$

$$i(k) \in \left\{ \begin{bmatrix} 1 \\ 0 \\ \vdots \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ 1 \\ \vdots \\ 0 \end{bmatrix}, \dots, \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 1 \end{bmatrix} \right\} \subset \{0, 1\}^s$$

Logic and Inequalities

Glover 1975, Williams 1977

$X_1 \vee X_2$	$X_1 + X_2 \geq 1$
Any logic statement $f(X) = \text{TRUE}$	$A\delta \leq B$
$\bigwedge_{j=1}^m (\bigvee_{i \in P_j} X_i \vee \bigvee_{i \in N_j} \neg X_i)$ (CNF) $N_j, P_j \subseteq \{1, \dots, n\}$	$\begin{cases} 1 \leq \sum_{i \in P_1} X_i + \sum_{i \in N_1} (1 - X_i) \\ \vdots \\ 1 \leq \sum_{i \in P_m} X_i + \sum_{i \in N_m} (1 - X_i) \end{cases}$
$[\delta_e^i(k) = 1] \leftrightarrow [H^i x_r(k) \leq W^i]$	$H^i x_r(k) - W^i \leq M^i(1 - \delta_e^i)$ $H^i x_r(k) - W^i > m^i \delta_e^i$
IF δ THEN $z = a_1^T x + b_1^T u + f_1$ ELSE $z = a_2^T x + b_2^T u + f_2$	$(m_2 - M_1)\delta + z \leq a_2 x + b_2 u + f_2$ $(m_1 - M_2)\delta - z \leq -a_2 x - b_2 u - f_2$ $(m_1 - M_2)(1 - \delta) + z \leq a_1 x + b_1 u + f_1$ $(m_2 - M_1)(1 - \delta) - z \leq -a_1 x - b_1 u - f_1$



Transformation of Logic into Linear Integer Inequalities

0. Given a Boolean statement $F(X_1, X_2, \dots, X_n)$

1. Convert to Conjunctive Normal Form (CNF):

$$\bigwedge_{j=1}^m \left(\bigvee_{i \in P_j} X_i \bigvee_{i \in N_j} \bar{X}_i \right)$$

2. Transform into inequalities:

$$A\delta \leq B, \quad \delta \in \{0, 1\}$$

$$\begin{aligned} 1 &\leq \sum_{i \in P_1} \delta_i + \sum_{i \in N_1} (1 - \delta_i) \\ &\vdots \\ 1 &\leq \sum_{i \in P_m} \delta_i + \sum_{i \in N_m} (1 - \delta_i) \end{aligned}$$

→ Every logic proposition can be translated into linear integer inequalities

General Transformation into Linear Integer Inequalities

Geometric approach: The polytope $P \triangleq \{\delta : A\delta \leq B\}$ is the convex hull of the rows of the truth table T

	x_1	x_2	...	x_{n-1}	$x_n = F(x_1, \dots, x_{n-1})$
T:	0	0		0	1
	0	0		1	0
	:	:		:	
	1	1		1	1

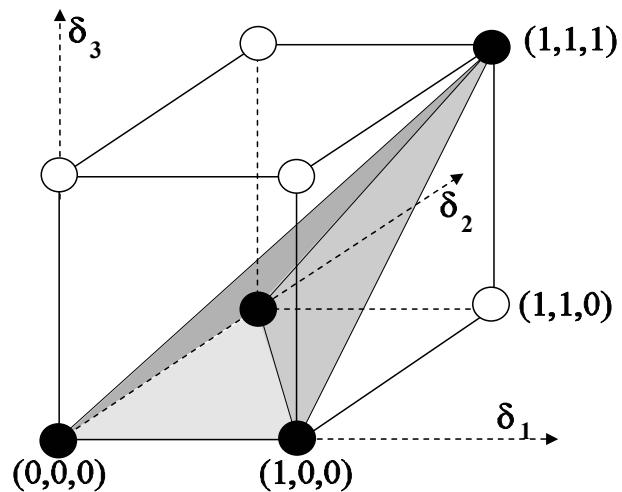
$$\Rightarrow A\delta \leq B, \quad \delta \in \{0, 1\}^n$$

→ Every logic proposition can be translated into linear integer inequalities

Truth Tables → Linear Integer Inequalities

Example: logic “AND”

δ_1	δ_2	δ_3
0	0	0
0	1	0
1	0	0
1	1	1



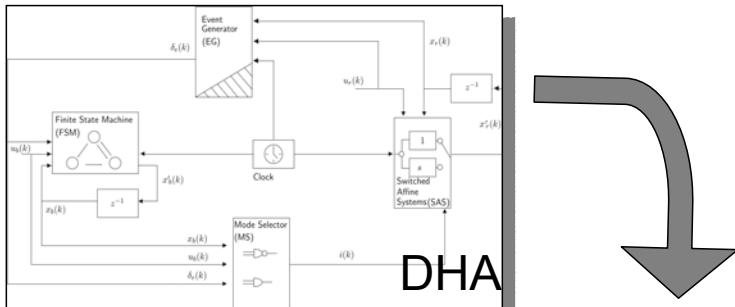
Key idea:

White points cannot be in the hull of black points

$$\text{conv} \left(\begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} \right) = \left\{ \delta : \begin{array}{l} -\delta_1 + \delta_3 \leq 0 \\ -\delta_2 + \delta_3 \leq 0 \\ \delta_1 + \delta_2 - \delta_3 \leq 1 \end{array} \right\}$$

Convex hull algorithms: `cdd`, `lrs`, `qhull`, `chD`, `Hull`, `Porto`

Mixed Logical Dynamical Systems



Mixed Logical Dynamical (MLD) Systems

$$\begin{aligned} x(t+1) &= Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) + B_5 \\ y(t) &= Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t) + D_5 \\ E_2\delta(t) + E_3z(t) &\leq E_4x(t) + E_1u(t) + E_5 \end{aligned}$$

(Bemporad, Morari 1999)

Continuous and binary variables

$$x \in \mathbb{R}^{n_r} \times \{0, 1\}^{n_b}, u \in \mathbb{R}^{m_r} \times \{0, 1\}^{m_b}$$

$$y \in \mathbb{R}^{p_r} \times \{0, 1\}^{p_b}, \delta \in \{0, 1\}^{r_b}, z \in \mathbb{R}^{r_r}$$

- Computationally oriented (Mixed Integer Programming)
- Suitable for optimal controller synthesis, verification, ...

A Simple Example

- System:

$$x(t+1) = \begin{cases} 0.8x(t) + u(t) & \text{if } x(t) \geq 0 \\ -0.8x(t) + u(t) & \text{if } x(t) < 0 \end{cases}$$

$$-10 \leq x(t) \leq 10, -1 \leq u(t) \leq 1$$

- Associate $[\delta(t) = 1] \leftrightarrow [x(t) \geq 0]$ and transform



$$\begin{aligned} -m\delta(t) &\leq x(t) - m & M = -m = 10 \\ -(M + \epsilon)\delta(t) &\leq -x(t) - \epsilon & \epsilon > 0 \quad \text{small} \end{aligned}$$

- Then $x(t+1) = 1.6\delta(t)x(t) - 0.8x(t) + u(t)$

$$z(t) = \delta(t)x(t)$$



$$\begin{aligned} z(t) &\leq M\delta(t) \\ z(t) &\geq m\delta(t) \\ z(t) &\leq x(t) - m(1 - \delta(t)) \\ z(t) &\geq x(t) - M(1 - \delta(t)) \end{aligned}$$

- Rewrite as linear equation



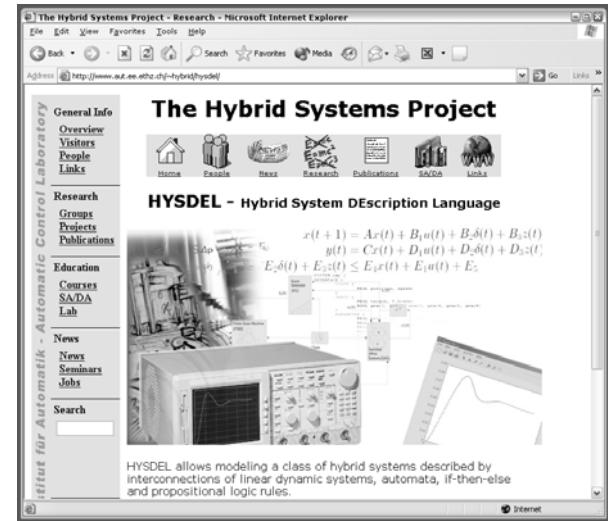
$$x(t+1) = 1.6z(t) - 0.8x(t) + u(t)$$

HYSDEL

(HYbrid Systems DEscription Language)

- Describe *hybrid systems*:

- Automata
- Logic
- Lin. Dynamics
- Interfaces
- Constraints



(Torrisi, Bemporad, 2003)

- Automatically generate MLD models in Matlab

- MLD model is not unique in terms of the number of auxiliary variables → optimize model (minimize # binary variables !)

Example 1: AD section



$$[s = T] \leftrightarrow [h \geq h_{\max}]$$

```

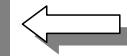
SYSTEM tank {
    INTERFACE {
        STATE {
            REAL h [-0.3,0.3];
        }
        INPUT {
            REAL Q [-10,10];
        }
        PARAMETER {
            REAL k     = 1;
            REAL e     = 1e-6;
        } /* end interface */
    }

    IMPLEMENTATION {
        AUX {
            BOOL s;
        }

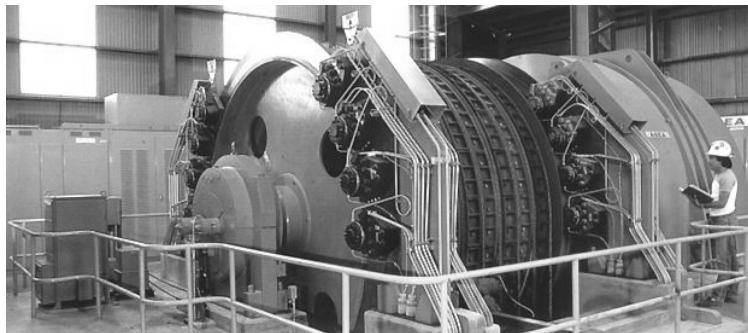
        AD {
            s = hmax - h <= 0;
        }

        CONTINUOUS {
            h = h + k * Q;
        } /* end implementation */
    } /* end system */
}

```



Example 2: DA section



Nonlinear amplification unit

$$u_{comp} = \begin{cases} u & (u < u_t) \\ 2.3u - 1.3u_t & (u \geq u_t) \end{cases}$$

```

SYSTEM motor {
    INTERFACE {
        STATE {
            REAL ucomp;
        }
        INPUT {
            REAL u [0,10];
        }
        PARAMETER {
            REAL ut    = 1;
            REAL e    = 1e-6;
        } /* end interface */
    }
}

```

```

IMPLEMENTATION {
    AUX {
        REAL unl;
        BOOL th;
    }

    AD {
        th = ut - u <= 0;
    }
}

```

```

DA {
    unl = { IF th THEN 2.3*u - 1.3*ut
             ELSE u };
}

```



```

CONTINUOUS {
    ucomp = unl;
} /* end implementation */
} /* end system */

```

Example 3: LOGIC section



$$u_{brake} = u_{alarm} \wedge (\neg s_{tunnel} \vee \neg s_{fire})$$

$$s_{fire} \rightarrow u_{alarm}$$

```

SYSTEM train {
    INTERFACE {
        STATE {
            BOOL brake; }
        INPUT {
            BOOL alarm, tunnel, fire; }

    } /* end interface */

    IMPLEMENTATION {
        AUX {
            BOOL decision; }

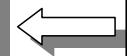
        LOGIC {
            decision =
                alarm & (~tunnel | ~fire); }

        AUTOMATA {
            brake = decision; }

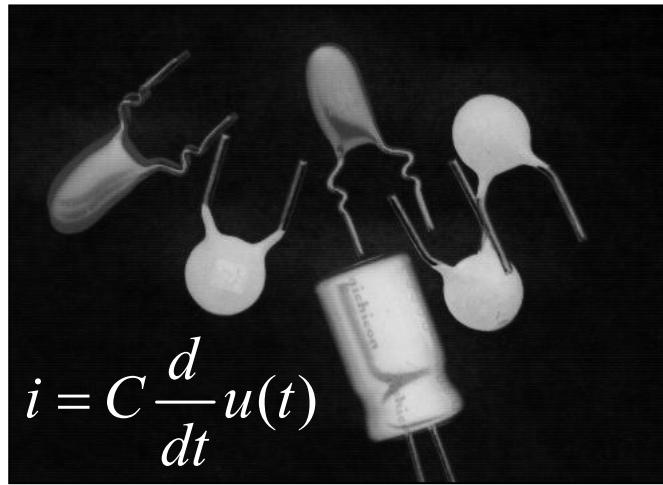
        MUST {
            fire -> alarm; }

    } /* end implementation */
} /* end system */

```



Example 4: CONTINUOUS section



Apply forward difference rule:

$$u(k+1) = u(k) + \frac{T}{C} i(k)$$

```

SYSTEM capacitorD {
    INTERFACE {
        STATE {
            REAL u; }

        PARAMETER {
            REAL R = 1e4;
            REAL C = 1e-4;
            REAL T = 1e-1; }

    } /* end interface */

    IMPLEMENTATION {

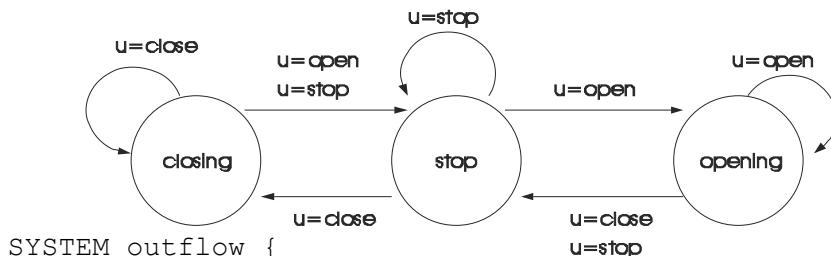
        CONTINUOUS {
            u = u - T/C/R*i; }

        } /* end implementation */
} /* end system */

```



Example 5: AUTOMATA section



```

SYSTEM outflow {
    INTERFACE {
        STATE {
            BOOL closing, stop, opening; }
        INPUT {
            BOOL uclose, uopen, ustope; }
    } /* end of interface */

    IMPLEMENTATION {

```



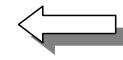
```

        AUTOMATA {
            closing = (uclose & closing) | (uclose & stop);
            stop   = ustope | (uopen & closing) | (uclose & opening);
            opening = (uopen & stop) | (uopen & opening); }
    
```

```

        MUST {
            ~ (uclose & uopen);
            ~ (uclose & ustope);
            ~ (uopen & ustope); }
    } /* end implementation */
} /* end system */

```



Example 6: MUST section



```

SYSTEM watertank {
    INTERFACE {
        STATE {
            REAL h; }
        INPUT {
            REAL Q; }
        PARAMETER {
            REAL hmax = 0.3;
            REAL k      = 1; }
    } /* end interface */

```

```

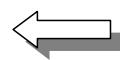
    IMPLEMENTATION {
        CONTINUOUS {
            h = h + k*Q; }
    }

```

```

    MUST {
        h - hmax <= 0;
        -h           <= 0; }

```



```

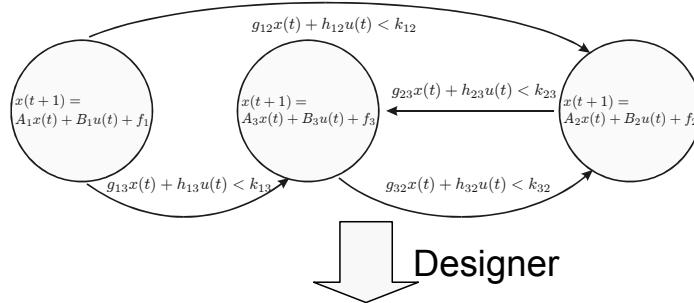
    } /* end implementation */
} /* end system */

```

$$0 \leq h \leq h_{\max}$$

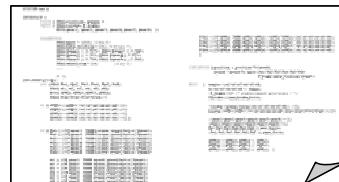
Modeling Flow

Process

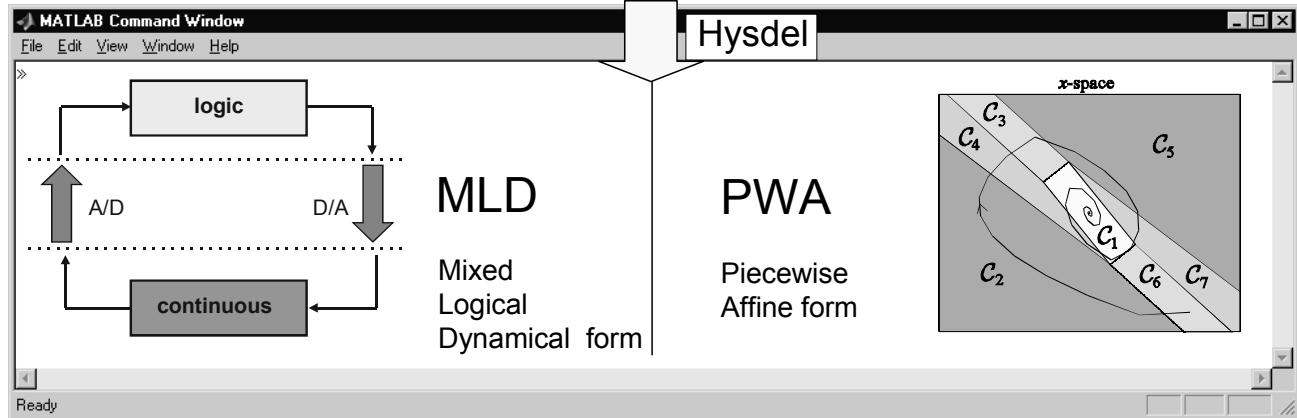


Text

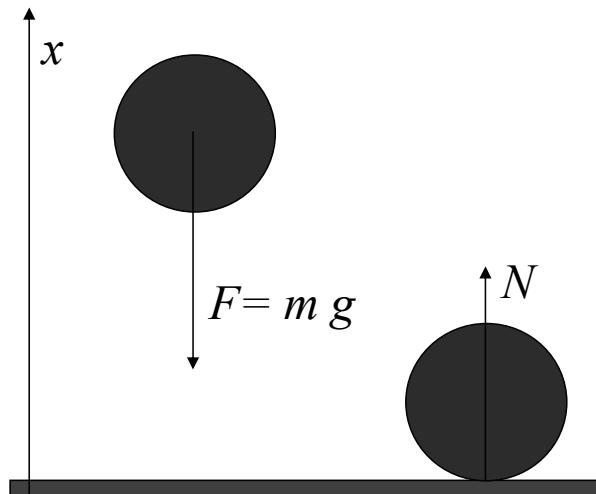
HYSDEL Source



Matlab



Example: Bouncing Ball



$$\begin{aligned}\ddot{x} &= -g \\ x \leq 0 \Rightarrow \dot{x}(t^+) &= -(1-\alpha)\dot{x}(t^-)\end{aligned}$$

$$\alpha \in [0, 1]$$

How to model this system in MLD form?

HYSDEL - Bouncing Ball

```
SYSTEM bouncing_ball {
INTERFACE {
/* Description of variables and constants */
    STATE { REAL height [-10,10];
              REAL velocity [-100,100];      }

PARAMETER {
    REAL g=9.8;
    REAL dissipation=.4; /* 0=elastic, 1=完全ly anelastic */
    REAL Ts=.05;  }

IMPLEMENTATION {
AUX {   REAL z1;
        REAL z2;
        BOOL negative;      }

AD {   negative = height <= 0;  }

DA {   z1 = { IF negative THEN height-Ts*velocity
            ELSE height+Ts*velocity-Ts*Tg};
        z2 = { IF negative THEN -(1-dissipation)*velocity
            ELSE velocity-Ts*g};  }

CONTINUOUS {
height = z1;
velocity=z2; }

}}}
```



System Theory for Hybrid Systems

- Analysis
 - Well-posedness
 - Realization & Transformation
 - Stability
 - Reachability (=Verification)
 - Observability
- Synthesis
 - Control
 - State estimation
 - Identification
 - Modeling language

Well-posedness

Are state and output trajectories defined ?
Uniquely defined ? Persistently defined ?

- MLD well-posedness :

$$\delta(t) = F(x(t), u(t))$$

$$z(t) = G(x(t), u(t))$$

$$x(t+1) = Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t)$$

$$y(t) = Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t)$$

$$E_2\delta(t) + E_3z(t) \leq E_1u(t) + E_4x(t) + E_5$$

$$\begin{aligned} \{x(t), u(t)\} &\rightarrow \{x(t+1)\} \\ \{x(t), u(t)\} &\rightarrow \{y(t)\} \end{aligned}$$

are single valued

Definition 1 Let $\Omega \subseteq \mathbb{R}^n \times \mathbb{R}^m$ be a set of input+state pairs. A hybrid MLD system is called well-posed on Ω , if for all pairs $(x(t), u(t)) \in \Omega$ there exists a solution $x(t+1), y(t), \delta(t), z(t)$ and moreover, $x(t+1), y(t)$ are uniquely determined.

Numerical test based on mixed-integer programming available

(Bemporad, Morari, *Automatica*, 1999)

Realization and Transformation (State-Space Hybrid Models)

Other Existing Hybrid Models

- Linear complementarity (LC) systems (Heemels, 1999)

$$\begin{aligned}x(t+1) &= Ax(t) + B_1 u(t) + B_2 w(t) \\y(t) &= Cx(t) + D_1 u(t) + D_2 w(t) \\v(t) &= E_1 x(t) + E_2 u(t) + E_3 w(t) + e_4 \\0 \leq v(t) \perp w(t) &\geq 0\end{aligned}$$

Ex: mechanical systems
circuits with diodes etc.

- Extended linear complementarity (ELC) systems

Generalization of LC systems

(De Schutter, De Moor, 2000)

- Min-max-plus-scaling (MMPS) systems

(De Schutter, Van den Boom, 2000)

$$\begin{aligned}x(t+1) &= M_x(x(t), u(t), d(t)) \\y(t) &= M_y(x(t), u(t), d(t)) \\0 \geq M_c(x(t), u(t), d(t)) &\geq 0\end{aligned}$$

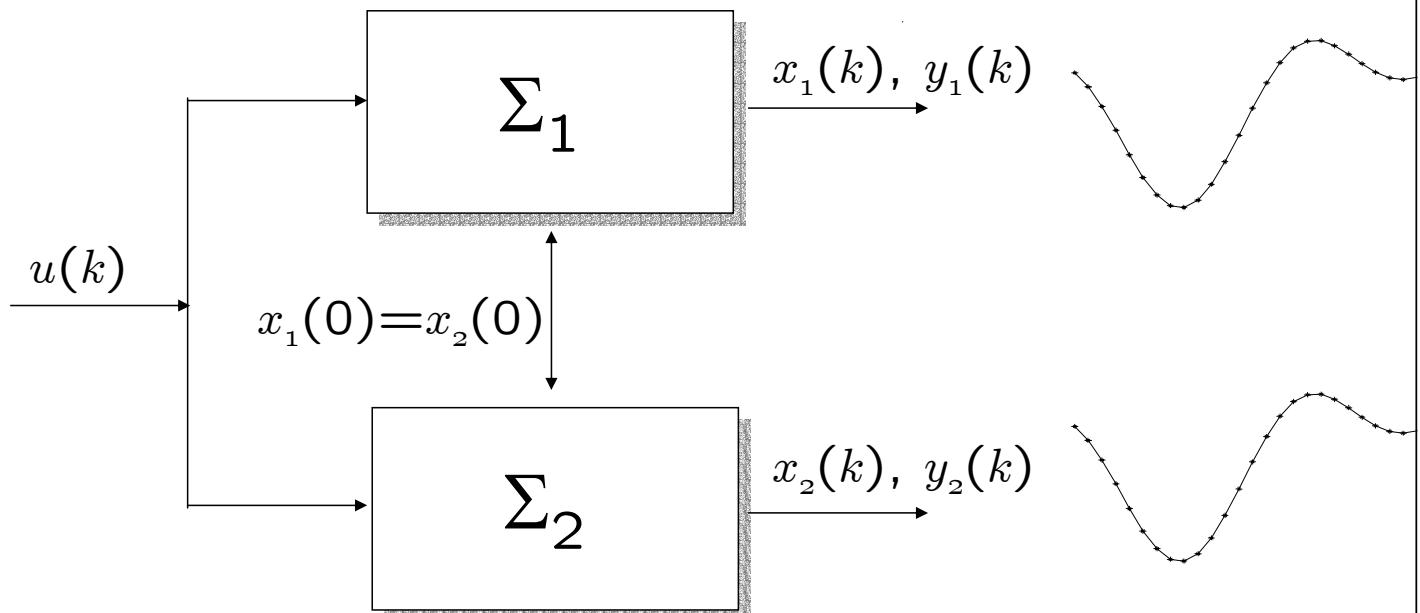
MMPS function: defined by the grammar
 $M := x_i | \alpha | \max(M_1, M_2) | \min(M_1, M_2) | M_1 + M_2 | \beta M_1$

Example: $x(t+1) = 2 \max(x(t), 0) + \min(-\frac{1}{2}u(t), 1)$

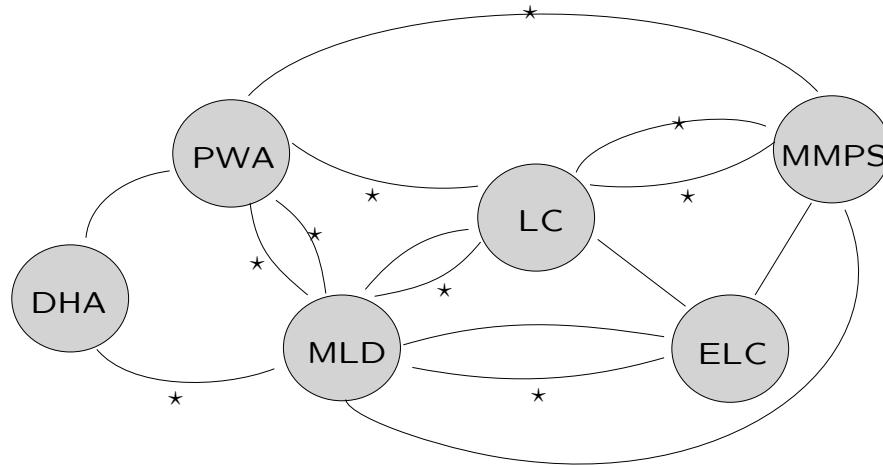
Used for modeling discrete-event systems (t=event counter)

Equivalences of Hybrid Models

Definition 1 Two hybrid systems Σ_1, Σ_2 are equivalent if for all initial conditions $x_1(0) = x_2(0)$ and input $\{u_1(k)\}_{k \in \mathbb{Z}_+} = \{u_2(k)\}_{k \in \mathbb{Z}_+}$ then $x_1(k) = x_2(k)$ and $y_1(k) = y_2(k)$, for all $k \in \mathbb{Z}_+$.



Equivalence Results



Theorem 1 All the above six classes of discrete-time hybrid models are equivalent (possibly under some additional assumptions, such as boundedness of input and state variables)

(Heemels, De Schutter, Bemporad, *Automatica*, 2001)

(Torrisi, Bemporad, *IEEE CST*, 2003)

(Bemporad and Morari, *Automatica*, 1999)

(Bemporad, Ferrari-T., Morari, *IEEETAC*, 2000)

Theoretical properties and analysis/synthesis tools can be transferred from one class to another

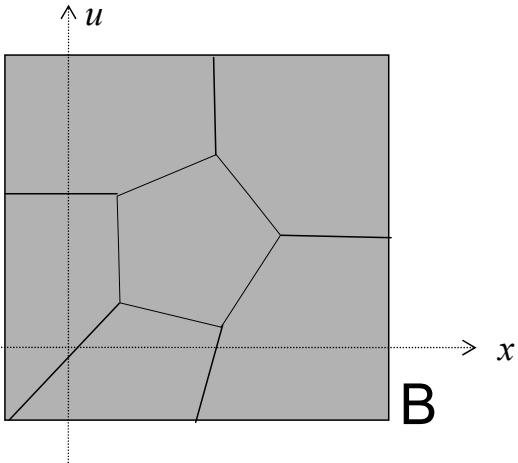
Efficient MLD to PWA Conversion

- Proof is constructive: given an MLD system it returns its equivalent PWA form
- Drawback: it needs the enumeration of all possible combinations of binary states, binary inputs, and δ variables
- Most of such combinations lead to empty regions Ω
- IDEA: use techniques from multiparametric programming to avoid such an enumeration

MLD to PWA Conversion Algorithm

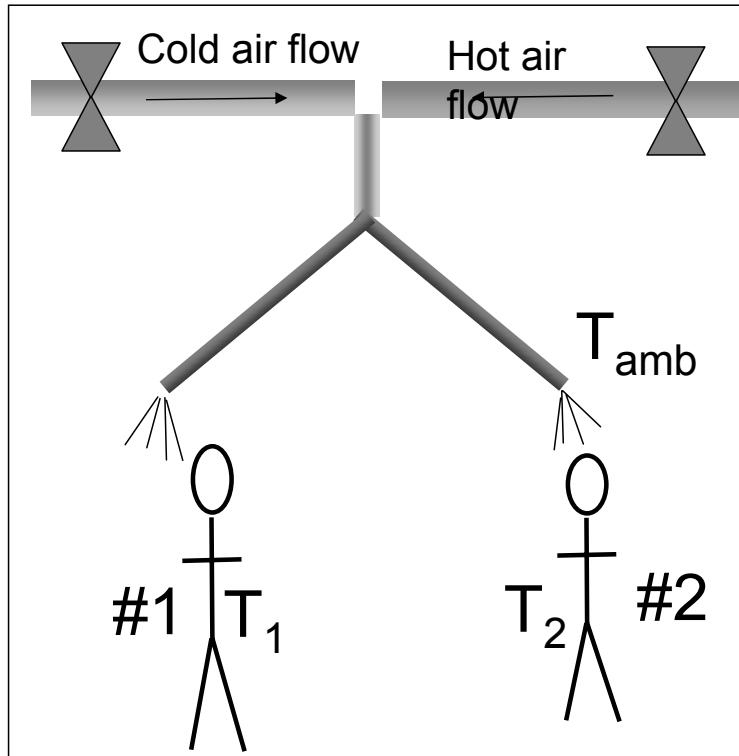
(Bemporad, 2002)

GOAL: split a given set B of states+inputs into polyhedral cells and find the affine dynamics in each cell, therefore determining the PWA system which is equivalent to the given MLD system



Note: the cells Ω_i are embedded in \mathbb{R}^{n+m} by treating integer states and integer inputs as continuous vars:
 $x_b, u_b \in [0,1] \rightarrow x_b, u_b \in [-1/2, 1/2) \cup [1/2, 3/2]$

Example: Heat and Cool



Rules of the game:

- #1 turns the heater (air conditioning) on whenever he is cold (hot)
- If #2 is cold he turns the heater on, unless #1 is hot
- If #2 is hot he turns the air conditioning on, unless #2 is cold
- Otherwise, heater and air conditioning are off

- $\dot{T}_1 = -\alpha_1(T_1 - T_{\text{amb}}) + k_1(u_{\text{hot}} - u_{\text{cold}})$ (body temperature dynamics of #1)
- $\dot{T}_2 = -\alpha_2(T_2 - T_{\text{amb}}) + k_2(u_{\text{hot}} - u_{\text{cold}})$ (body temperature dynamics of #2)

Hybrid HYSDEL Model

```

/* Heat and Cool example
(C) 2002 by A. Bemporad, Las Vegas, December 9, 2002 */

SYSTEM heatcool {
    INTERFACE {
        STATE { REAL T1 [-10,50];
                 REAL T2 [-10,50];
            }
        INPUT { REAL Tamb [-10,50];
            }
        PARAMETER {
            REAL Ts = .5; /* sampling time, second
            REAL alpha1 = 1;
            REAL alpha2 = 0.5;
            REAL k1 = .8;
            REAL k2 = .4;
            REAL Thot1 = 30;
            REAL Tcold1 = 15;
            REAL Thot2 = 35;
            REAL Tcold2 = 10;
            REAL Uc = 2; /* cold water flow */
            REAL Uh = 2; /* hot water flow */
        }
    }
    IMPLEMENTATION {
        AUX { REAL uhot, ucold;
              BOOL hot1, hot2, cold1, cold2;
            }
        AD { hot1 = T1>=Thot1;
              hot2 = T2>=Thot2;
              cold1 = T1<=Tcold1;
              cold2 = T2<=Tcold2;
            }
        DA { uhot = {IF cold1 | (cold2 & ~hot1) THEN Uh ELSE 0};
              ucold = {IF hot1 | (hot2 & ~cold1) THEN Uc ELSE 0};
            }
        CONTINUOUS { T1 = T1+Ts*(-alpha1*(T1-Tamb)+k1*(uhot-ucold));
                     T2 = T2+Ts*(-alpha2*(T2-Tamb)+k2*(uhot-ucold));
            }
    }
}

```

Hybrid MLD Model

- MLD model

$$\begin{aligned}
 x(t+1) &= Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) \\
 y(t) &= Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t) \\
 E_2\delta(t) + E_3z(t) &\leq E_1u(t) + E_4x(t) + E_5
 \end{aligned}$$

- 2 continuous states: (T_1, T_2)
- 1 continuous input: (T_{amb})
- 2 auxiliary continuous vars: (u_{hot}, u_{cold})
- 6 auxiliary binary vars: $(4 \text{ thresholds} + 2 \text{ for OR condition})$
- 20 mixed-integer inequalities

Possible combination of integer variables: $2^6 = 64$

Hybrid PWA Model

- PWA model

$$\begin{aligned}x'(k) &= A_{i(k)}x(k) + B_{i(k)}u(k) + f_{i(k)} \\y(k) &= C_{i(k)}x(k) + D_{i(k)}u(k) + g_{i(k)} \\i(k) \text{ s.t. } &H_{i(k)}x(k) + J_{i(k)}u(k) \leq K_{i(k)}\end{aligned}$$

- 2 continuous states:

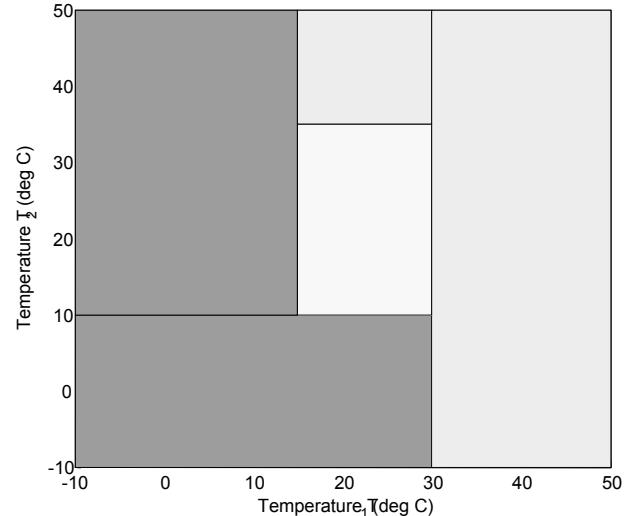
(temperatures T_1, T_2)

- 1 continuous input:

(room temperature T_{amb})

- 5 polyhedral regions

(partition does not depend on input)



Computation time: 0.72 s in Matlab

$$\begin{cases} u_{\text{hot}} = 0 \\ u_{\text{cold}} = 0 \end{cases}$$

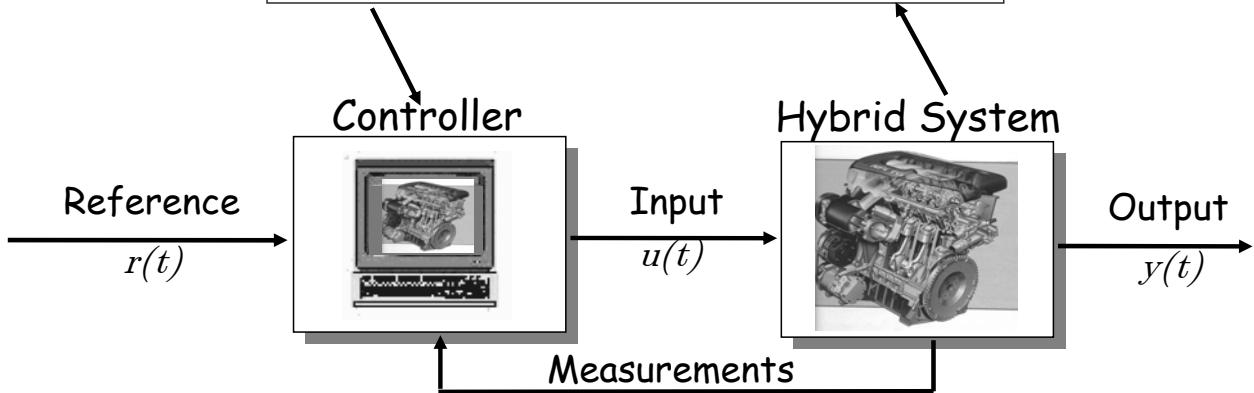
$$\begin{cases} u_{\text{hot}} = 0 \\ u_{\text{cold}} = \bar{U}_C \end{cases}$$

$$\begin{cases} u_{\text{hot}} = \bar{U}_H \\ u_{\text{cold}} = 0 \end{cases}$$

Controller Synthesis

Model Predictive Control of Hybrid Systems

$$\begin{aligned}x(t+1) &= Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) \\y(t) &= Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t) \\E_2\delta(t) + E_3z(t) &\leq E_4x(t) + E_1u(t) + E_5\end{aligned}$$



- MODEL: a model of the plant is needed to predict the future behavior of the plant
- PREDICTIVE: optimization is based on the predicted future evolution of the plant
- CONTROL: control complex constrained multivariable plants

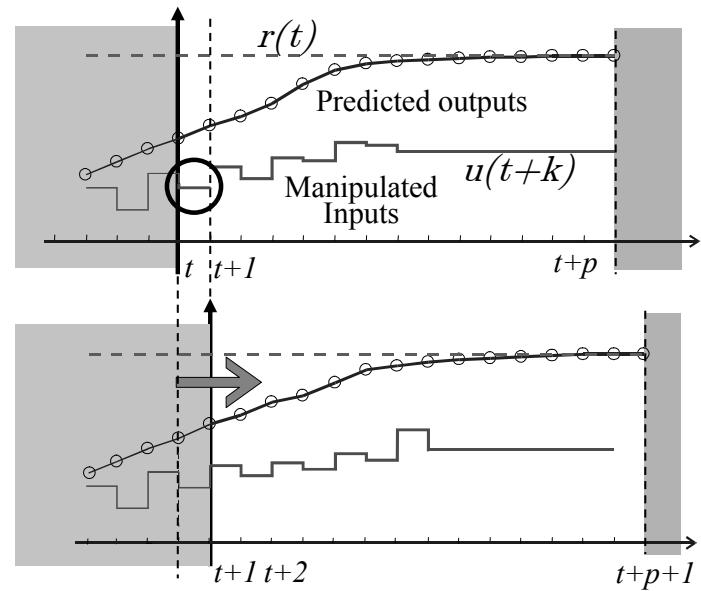
Receding Horizon Philosophy

- At time t :

Solve an optimal control problem over a finite future horizon p :

- minimize $|y - r| + \rho|u|$
- subject to constraints

$$\begin{aligned}u_{min} \leq u \leq u_{max} \\y_{min} \leq y \leq y_{max}\end{aligned}$$

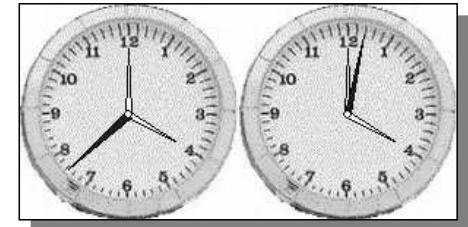


- Only apply the first optimal move $u^*(t)$
- Get new measurements, and repeat the optimization at time $t + 1$

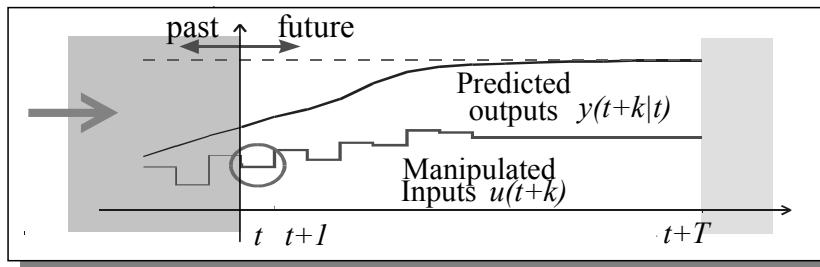
Advantage of on-line optimization: **FEEDBACK!**

Receding Horizon - Example

MPC is like playing chess !



MPC for Hybrid Systems



Model
Predictive (MPC)
Control

- At time t solve with respect to $U \triangleq \{u(t), u(t+1), \dots, u(t+T-1)\}$ the finite-horizon open-loop, optimal control problem:

$$\begin{aligned} \min_U J(U, x(t), r) \triangleq & \sum_{k=0}^{T-1} \|Q(y(t+k+1|t) - r)\| + \|R(u(t+k) - u_r)\| \\ & + \sigma (\|\delta(t+k|t) - \delta_r\| + \|z(t+k|t) - z_r\| + \|x(t+k|t) - x_r\|) \end{aligned}$$

subj. to $\left\{ \begin{array}{l} \text{MLD model} \\ x(t|t) = x(t) \\ x(t+T|t) = x_r \end{array} \right.$

- Apply only $u(t) = u^*(t)$ (discard the remaining optimal inputs)
- Repeat the whole optimization at time $t+1$

Closed-Loop Stability

Theorem 1 Let $(x_r, u_r, \delta_r, z_r)$ be the equilibrium values corresponding to the set point r , and assume $x(0)$ is such that the MPC problem is feasible at time $t = 0$. Then $\forall Q, R > 0$, $\forall \sigma > 0$, the MPC controller stabilizes the MLD system

$$\begin{aligned}\lim_{t \rightarrow \infty} y(t) &= r \\ \lim_{t \rightarrow \infty} u(t) &= u_r\end{aligned}$$

$\lim_{t \rightarrow \infty} x(t) = x_r$, $\lim_{t \rightarrow \infty} \delta(t) = \delta_r$, $\lim_{t \rightarrow \infty} z(t) = z_r$, and all constraints are fulfilled.

(Bemporad, Morari 1999)

Proof: Easily follows from standard Lyapunov arguments

Stability Proof

- Assume we set the terminal constraint $x(T) = x_r$ in the optimal control problem
 - Let \mathcal{U}_t^* denote the optimal control sequence $\{u_t^*(0), \dots, u_t^*(T-1)\}$
 - Let $V(t) \triangleq J(\mathcal{U}_t^*, x(t))$ = value function
 - By construction, $\mathcal{U}_1 = \{u_t^*(1), \dots, u_t^*(T-1), u_r\}$ is feasible @ $t+1$
 - Hence,
- $$V(t+1) \leq J(\mathcal{U}_1, x(t+1)) = V(t) - \|y(t) - r\|_Q - \|u(t) - u_r\|_R - \sigma(\|\delta(t) - \delta_r\| - \|z(t) - z_r\| - \|x(t) - x_r\|)$$
- Hence $V(t)$ is decreasing and lower-bounded by 0 $\Rightarrow \exists V_\infty = \lim_{t \rightarrow \infty} V(t) \Rightarrow V(t+1) - V(t) \rightarrow 0$
 - Hence, $\|y(t) - r\|_Q \rightarrow 0, \|u(t) - u_r\|_R \rightarrow 0, \dots, \|x(t) - x_r\| \rightarrow 0$

Hybrid MPC - Example

Switching System:

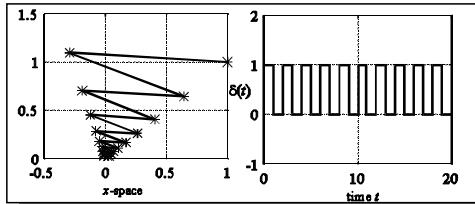
$$x(t+1) = 0.8 \begin{bmatrix} \cos \alpha(t) & -\sin \alpha(t) \\ \sin \alpha(t) & \cos \alpha(t) \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

$$y(t) = [0 \ 1] x(t)$$

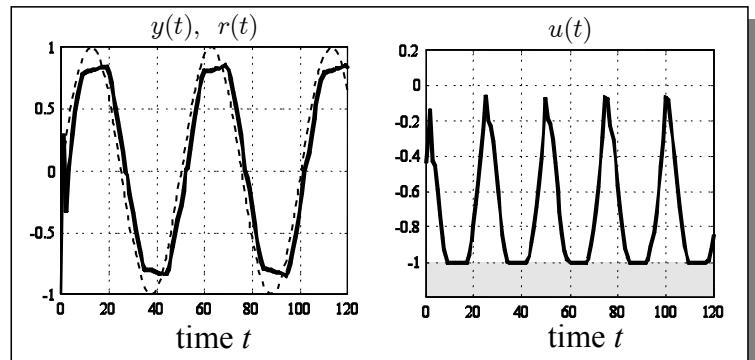
$$\alpha(t) = \begin{cases} \frac{\pi}{3} & \text{if } [1 \ 0]x(t) \geq 0 \\ -\frac{\pi}{3} & \text{if } [1 \ 0]x(t) < 0 \end{cases}$$

Constraint: $-1 \leq u(t) \leq 1$

Open loop:



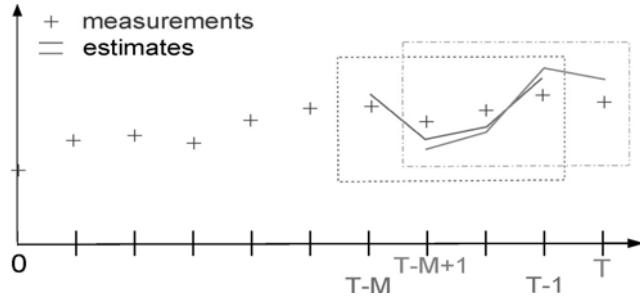
Closed loop:



State Estimation / Fault Detection

State Estimation / Fault Detection

- Problem: given past output measurements and inputs, estimate the current state/faults
- Solution: Use Moving Horizon Estimation for MLD systems (dual of MPC)



At each time t solve the optimization:

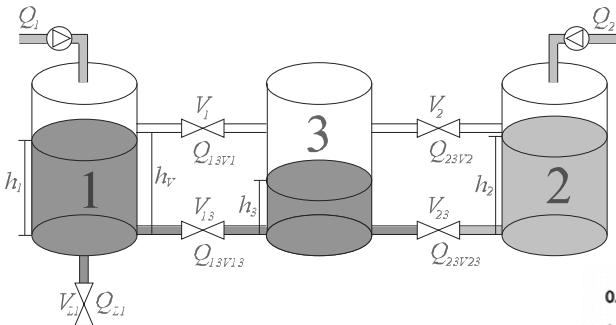
$$\min \sum_{k=1}^T \|\hat{y}(t - k|t) - y(t - k)\|^2 + \dots$$

and get estimates $\hat{x}(t)$

→ MHE optimization = MIQP (Bemporad, Mignone, Morari, ACC 1999)
 → Convergence can be guaranteed (Ferrari-T., Mignone, Morari, 2002)

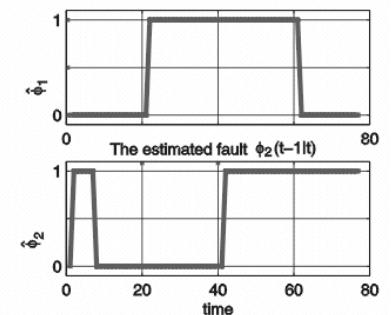
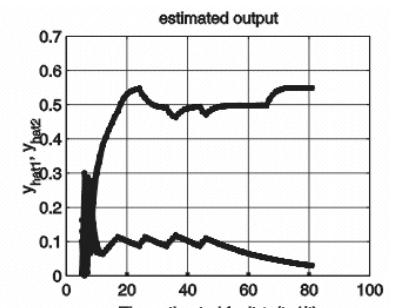
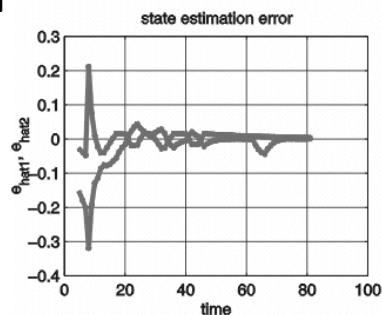
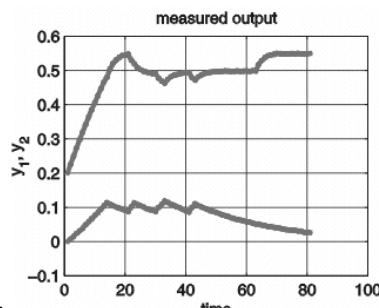
Fault detection: augment MLD with unknown **binary** disturbances $\phi \in \{0,1\}^p$

Example: Three Tank System

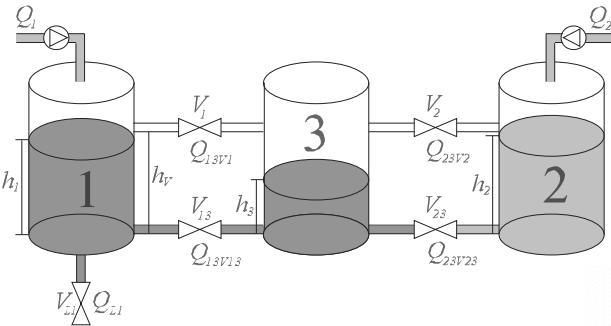


 COSY Benchmark problem, ESF

- ϕ_1 : leak in tank 1 for $20s \leq t \leq 60s$
- ϕ_2 : valve V_1 blocked for $t \geq 40s$

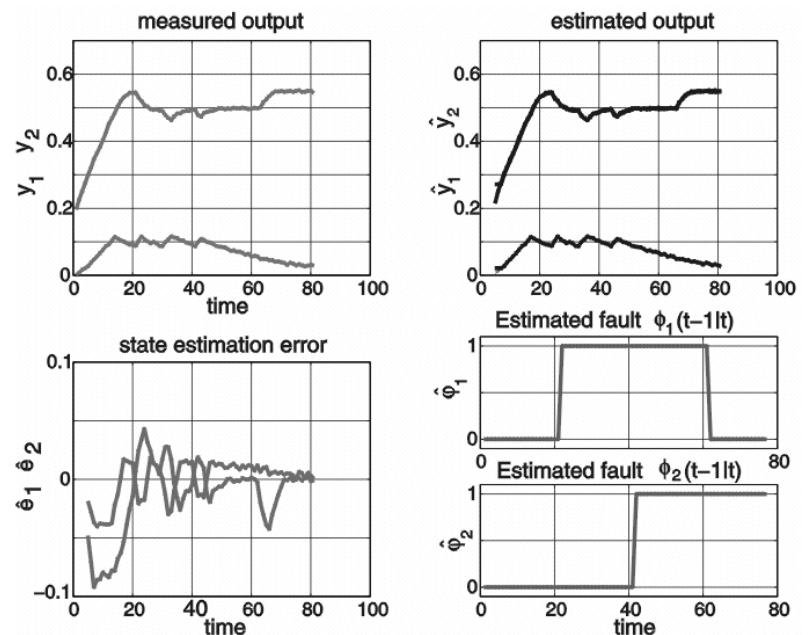


Example: Three Tank System



COSY Benchmark problem, ESF

- ϕ_1 : leak in tank 1
for $20s \leq t \leq 60s$
- ϕ_2 : valve V_1 blocked
for $t \geq 40s$
- Add logic constraint
 $[h_1 \leq h_v] \Rightarrow \phi_2 = 0$



Optimal Control of Hybrid Systems: Computational Aspects

MIQP Formulation of MPC

(Bemporad, Morari, 1999)

$$\min \sum_{k=0}^{T-1} \left\{ y'(t+k+1|t)Qy(t+k+1|t) + u'(t+k)Ru(t+k) \right\}$$

s.t. MLD dynamics

- Set $\xi \triangleq [u(t), \dots, u(t+T-1), \delta(t|t), \dots, \delta(t+T-1|t), z(t|t), \dots, z(t+T-1|t)]$



$$\begin{aligned} \min_{\xi} J(\xi, x(t)) &= \frac{1}{2} \xi' H \xi + x'(t) F \xi + \frac{1}{2} x'(t) Y x(t) \\ \text{s.t. } G \xi &\leq W + S x(t) \end{aligned}$$

Mixed Integer Quadratic Program (MIQP)

$$\begin{aligned} u &\in \mathbf{R}^{n_u} \\ \delta &\in \{0, 1\}^{n_\delta} \\ z &\in \mathbf{R}^{n_z} \end{aligned}$$



$$\xi \in \mathbf{R}^{T(n_u+n_z)} \times \{0, 1\}^{Tn_\delta}$$

MILP Formulation of MPC

(Bemporad, Borrelli, Morari, 2000)

$$\begin{aligned} \min \sum_{k=0}^{T-1} \|Qy(t+k+1|t)\|_\infty + \|Ru(t+k)\|_\infty \\ \text{s.t. MLD dynamics} \end{aligned}$$

- Introduce slack variables: $\min |x| \rightarrow$

$$\begin{aligned} \min \epsilon \\ \text{s.t. } \epsilon \geq x \\ \epsilon \geq -x \end{aligned}$$

$$\begin{aligned} \epsilon_k^x &\geq \|Qy(t+k|t)\|_\infty & i = 1, \dots, p, \quad k = 1, \dots, T-1 \\ \epsilon_k^u &\geq \|Ru(t+k)\|_\infty & i = 1, \dots, p, \quad k = 1, \dots, T-1 \\ \epsilon_k^x &\geq -[Qy(t+k|t)]_i & i = 1, \dots, p, \quad k = 1, \dots, T-1 \\ \epsilon_k^u &\geq [Ru(t+k)]_i & i = 1, \dots, m, \quad k = 0, \dots, T-1 \\ \epsilon_k^u &\geq -[Ru(t+k)]_i & i = 1, \dots, m, \quad k = 0, \dots, T-1 \end{aligned}$$

- Set $\xi \triangleq [\epsilon_1^x, \dots, \epsilon_{N_y}^x, \epsilon_1^u, \dots, \epsilon_{T-1}^u, U, \delta, z]$

Mixed Integer Linear Program
(MILP)

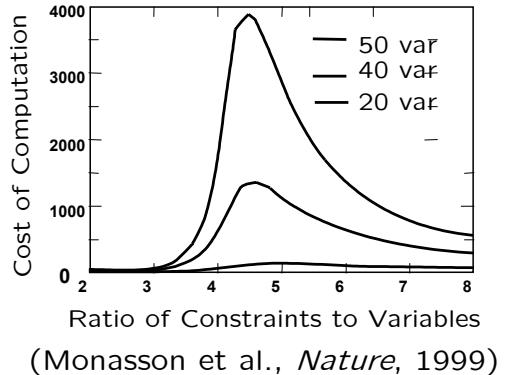
$$\begin{aligned} \min_{\xi} J(\xi, x(t)) &= \sum_{k=0}^{T-1} \epsilon_i^x + \epsilon_i^u \\ \text{s.t. } G \xi &\leq W + S x(t) \end{aligned}$$

Mixed-Integer Program Solvers

- Mixed-Integer Programming is NP -hard

Phase transitions have been found in computationally hard problems.

BUT



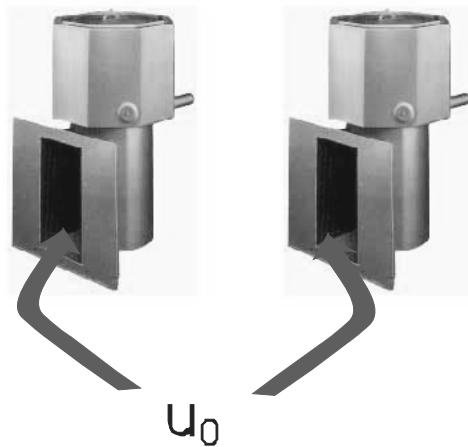
- General purpose Branch & Bound/Branch & Cut solvers available for MILP and MIQP (Ilog CPLEX, XPRESS-MP, Sahinidis, ...)
More solvers and benchmarks: <http://plato.la.asu.edu/bench.html>
- No need to reach global optimum (see proof of the theorem), although performance deteriorates

Good for large sampling times (e.g., 1 h) / expensive hardware ...
... but not for fast sampling (e.g. 10 ms) / cheap hardware !

Hybrid Control Example: Heat Exchange

Hybrid Control - An Example

(Hedlund and Rantzer, CDC1999)



$$\begin{bmatrix} \dot{T}_1 \\ \dot{T}_2 \end{bmatrix} = \begin{bmatrix} -1 & 0 \\ 0 & -2 \end{bmatrix} \begin{bmatrix} T_1 \\ T_2 \end{bmatrix} + B_i u_0$$

$$B_i = \begin{cases} \begin{bmatrix} 1 \\ 0 \end{bmatrix} & \text{if first furnace heated} \\ \begin{bmatrix} 0 \\ 1 \end{bmatrix} & \text{if second furnace heated} \\ \begin{bmatrix} 0 \\ 0 \end{bmatrix} & \text{if no heating} \end{cases}$$

- Objective:
 - Control the temperature to a given set-point T_1, T_2
- Constraints:
 - Only three operation modes:
 - 1- Heat only the first furnace
 - 2- Heat only the second furnace
 - 3- Do not heat any furnaces

Amount of heating power u_0 is constant

Alternate Heating of Two Furnaces

- HYSDEL model:

```
/* HEAT EXCHANGE model - (C) 2001 by A. Bemporad, Zurich, March 13, 2001 */

SYSTEM furnaces {

INTERFACE {
    STATE {
        REAL t1,t2,u0; }
    INPUT {
        BOOL heat1,heat2; }
    PARAMETER {
        REAL Ts=0.08; /* sampling time, seconds */
        REAL Bd1=0.07688365361336; /* discretization of B matrix, heat 1 active */
        REAL Bd2=0.07392810551689; /* discretization of B matrix, heat 2 active */
        REAL A11=.9231; /* discretization of A matrix */
        REAL A22=.8521; /* discretization of A matrix */

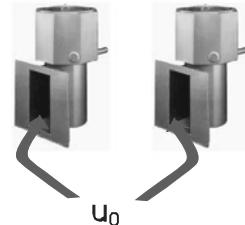
        REAL u0Max=10; /* upperbound on u0 */
        REAL e = 1e-6; /* precision for strict inequalities */ }
}

IMPLEMENTATION {
    AUX (REAL z1,z2; )

    DA {
        z1 = {IF heat1 THEN Bd1*u0 [Bd1*u0Max,0,e]};
        z2 = {IF heat2 THEN Bd2*u0 [Bd2*u0Max,0,e]}; }

    CONTINUOUS { t1 = A11*t1+z1;
                 t2 = A22*t2+z2;
                 u0 = u0; }

    MUST { ~heat1 | ~heat2; /* heat1 and heat2 cannot be both active */ }
}
}
```

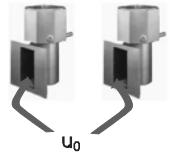


- MLD model:

$$x(t+1) = \begin{bmatrix} 0.9231 & 0 & 0 \\ 0 & 0.8521 & 0 \\ 0 & 0 & 1.0000 \end{bmatrix} x(t) + \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix} z(t)$$

$$\begin{bmatrix} -1 & 0 \\ 1 & 0 \\ -1 & 0 \\ 1 & 0 \\ 0 & -1 \\ 0 & 1 \\ 0 & -1 \\ 0 & 1 \\ 0 & 0 \end{bmatrix} z(t) \square \begin{bmatrix} -0.7688 & 0 \\ 0 & 0 \\ 0.7688 & 0 \\ 0 & -0.7393 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ -1.0000 & -1.0000 \end{bmatrix} u(t) + \begin{bmatrix} -0.0769 \\ 0.0769 \\ 0 \\ 0 \\ -0.0739 \\ 0.0739 \\ 0 \\ 0 \\ 0 \end{bmatrix} x_3(t) + \begin{bmatrix} 0.7688 \\ 0 \\ 0 \\ 0 \\ 0.7393 \\ 0 \\ 0 \\ 0 \\ 1.0000 \end{bmatrix}$$

Alternate Heating of Two Furnaces



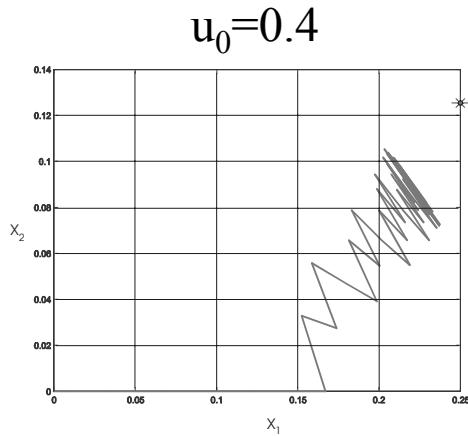
- Performance index

$$\min_{\{u(t), u(t+1), u(t+2)\}} \sum_{k=0}^2 \|R(u(t+k+1) - u(t+k))\|_\infty + \|Q(x(t+k|t) - x_e)\|_\infty$$

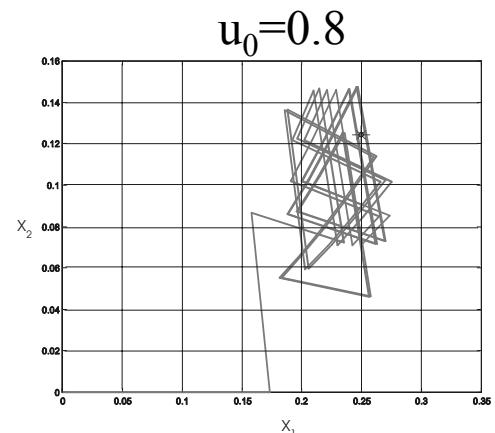
- Constraints

$$\begin{array}{lcl} -1 & \leq & x_1 \leq & 1 \\ -1 & \leq & x_2 \leq & 1 \end{array}$$

Closed-Loop Behavior



Set point cannot be reached



Set point is reached

- Computational complexity of on-line MILP

(Bemporad, Borrelli, Morari, 2000)

Linear constraints	168
Continuous variables	33
Binary variables	6
Parameters	3
Time to solve MILP (av.)	1,09 s

On-Line vs. Off-Line Optimization

$$\min_U J(U, \underline{x}(t)) \triangleq \sum_{k=0}^{T-1} \|Qy(t+k+1|t)\|_\infty + \|Ru(t+k)\|_\infty$$

subj. to $\begin{cases} \text{MLD model} \\ x(t|t) = \underline{x}(t) \\ x(t+T|t) = 0 \end{cases}$

- On-line optimization: given $x(t)$, solve the problem at each time step t

Mixed-Integer Linear Program (MILP)

- Off-line optimization: solve the MILP for all $x(t)$

$$\begin{aligned} \min_{\xi} J(\xi, \underline{x}(t)) &\triangleq f'\xi \\ \text{s.t. } G\xi &\leq W + F\underline{x}(t) \end{aligned}$$

multi-parametric Mixed Integer Linear Program (mp-MILP)

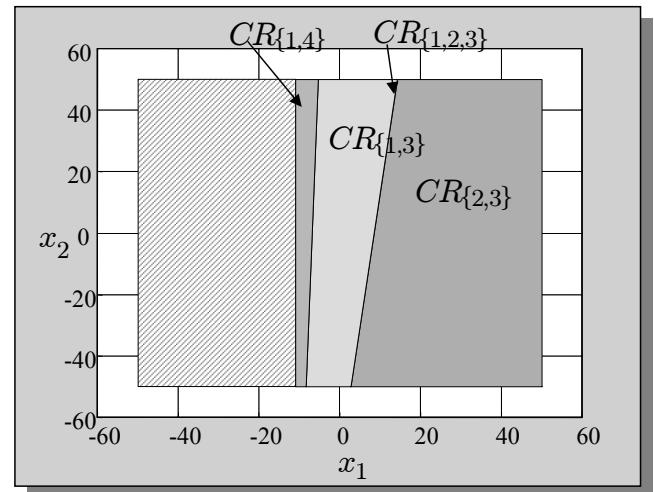
Explicit Form of
Model Predictive Control

via Multiparametric Programming

Example of Multiparametric Solution

Multiparametric LP ($\xi \in \mathbf{R}^2$)

$$\begin{array}{ll} \min_{\xi} & -3\xi_1 - 8\xi_2 \\ \text{s.t.} & \left\{ \begin{array}{lcl} \xi_1 + \xi_2 & \leq & 13 + x_1 \\ 5\xi_1 - 4\xi_2 & \leq & 20 \\ -8\xi_1 + 22\xi_2 & \leq & 121 + x_2 \\ -4\xi_1 - \xi_2 & \leq & -8 \\ -\xi_1 & \leq & 0 \\ -\xi_2 & \leq & 0 \end{array} \right. \end{array}$$



$$\xi(x) = \begin{cases} \begin{bmatrix} 1.33 & 0 \\ -0.33 & 0.00 \end{bmatrix} x + \begin{bmatrix} 1.41 \\ -1.41 \end{bmatrix} \quad \text{if } \begin{bmatrix} -0.00 \\ 0.12 \\ 0.00 \\ 0.00 \\ 0.00 \\ 0.03 \end{bmatrix} x \leq \begin{bmatrix} 1.00 \\ -1.00 \\ 1.00 \\ 1.00 \end{bmatrix} \quad \text{CB}^{\{1,4\}} \\ \begin{bmatrix} 0.33 & 0.03 \\ 0.13 & -0.03 \end{bmatrix} x + \begin{bmatrix} 1.20 \\ -1.20 \end{bmatrix} \quad \text{if } \begin{bmatrix} -0.12 \\ 0.15 \\ 0.00 \\ 0.00 \\ 0.00 \\ 0.03 \end{bmatrix} x \leq \begin{bmatrix} 1.00 \\ 1.00 \\ 1.00 \\ 1.00 \end{bmatrix} \quad \text{CB}^{\{1,3\}} \\ \begin{bmatrix} 0.00 & 0.00 \\ 0.00 & 0.02 \end{bmatrix} x + \begin{bmatrix} 0.80 \\ -1.80 \end{bmatrix} \quad \text{if } \begin{bmatrix} -0.13 \\ 0.00 \\ 0.00 \\ 0.03 \\ 0.00 \end{bmatrix} x \leq \begin{bmatrix} 1.00 \\ 1.00 \\ 1.00 \\ 1.00 \end{bmatrix} \quad \text{CB}^{\{3,3\}} \end{cases}$$

Linear MPC

- Linear Model:

$$\begin{aligned} x(t+1) &= Ax(t) + Bu(t) \\ y(t) &= Cx(t) \end{aligned}$$

$$\Delta u(t) = u(t) - u(t-1) \quad \text{input increments (for integral action)}$$

- Prediction:

$$y(t+k|t) = CA^k x(t) + \sum_{i=0}^{k-1} CA^i Bu(t+k-1-i)$$

$$u(t+k) = u(t-1) + \sum_{i=0}^{k-1} \Delta u(t+i)$$

Linear MPC

Quadratic Performance Index

$$U \triangleq \{\Delta u(t), \dots, \Delta u(t + m - 1)\}$$

$$\begin{aligned} \min_U \sum_{k=0}^{p-1} & \|W^y [y(t+k+1|t) - r(t)]\|^2 \\ & + \|W^{\Delta u} \Delta u(t+k)\|^2 + W^u [u(t+k) - u_{\text{target}}(t)]\|^2 \end{aligned}$$

→ $\min_U J(U, x(t)) = \frac{1}{2} U' H U + [x'(t) \ r'(t)] F U + \frac{1}{2} [x'(t) \ r'(t)] Y \begin{bmatrix} x(t) \\ r(t) \end{bmatrix}$

subject to $GU \leq W + K \begin{bmatrix} x(t) \\ r(t) \end{bmatrix}$

(convex)
**QUADRATIC
PROGRAM
(QP)**

$$\begin{aligned} y_{\min} &\leq y(t+k|t) \leq y_{\max}, \quad k = 1, \dots, p \\ u_{\min} &\leq u(t+k) \leq u_{\max}, \quad k = 0, 1, \dots, m-1 \\ \Delta u_{\min} &\leq \Delta u(t+k) \leq \Delta u_{\max}, \quad k = 0, 1, \dots, m-1 \end{aligned}$$

Constraints

Multiparametric Quadratic Programming

(Bemporad et al., 2002)

$$\begin{aligned} \min_U & \frac{1}{2} U' H U + x F' U + \frac{1}{2} x' X x \\ \text{subj. to} & GU \leq W + Sx, \end{aligned}$$

$$U \triangleq [u'_0 \ \dots \ u'_{T-1}]'$$

$$U \in \mathbb{R}^r, \quad r \triangleq n_u T$$

$$x \in \mathbb{R}^n$$

- Objective: solve the QP for all $x \in X \subseteq \mathbb{R}^n$

Linearity of the Solution

- $x_0 \in X$
- solve QP to find $U^*(x_0), \lambda^*(x_0)$
 - identify active constraints at $U^*(x_0)$
 - form matrices $\tilde{G}, \tilde{W}, \tilde{S}$ by collecting active constraints $\tilde{G}U^*(x_0) - \tilde{W} - \tilde{S}x_0 = 0$

KKT optimality conditions:

$$\begin{aligned} (1) \quad & HU + Fx + G'\lambda = 0, & (2) \quad & \tilde{G}U - \tilde{W} - \tilde{S}x = 0 \\ (3) \quad & \lambda_i(G^i U - W^i - s^i x) = 0, & (4) \quad & \tilde{G}U \leq \tilde{W} + \tilde{S}x = 0 \\ (5) \quad & \tilde{\lambda}_i \geq 0, \quad \tilde{\lambda}_i = 0 \end{aligned}$$

From (1) : $U = -H^{-1}(Fx + \tilde{G}'\tilde{\lambda})$

From (2) : $\tilde{\lambda}(x) = -(\tilde{G}H^{-1}\tilde{G}')^{-1}(\tilde{W} + (\tilde{S} + \tilde{G}H^{-1}F)x).$
 $U(x) = H^{-1}[\tilde{G}'(\tilde{G}H^{-1}\tilde{G}')^{-1}(\tilde{W} + (\tilde{S} + \tilde{G}H^{-1}F)x) - Fx]$

→ In some neighborhood of x_0 , λ and U are explicit affine functions of x !

Determining a Critical Region

- Impose primal and dual feasibility:

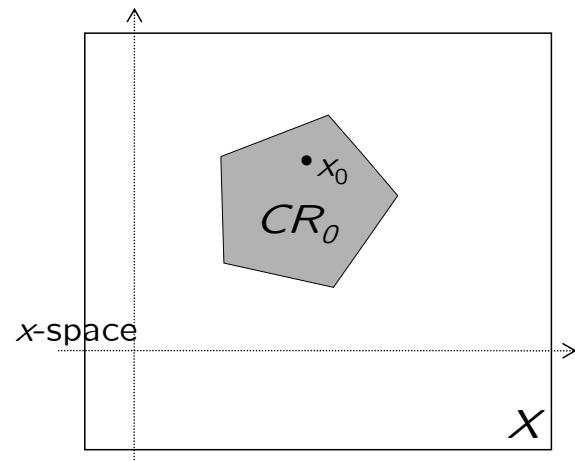
→ linear inequalities in x !

$$\begin{aligned} \tilde{G}U(x) &\leq \tilde{W} + \tilde{S}x \\ \tilde{\lambda}(x) &\geq 0 \end{aligned}$$

- Remove redundant constraints

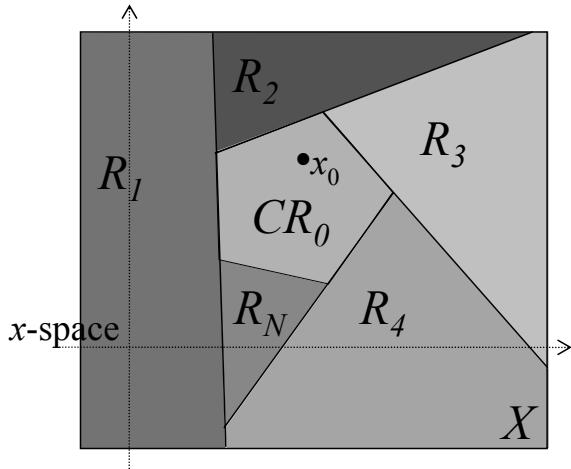
→ critical region CR_0

$$CR_0 = \{Ax \leq B\}$$



- CR_0 is the set of all and only parameters x for which $\tilde{G}, \tilde{W}, \tilde{S}$ is the optimal combination of active constraints at the optimizer

Multiparametric QP



$$CR_0 = \{\mathcal{A}x \leq \mathcal{B}\}$$

$$R_i = \{x \in X : \mathcal{A}^i x > \mathcal{B}^i, \mathcal{A}^j z \leq \mathcal{B}^j, \forall j < i\}$$

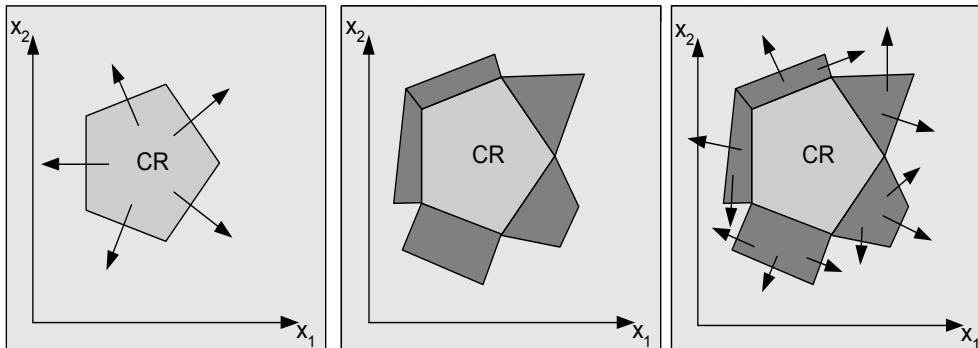
Theorem: $\{CR_0, R_1, \dots, R_N\}$ is a partition of $X \subseteq \mathbf{R}^n$

Proceed iteratively: for each region R_i repeat the whole procedure with $X \leftarrow R_i$

The recursive algorithm terminates after a finite number of steps, because the number of combinations of active constraints is finite

Mp-QP – More efficient method

(Tøndel, Johansen, Bemporad, 2003)



The active set of a neighboring region is found by using the active set of the current region + knowledge of the type of hyperplane we are crossing:

$\hat{G}^i U(x) \leq \hat{W}^i + \hat{S}^i x \Rightarrow$ The corresponding constraint is added to the active set

$\tilde{\lambda}_j(x) \geq 0 \Rightarrow$ The corresponding constraint is withdrawn from the active set

Multiparametric QP

Theorem 1 Consider a multi-parametric quadratic program with $H \succ 0$. The set X^* of parameters x for which the problem is feasible is a polyhedral set, the value function $J^* : X^* \mapsto \mathbb{R}$ is piecewise quadratic, convex and continuous and the optimizer $U^* : X^* \mapsto \mathbb{R}^r$ is piecewise affine and continuous.

$$U^*(x) = \arg \min_U \frac{1}{2} U' H U + x' F' U \quad \begin{array}{l} \text{continuous,} \\ \text{piecewise affine} \end{array}$$

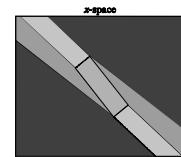
subj. to $Gz \leq W + Sx$

$$V^*(x) = \frac{1}{2} x' Y x + \min_U \frac{1}{2} U' H U + x' F' U \quad \begin{array}{l} \text{convex, continuous,} \\ \text{piecewise quadratic,} \\ C^1 \text{ (if no degeneracy)} \end{array}$$

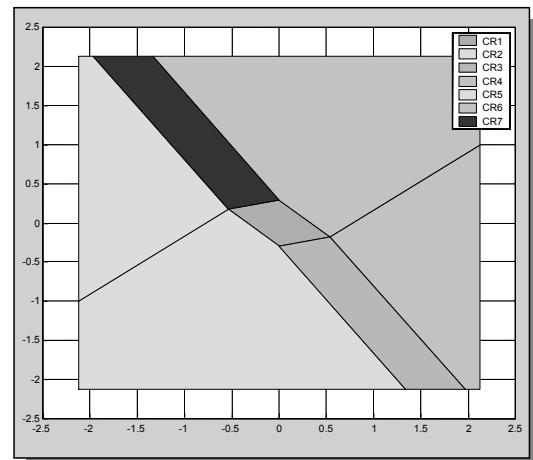
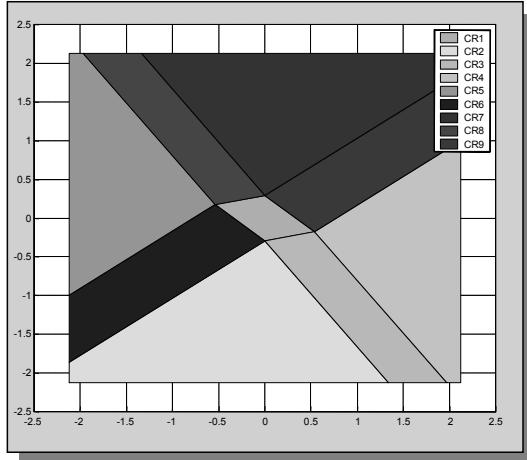
subj. to $Gz \leq W + Sx$

Corollary: The linear MPC controller is a continuous piecewise affine function of the state

$$u(x) = \begin{cases} F_1 x + G_1 & \text{if } H_1 x \leq K_1 \\ \vdots & \vdots \\ F_N x + G_N & \text{if } H_N x \leq K_N \end{cases}$$



Complexity Reduction



$$U(x(0)) \triangleq [u_0(x(0)), \dots, u_{T-1}(x(0))]$$

Regions where the first component of the solution is the same can be joined (when their union is convex). (Bemporad, Fukuda, Torrisi, *Computational Geometry*, 2001)

Double Integrator Example

- System: $y(t) = \frac{1}{s^2} u(t)$ \rightarrow $x(t+1) = \begin{bmatrix} 1 & 1 \\ 0 & 1 \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$
 sampling + ZOH
 $T=1$ s $y(t) = [1 \ 0] x(t)$

- Constraints: $-1 \leq u(t) \leq 1$

- Control objective: minimize $\sum_{t=0}^{\infty} y'(t)y(t) + \frac{1}{100}u^2(t)$
 $u_{t+k} = K_{LQ} x(t+k|t) \quad \forall k \geq N_u$

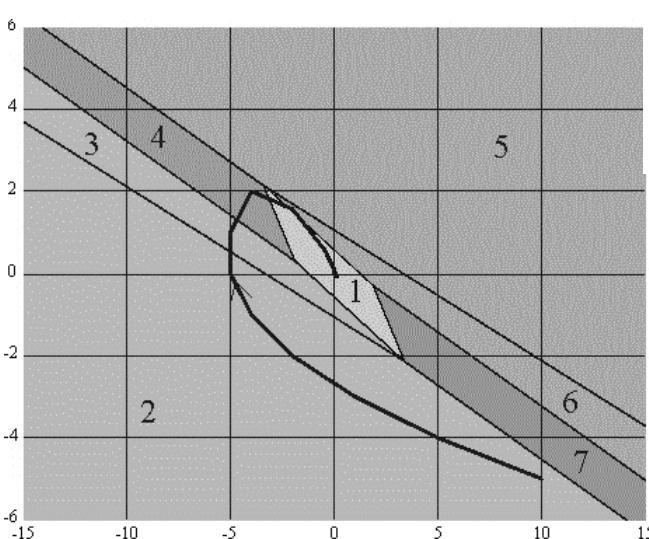
- Optimization problem: for $N_u=2$

$$H = \begin{bmatrix} 0.8365 & 0.3603 \\ 0.3603 & 0.2059 \end{bmatrix}, \quad F = \begin{bmatrix} 0.4624 & 1.2852 \\ 0.1682 & 0.5285 \end{bmatrix} \quad (\text{cost function is normalized by } \max \lambda(H))$$

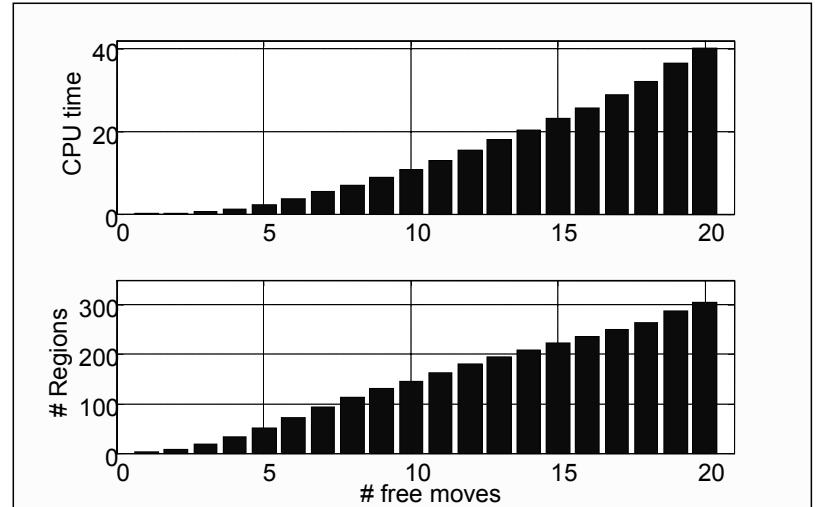
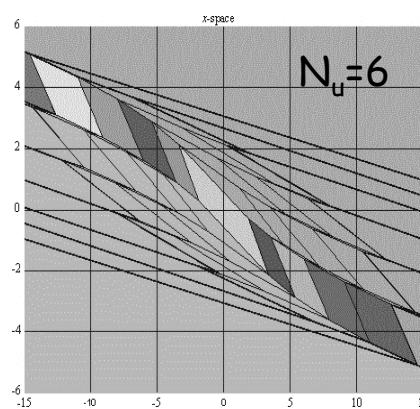
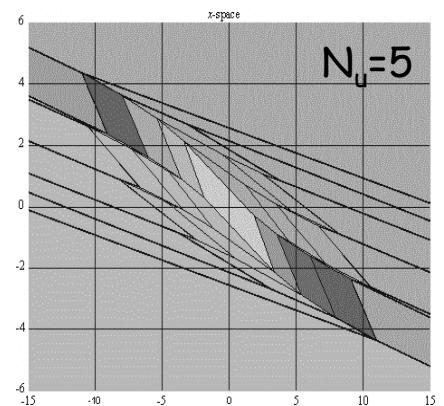
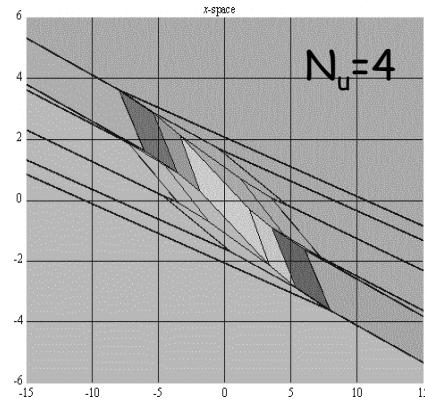
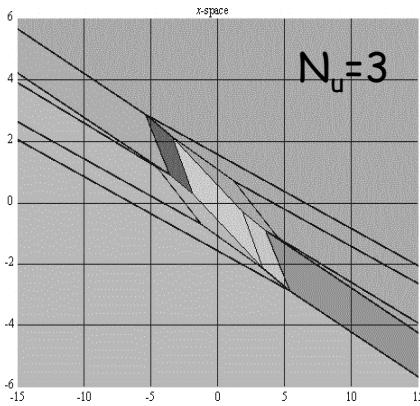
$$G = \begin{bmatrix} 1 & 0 \\ -1 & 0 \\ 0 & 1 \\ 0 & -1 \end{bmatrix}, \quad W = \begin{bmatrix} 1 \\ 1 \\ 1 \\ 1 \end{bmatrix}, \quad S = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 0 & 0 \\ 0 & 0 \end{bmatrix}$$

mp-QP solution

$$N_u=2 \quad u(x) = \begin{cases} [-0.8166 \ -1.7499]x & \text{if } \begin{bmatrix} -0.8166 & -1.7499 \\ 0.8166 & 1.7499 \\ -0.6124 & -0.4957 \end{bmatrix} x \leq \begin{bmatrix} 1.0000 \\ 1.0000 \\ 1.0000 \end{bmatrix} \\ 1.0000 & \text{if } \begin{bmatrix} 0.3864 & 1.0738 \\ 0.2970 & 0.9333 \end{bmatrix} x \leq \begin{bmatrix} -1.0000 \\ -1.0000 \end{bmatrix} \\ 1.0000 & \text{if } \begin{bmatrix} 0.9712 & 2.6991 \\ 0.8166 & 1.7499 \end{bmatrix} x \leq \begin{bmatrix} -1.0000 \\ -1.0000 \end{bmatrix} \\ [-0.5528 \ -1.5364]x + 0.4308 & \text{if } \begin{bmatrix} -0.9712 & -2.6991 \\ 0.3864 & 1.0738 \\ 0.6124 & 0.4957 \end{bmatrix} x \leq \begin{bmatrix} 1.0000 \\ 1.0000 \\ -1.0000 \end{bmatrix} \\ -1.0000 & \text{if } \begin{bmatrix} -0.3864 & -1.0738 \\ -0.2970 & -0.9333 \end{bmatrix} x \leq \begin{bmatrix} -1.0000 \\ -1.0000 \end{bmatrix} \\ -1.0000 & \text{if } \begin{bmatrix} -0.9712 & -2.6991 \\ -0.8166 & -1.7499 \end{bmatrix} x \leq \begin{bmatrix} -1.0000 \\ -1.0000 \end{bmatrix} \\ [-0.5528 \ -1.5364]x - 0.4308 & \text{if } \begin{bmatrix} -0.3864 & -1.0738 \\ 0.9712 & 2.6991 \\ -0.6124 & -0.4957 \end{bmatrix} x \leq \begin{bmatrix} 1.0000 \\ 1.0000 \\ -1.0000 \end{bmatrix} \end{cases} \quad (\text{Region } \#1 \text{ to } \#7)$$



Complexity



Extensions

- Tracking of reference $r(t)$: $\delta u(t) = F(x(t), u(t-1), r(t))$

- Rejection of measured disturbance $v(t)$: $\delta u(t) = F(x(t), u(t-1), v(t))$

- Soft constraints: $u(t) = F(x(t))$

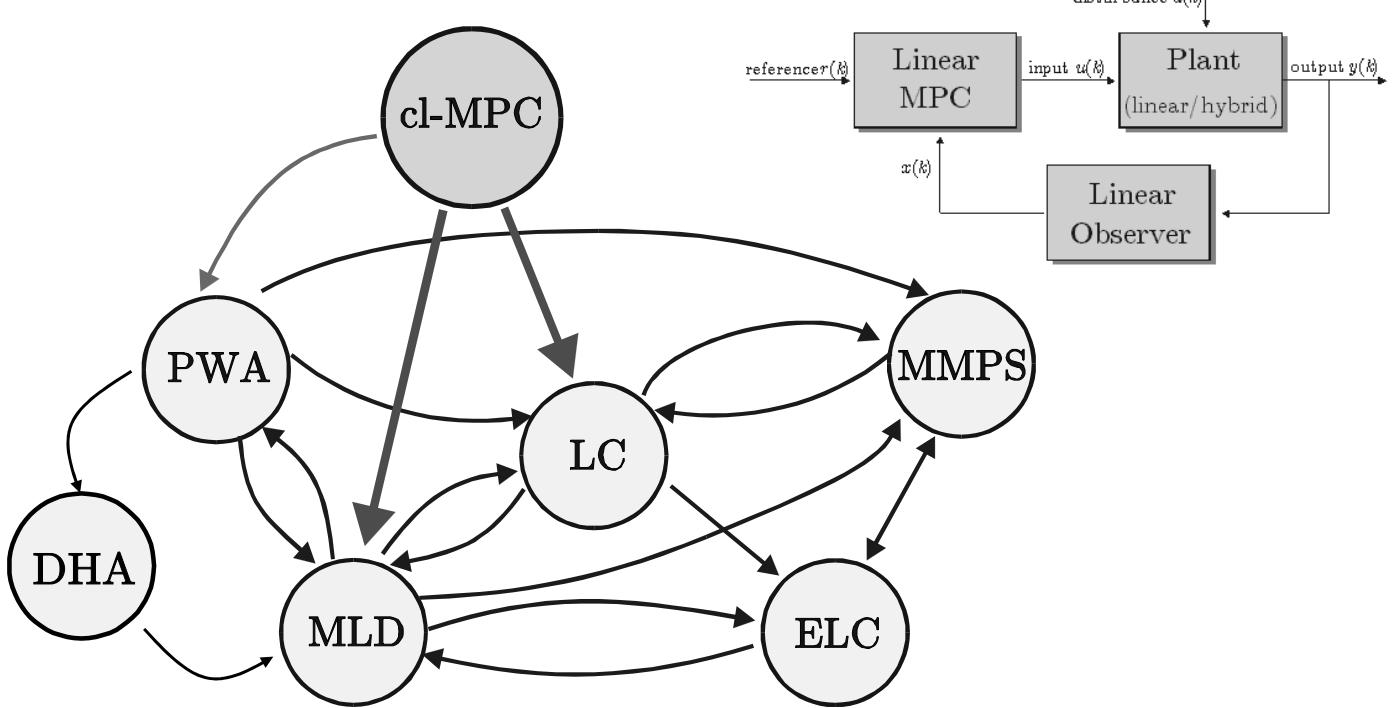
$$y_{\min} - \epsilon \leq y(t+k|t) \leq y_{\max} + \epsilon$$

- Variable constraints: $u(t) = F(x(t), u_{\min}(t), \dots, y_{\max}(t))$

$$\begin{aligned} u_{\min}(t) &\leq u(t+k) \leq u_{\max}(t) \\ y_{\min}(t) &\leq y(t+k|t) \leq y_{\max}(t) \end{aligned}$$

- Linear norms: $\min_U J(U, x(t)) \triangleq \sum_{k=0}^p \|Qy(t+k|t)\|_\infty + \|Ru(t+k)\|_\infty$
(Bemporad, Borrelli, Morari, IEEE TAC, 2002)

Closed-Loop MPC and Hybrid Systems



Motivation:

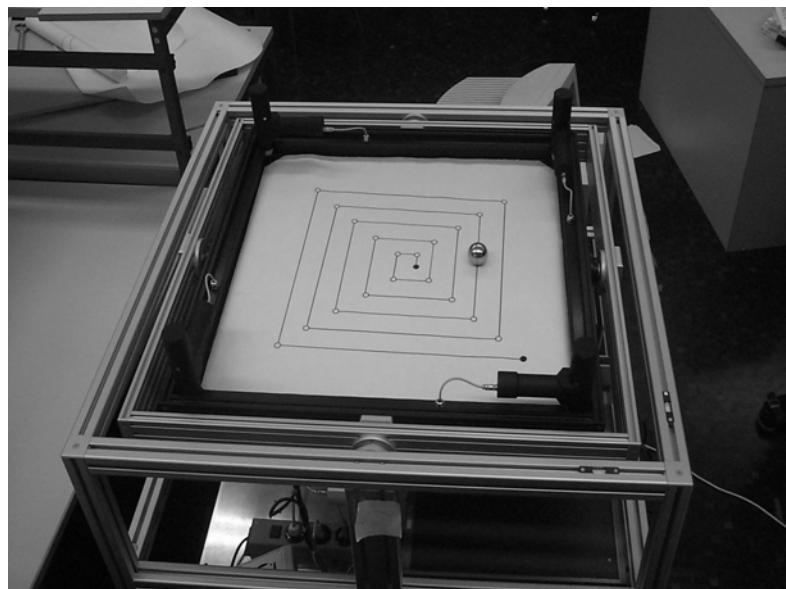
Use hybrid techniques to analyze closed-loop MPC systems !

(Bemporad, Heemels, De Schutter IEEE TAC, 2002)

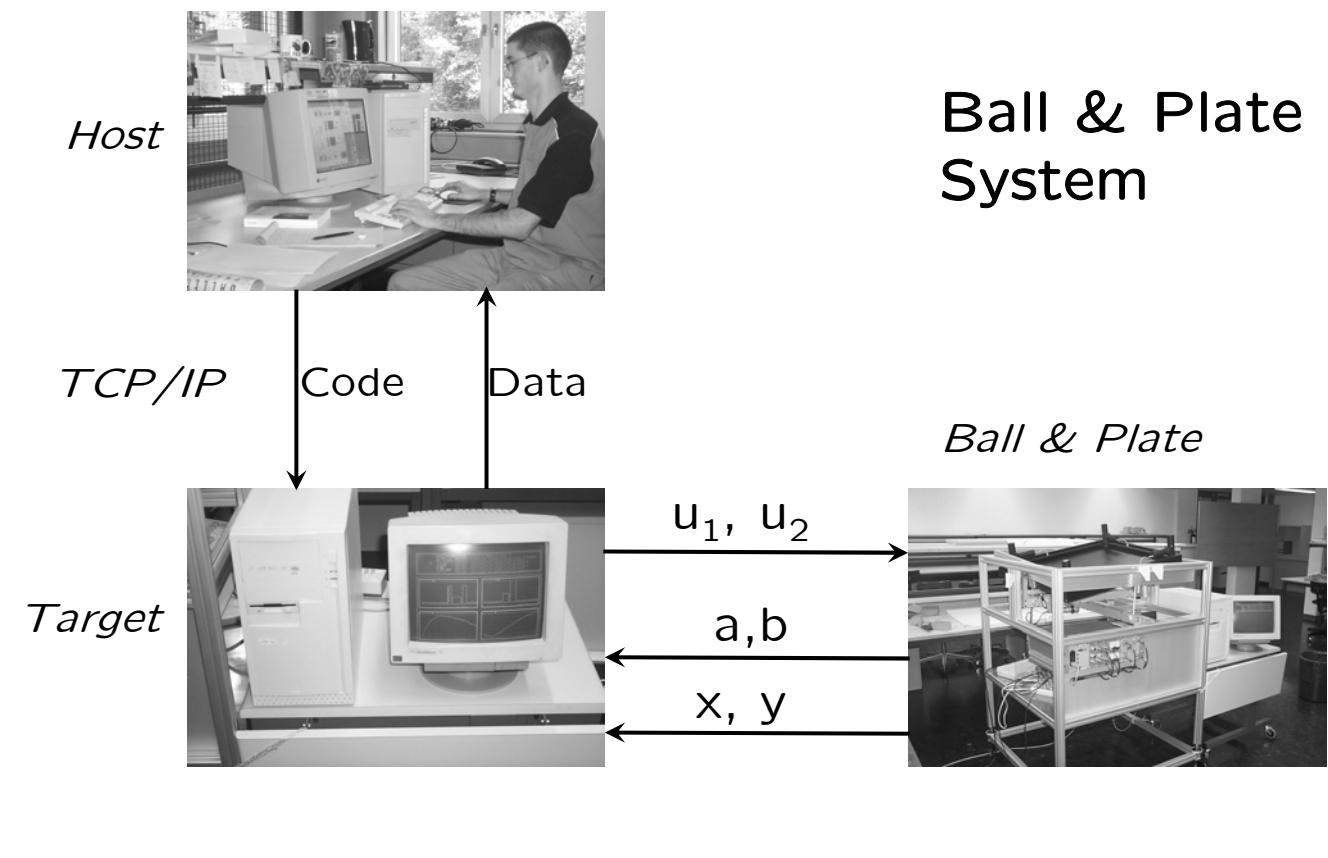
MPC Regulation of a Ball on a Plate

Task:

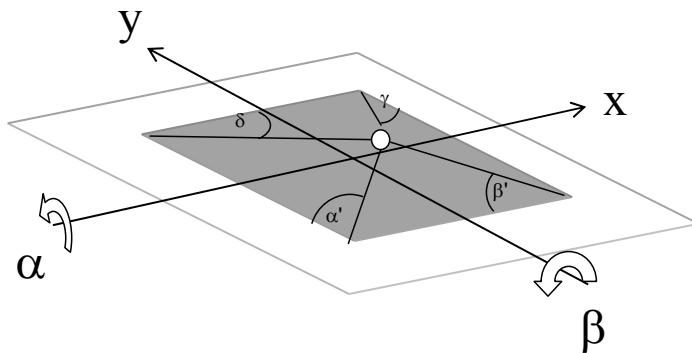
- Tune an MPC controller by simulation, using the *MPC Simulink Toolbox*
- Get the *explicit solution* of the MPC controller.
- Validate the controller on *experiments*.



B&P Experimental Setup



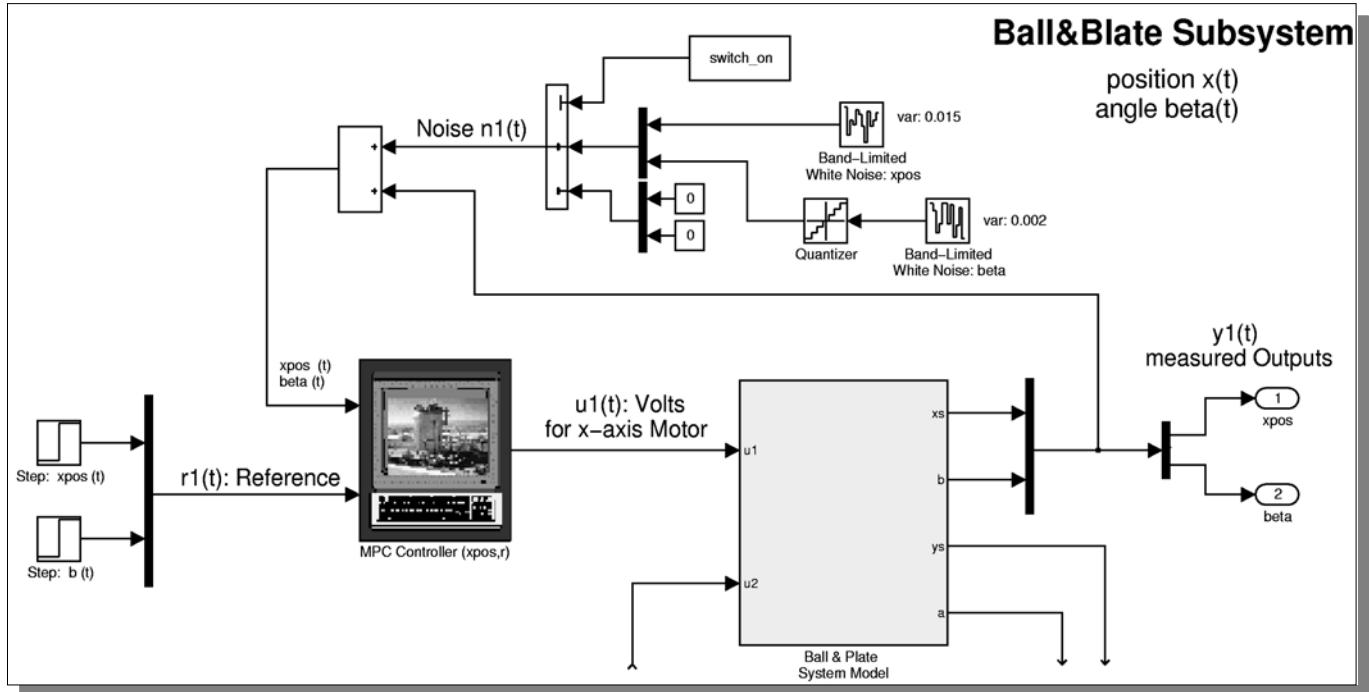
Ball&Plate Experiment



- Specifications:
 - Angle: -17 deg ... +17deg
 - Plate: -30 cm ...+30 cm
 - Input Voltage: -10 V... +10 V
 - Computer: PENTIUM166
 - Sampling Time: 30 ms
- Model: LTI 14 states
Constraints on inputs and states

General Philosophy: (1) MPC Design

- Step 1: Tune the MPC controller (in simulation)

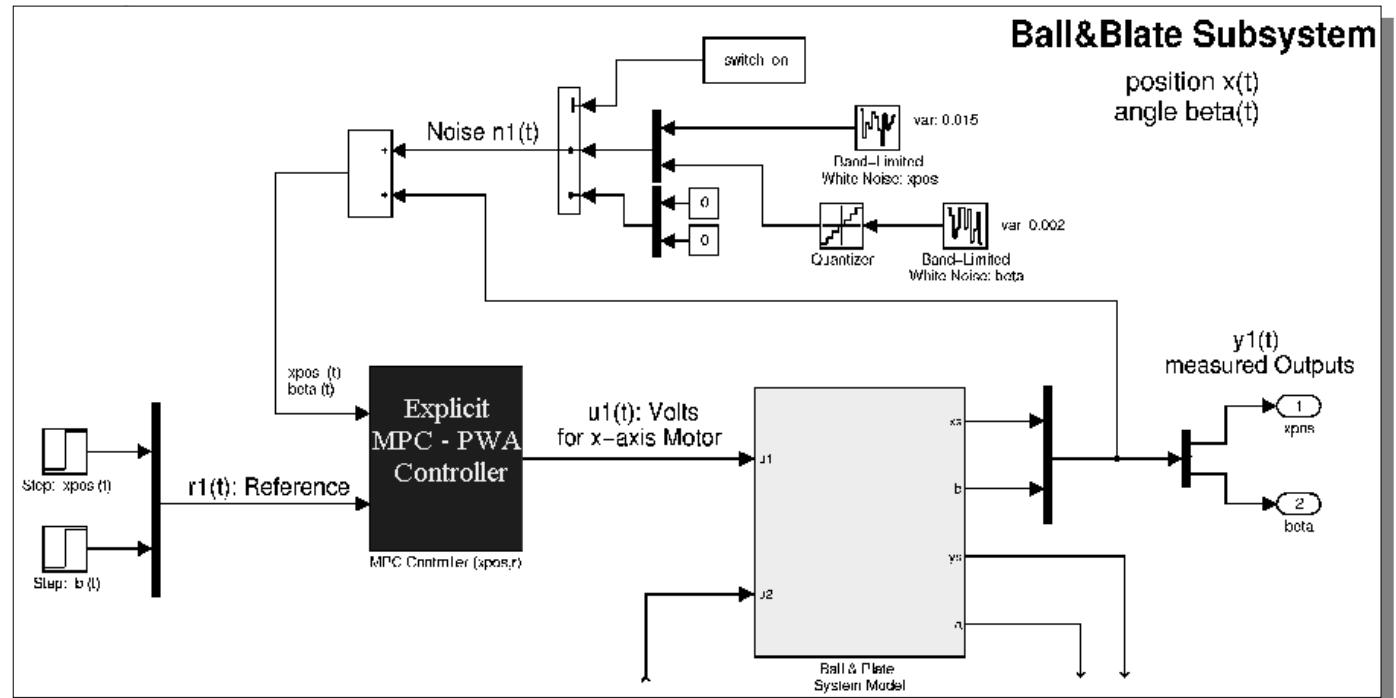


E.g: MPC Toolbox for Matlab

(Bemporad, Morari, Ricker, 2003)

General Philosophy: (2) Implementation

- Step 2: Solve mp-QP and implement Explicit

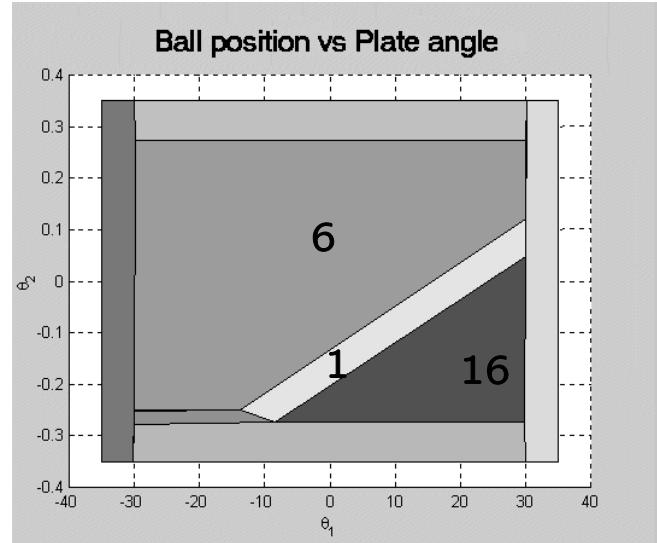
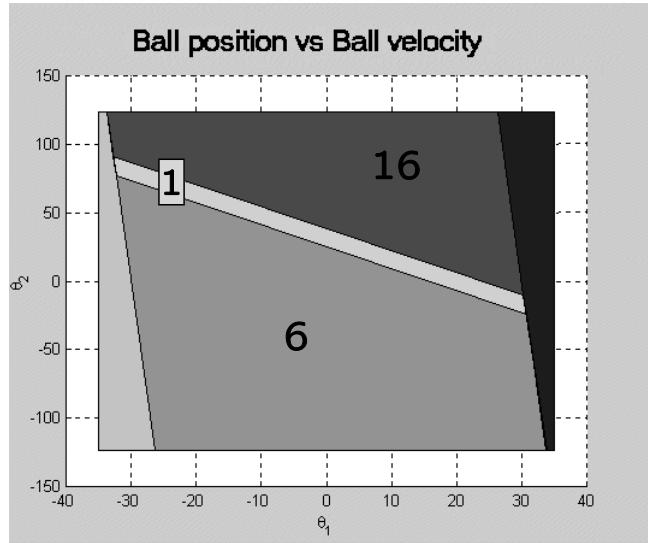


E.g: Real-Time Workshop + xPC Toolbox

Explicit MPC Solution

Controller: $x: 22$ Regions, $y: 23$ Regions

x -MPC: sections at $\alpha_x^{\circ}=0, \alpha_x=0, u_x=0, r_x=18, r_{\alpha}=0$



Region 1: LQR Controller (near Equilibrium)

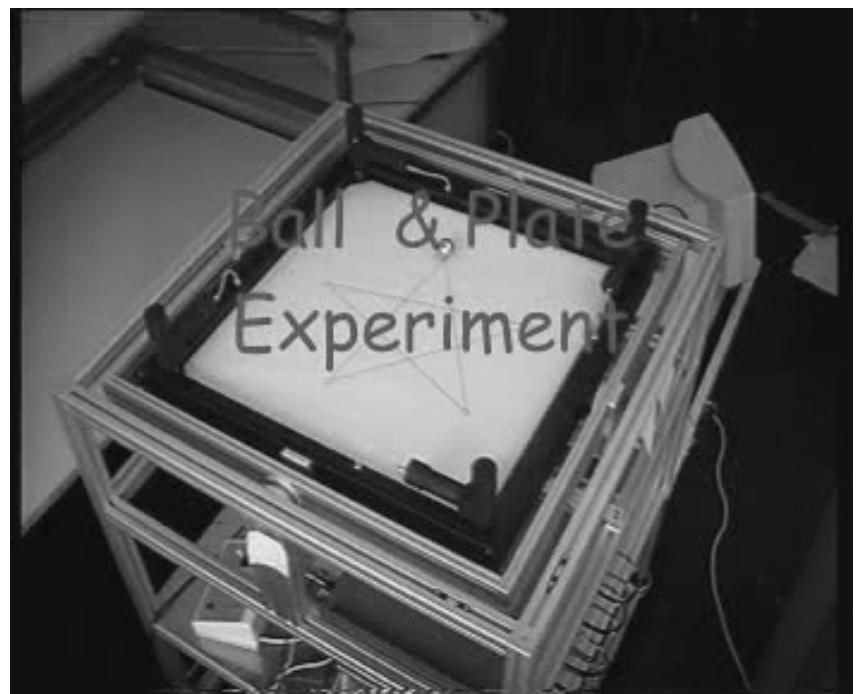
Region 6: Saturation at -10

Region 16: Saturation at +10

MPC Regulation of a Ball on a Plate

Design Steps:

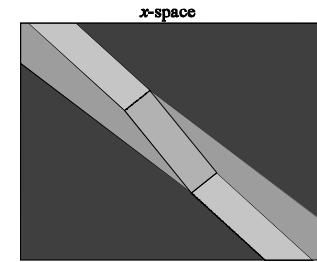
- Tune an MPC controller by simulation, using the *MPC Simulink Toolbox*
-
- Get the *explicit solution* of the MPC controller.
- ✓ Validate the controller on *experiments*.



Comments on Explicit MPC

- Multiparametric Quadratic Programs (mp-QP) can be solved efficiently
- Model Predictive Control (MPC) can be solved off-line via mp-QP
- Explicit solution of MPC controller $u = f(x)$ is Piecewise Affine

$$u(x) = \begin{cases} F_1x + G_1 & \text{if } H_1x \leq K_1 \\ \vdots & \vdots \\ F_Nx + G_N & \text{if } H_Nx \leq K_N \end{cases}$$



- Eliminate heavy on-line computation for MPC
 → Make MPC suitable for fast/small/cheap processes

MILP Formulation of MPC

(Bemporad, Borrelli, Morari, 2000)

$$\begin{aligned} & \min \sum_{k=0}^{T-1} \|Qy(t+k|t)\|_\infty + \|Ru(t+k)\|_\infty \\ & \text{s.t. MLD dynamics} \end{aligned}$$

- Introduce slack variables: $\min |x| \rightarrow$

$$\begin{aligned} & \min \epsilon \\ & \text{s.t. } \epsilon \geq x \\ & \quad \epsilon \geq -x \end{aligned}$$

$$\begin{aligned} \epsilon_k^x & \geq \|Qy(t+k|t)\|_\infty \\ \epsilon_k^u & \geq \|Ru(t+k)\|_\infty \end{aligned} \rightarrow \begin{aligned} \epsilon_k^x & \geq [Qy(t+k|t)]_i & i = 1, \dots, p, \quad k = 1, \dots, T-1 \\ \epsilon_k^x & \geq -[Qy(t+k|t)]_i & i = 1, \dots, p, \quad k = 1, \dots, T-1 \\ \epsilon_k^u & \geq [Ru(t+k)]_i & i = 1, \dots, m, \quad k = 0, \dots, T-1 \\ \epsilon_k^u & \geq -[Ru(t+k)]_i & i = 1, \dots, m, \quad k = 0, \dots, T-1 \end{aligned}$$

- Set $\xi \triangleq [\epsilon_1^x, \dots, \epsilon_{N_y}^x, \epsilon_1^u, \dots, \epsilon_{T-1}^u, U, \delta, z]$

Mixed Integer Linear Program
(MILP)

$$\begin{aligned} & \min J(\xi, x(t)) = \sum_{k=0}^{T-1} \epsilon_i^x + \epsilon_i^u \\ & \text{s.t. } G\xi \leq W + Sx(t) \end{aligned}$$

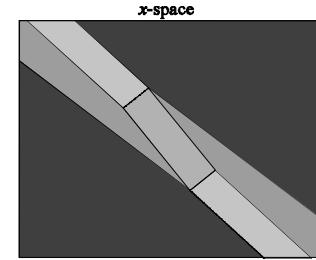
Multiparametric MILP

$$\begin{aligned} \min_{\xi=\{\xi_c, \xi_d\}} \quad & f' \xi_c + d' \xi_d \\ \text{s.t.} \quad & G \xi_c + E \xi_d \leq W + Fx \end{aligned}$$

$\xi_c \in \mathbf{R}^n$
 $\xi_d \in \{0, 1\}^m$

- mp-MILP can be solved (by alternating MILPs and mp-LPs)
 (Dua, Pistikopoulos, 1999)
- Theorem: The multiparametric solution $\xi^*(x)$ is piecewise affine
- Corollary: The MPC controller is piecewise affine in x

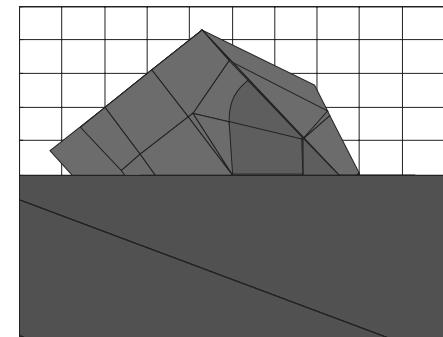
$$u(x) = \begin{cases} F_1 x + G_1 & \text{if } H_1 x \leq K_1 \\ \vdots & \vdots \\ F_N x + G_N & \text{if } H_N x \leq K_N \end{cases}$$



Solutions via Dynamic Programming

(Borrelli, Bemporad, Baotic, Morari, 2003)
 (Mayne, ECC 2001)

- Explicit solutions to finite-time optimal control problems for PWA systems can be obtained using a combination of
 - Dynamic Programming
 - Multiparametric Linear (1-norm, ∞ -norm), or or Quadratic (squared 2-norm) programming



Note: in the 2-norm case,
 the partition may not be polyhedral

Hybrid Control Examples (Revisited)

Hybrid Control - Example

Switching System:

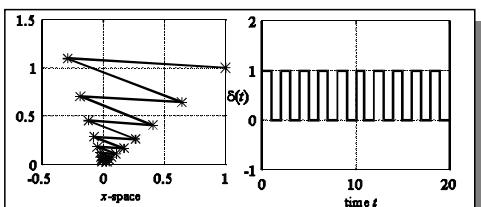
$$x(t+1) = 0.8 \begin{bmatrix} \cos \alpha(t) & -\sin \alpha(t) \\ \sin \alpha(t) & \cos \alpha(t) \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$

$$y(t) = [0 \ 1] x(t)$$

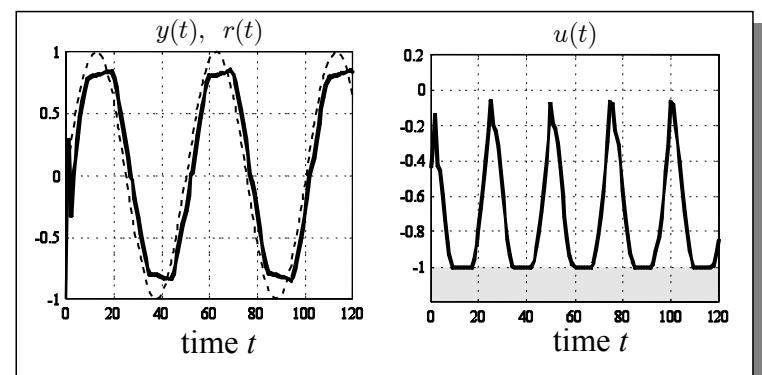
$$\alpha(t) = \begin{cases} \frac{\pi}{3} & \text{if } [1 \ 0]x(t) \geq 0 \\ -\frac{\pi}{3} & \text{if } [1 \ 0]x(t) < 0 \end{cases}$$

Constraint: $-1 \leq u(t) \leq 1$

Open loop:



Closed loop:



Hybrid MPC - Example

- MLD system

State $x(t)$	2 variables
Input $u(t)$	1 variables
Aux. binary vector $\delta(t)$	1 variables
Aux. continuous vector $z(t)$	4 variables

- mp-MILP optimization problem

$$\min_{\left\{ \begin{array}{l} v^1 \\ v^0 \end{array} \right\}} J(v_0^1, x(t)) \triangleq \sum_{k=0}^1 \|Q_1(v(k) - u_e)\|_\infty + \|Q_2(\delta(k|t) - \delta_e)\|_\infty + \|Q_3(z(k|t) - z_e)\|_\infty + \|Q_4(x(k|t) - x_e)\|_\infty$$

subject to constraints

to be solved in the region

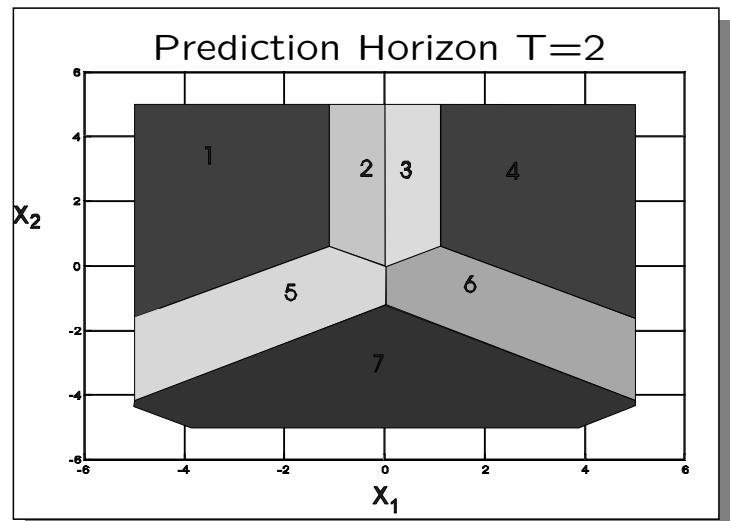
$$\begin{aligned} -5 &\leq x_1 \leq 5 \\ -5 &\leq x_2 \leq 5 \end{aligned}$$

- Computational complexity of mp-MILP

Linear constraints	84
Continuous variables	20
Binary variables	2
Parameters	2
Time to solve mp-MILP	3 min
Number of regions	7

mp-MILP Solution

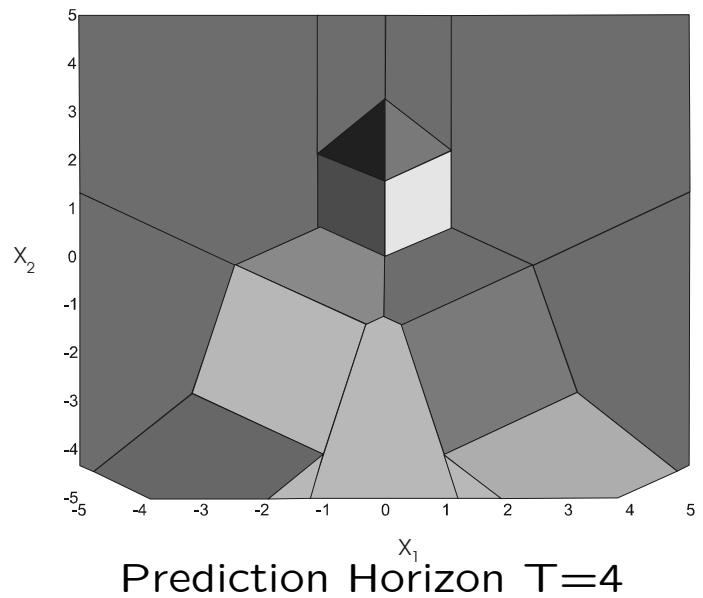
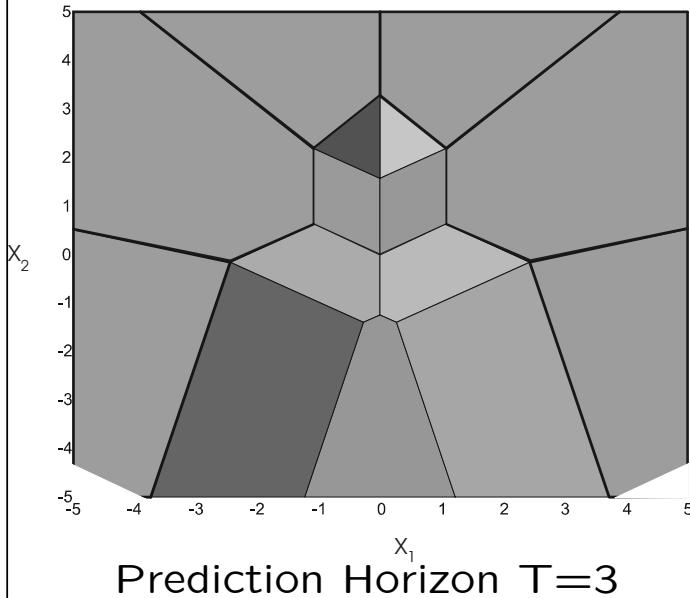
$$u = \begin{cases} -1.0000 & \text{if } \begin{bmatrix} 1.0000 & -1.7296 \\ -1.0000 & 0.0000 \\ 0.0000 & 1.0000 \\ 94.5067 & -1.0000 \end{bmatrix} x \leq \begin{bmatrix} -2.1680 \\ 5.0000 \\ 5.0000 \\ -105.1385 \end{bmatrix} \\ (\text{Region } \#1) & \\ [0.9238 \ 0.0000] x & \text{if } \begin{bmatrix} -188.1504 & 1.0000 \\ -20.3158 & -34.1997 \\ 1.0000 & 0.0000 \\ 0.0000 & 1.0000 \end{bmatrix} x \leq \begin{bmatrix} 204.3621 \\ 1.0000 \\ 0.0000 \\ 5.0000 \end{bmatrix} \\ (\text{Region } \#2) & \\ [-0.9238 \ -0.0000] x & \text{if } \begin{bmatrix} -217.00 & 1.0000 \\ 38.1919 & -64.29 \\ 1.0000 & 0.0000 \\ 0.0000 & 1.0000 \end{bmatrix} x \leq \begin{bmatrix} 4.9708 \\ 1.0000 \\ 1.0000 \\ 5.0000 \end{bmatrix} \\ (\text{Region } \#3) & \\ -1.0000 & \text{if } \begin{bmatrix} -188.15 & -1.0000 \\ -1.0000 & -1.6858 \\ 1.0000 & 0.0000 \\ 1.0000 & 133.68 \end{bmatrix} x \leq \begin{bmatrix} -204.39 \\ -2.1673 \\ 5.0000 \\ 665.60 \end{bmatrix} \\ (\text{Region } \#4) & \\ [0.4619 \ -0.8000] x & \text{if } \begin{bmatrix} 1.0000 & -1.7193 \\ -1.0000 & 1.7295 \\ -1.0000 & 0.0000 \\ 13.4029 & 23.6383 \end{bmatrix} x \leq \begin{bmatrix} 2.1387 \\ 2.1174 \\ 5.0000 \\ -1.0000 \end{bmatrix} \\ (\text{Region } \#5) & \\ [-0.4619 \ -0.8000] x & \text{if } \begin{bmatrix} 1.0000 & 1.7175 \\ -20.2527 & 34.1997 \\ -1.0000 & 0.0000 \\ -1.0000 & -1.7369 \\ 106.3247 & 1.0000 \end{bmatrix} x \leq \begin{bmatrix} -2.0413 \\ -1.0000 \\ 2.2345 \\ 525.0259 \end{bmatrix} \\ (\text{Region } \#6) & \\ 1.0000 & \text{if } \begin{bmatrix} 1.0000 & -1.6446 \\ -1.0000 & 1.7449 \\ -1.0000 & 0.0000 \\ -1.0000 & -1.6788 \\ 1.0000 & -265.137 \\ 1.0000 & 1.7369 \\ 1.0000 & 0.0000 \end{bmatrix} x \leq \begin{bmatrix} 12.0940 \\ -2.2448 \\ 5.0000 \\ 12.3141 \\ 1329.55 \\ -2.2345 \\ 5.0000 \end{bmatrix} \\ (\text{Region } \#7) & \end{cases}$$



PWA law \equiv MPC law

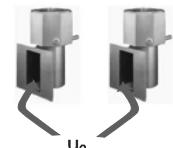
Linear constraints	84
Continuous variables	20
Binary variables	2
Parameters	2
Time to solve mp-MILP	3 min
Number of regions	7

mp-MILP Solution



Alternate Heating of Two Furnaces

(Bemporad, Borrelli, Morari, 2000)



- mp-MILP optimization problem

$$\min_{\{u(t), u(t+1), u(t+2)\}} \sum_{k=0}^2 \|R(u(t+k+1) - u(t+k))\|_\infty + \|Q(x(t+k|t) - x_e)\|_\infty$$

- Parameter set

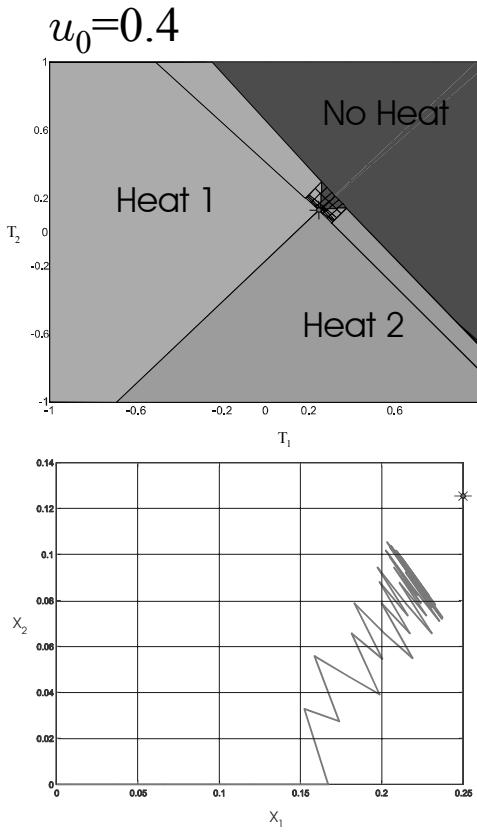
$$\begin{aligned} -1 &\leq x_1 \leq 1 \\ -1 &\leq x_2 \leq 1 \\ 0 &\leq u_0 \leq 1 \end{aligned}$$

parameterized !

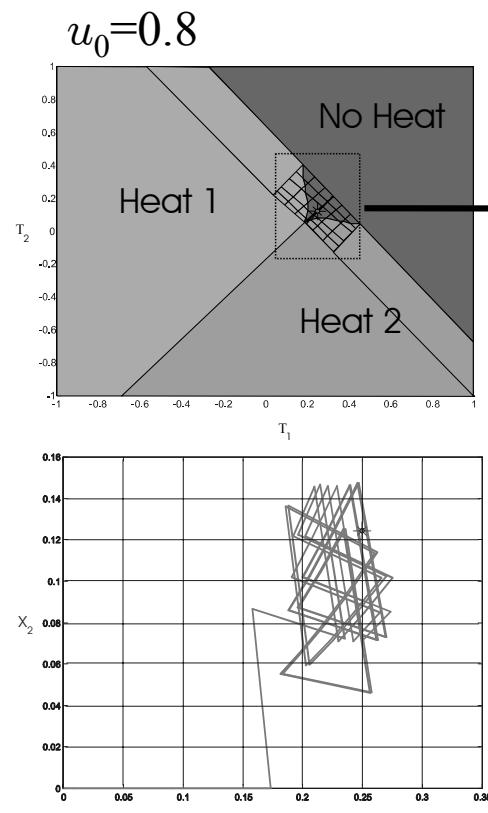
- Computational complexity of mp-MILP

Linear constraints	168
Continuous variables	33
Binary variables	6
Parameters	3
Time to solve mp-MILP	5 min
Number of regions	105

mp-MILP Solution



Set point cannot be reached



Set point is reached

NOTE: The control action of explicit and of implicit MPC are totally equal

Hybrid Control Example: Traction Control System



Vehicle Traction Control

Improve driver's ability to control a vehicle under adverse external conditions (wet or icy roads)

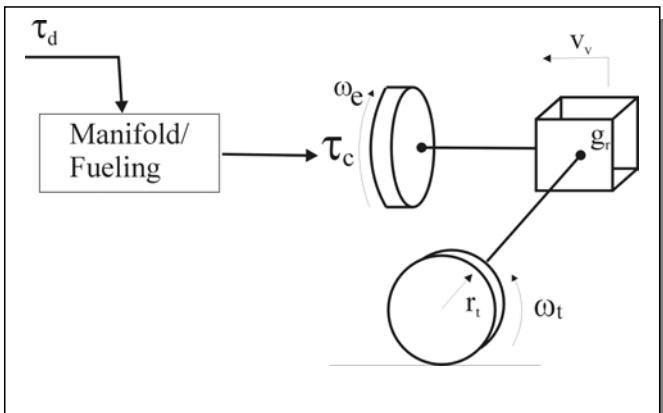


Model
nonlinear, uncertain,
constraints

Controller
suitable for real-time
implementation

MLD hybrid framework + optimization-based control strategy

Simple Traction Model



- Mechanical system

$$\dot{\omega}_e = \frac{1}{J_e} \left(\tau_c - b_e \omega_e - \frac{\tau_t}{g_r} \right)$$
$$\dot{v}_v = \frac{\tau_t}{m_v r_t}$$

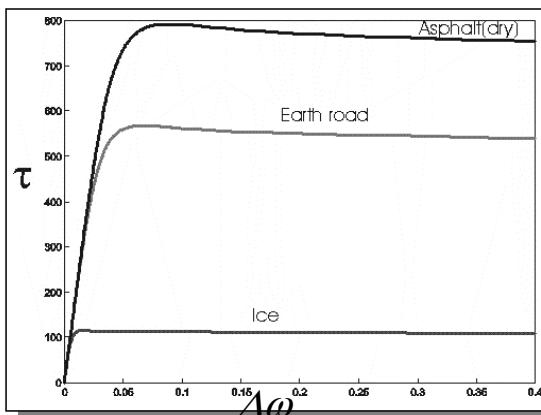
- Manifold/fueling dynamics

$$\tau_c = b_i \tau_d (t - \tau_f)$$

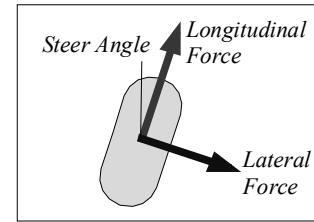
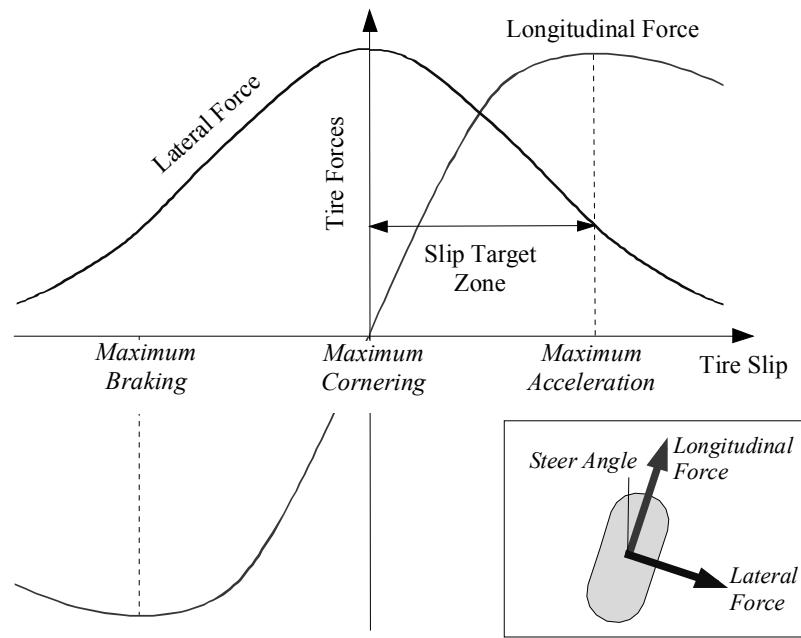
- Tire torque τ_t is a function of slip $\Delta\omega$ and road surface adhesion coefficient μ

$$\Delta\omega = \frac{\omega_e}{g_r} - \frac{v_v}{r_t}$$

Wheel slip

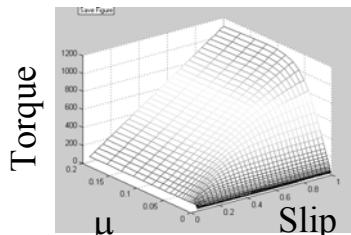


Tire Force Characteristics



Ford Motor Company

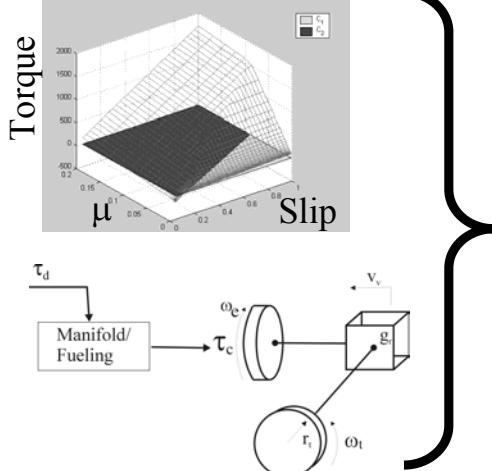
Hybrid Model



Nonlinear tire torque $\tau_t = f(\Delta\omega, \mu)$



PWA Approximation
(PWL Toolbox, Julian, 1999)



HYSDEL
(Hybrid Systems Description Language)

Mixed-Logical
Dynamical (MLD)
Hybrid Model
(discrete time)

MLD Model

$$\begin{aligned}
 x(t+1) &= Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) + B_5 \\
 y(t) &= Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t) + D_5 \\
 E_2\delta(t) + E_3z(t) &\leq E_4x(t) + E_1u(t) + E_5
 \end{aligned}$$

State $x(t)$	9 variables
Input $u(t)$	1 variable
Aux. Binary vars $\delta(t)$	3 variables
Aux. Continuous vars $z(t)$	4 variables

→ The MLD matrices are automatically generated in Matlab format by HYSDEL

Performance and Constraints

- Control objective:

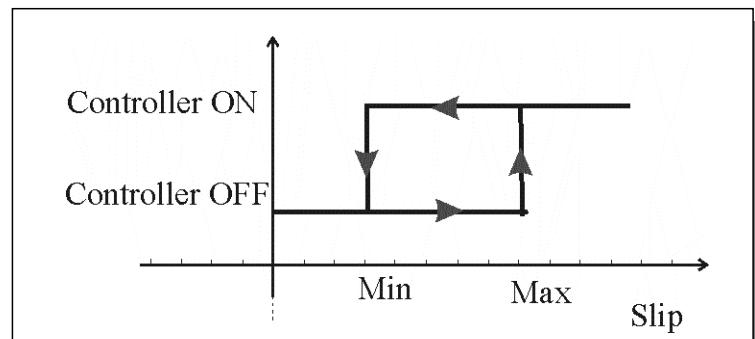
$$\begin{aligned}
 \min \sum_{k=0}^N & |\Delta\omega(k|t) - \Delta\omega_{des}| \\
 \text{subj. to.} & \quad \text{MLD Dynamics}
 \end{aligned}$$

- Constraints:

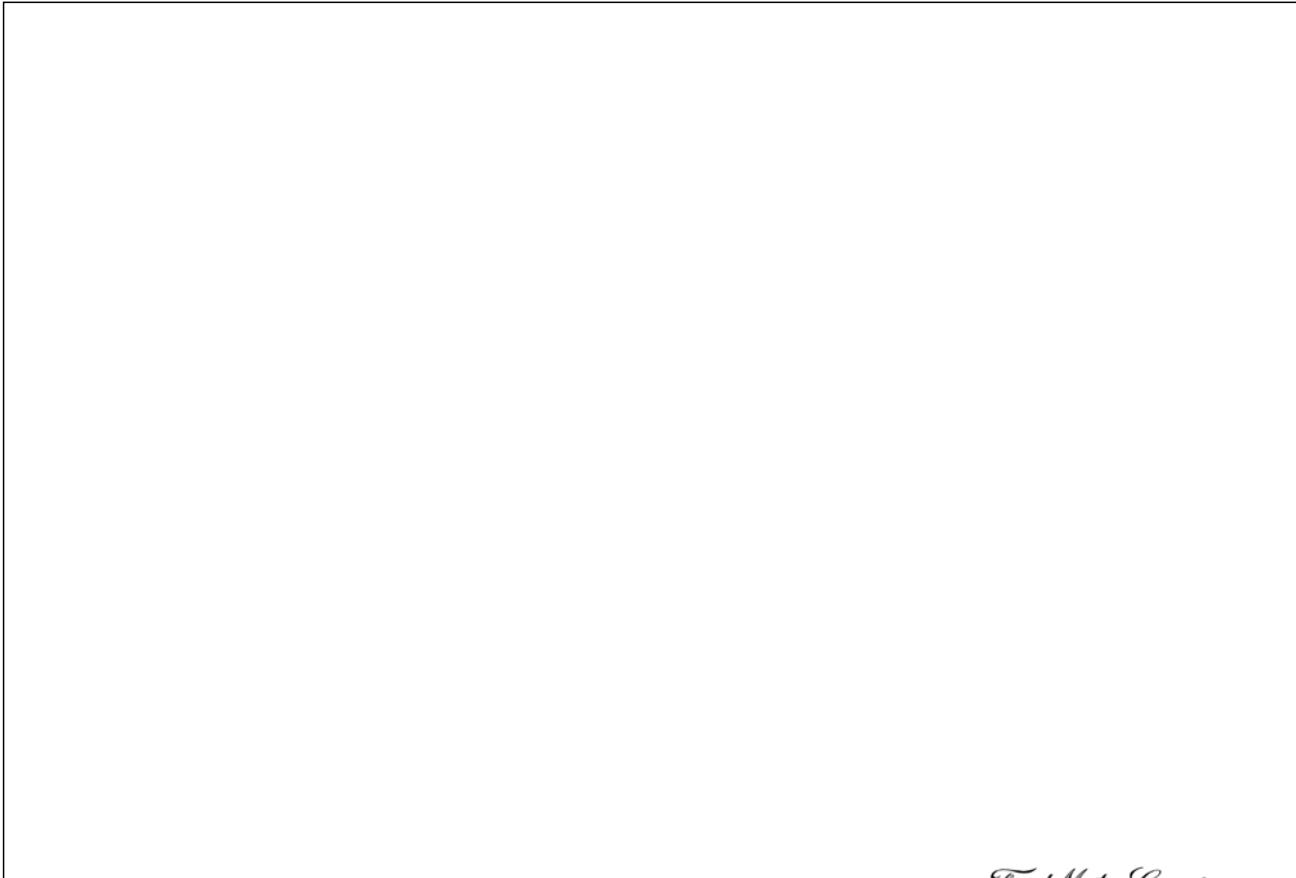
- Limits on the engine torque: $-20Nm \leq \tau_d \leq 176Nm$

- Logic Constraint:

- Hysteresis

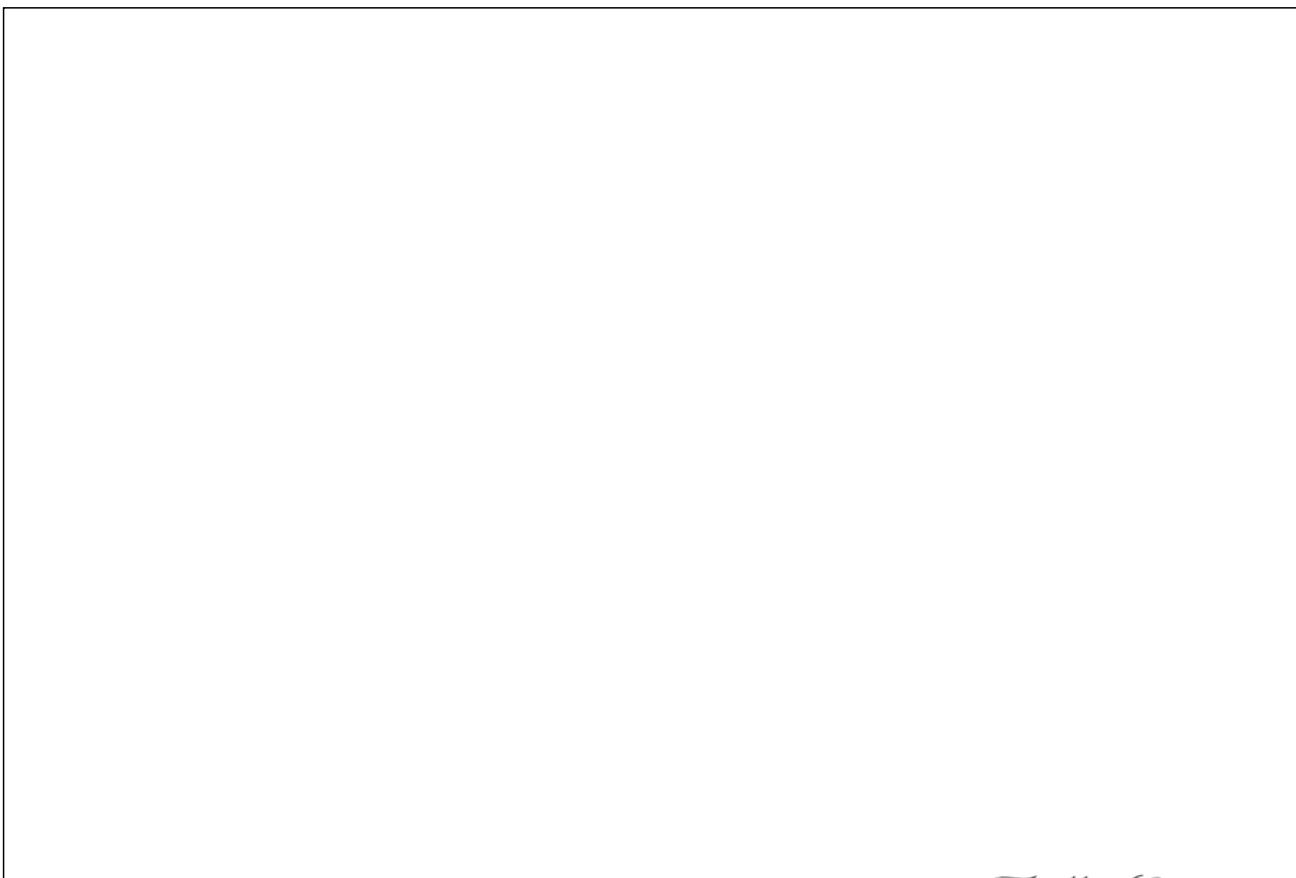


Experimental Apparatus



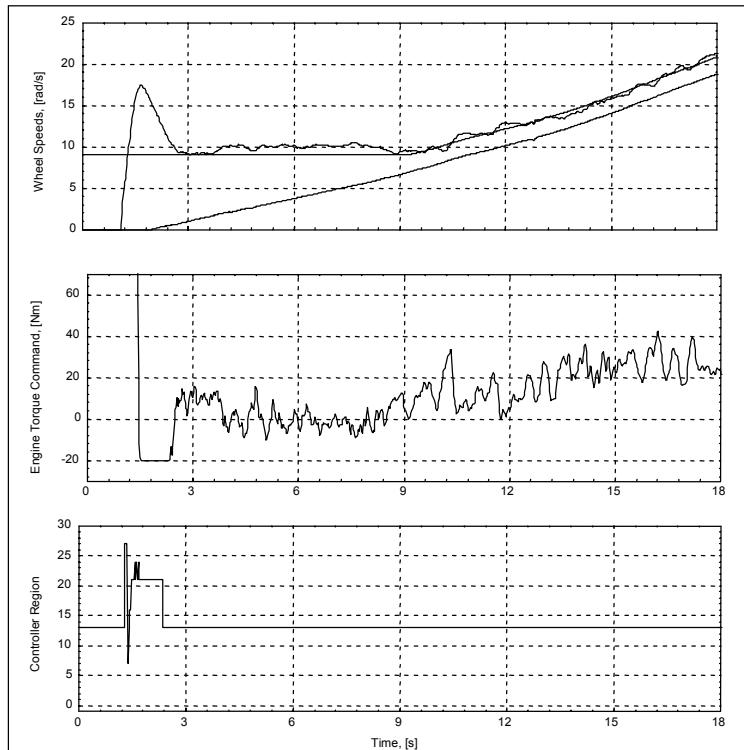
Ford Motor Company

Experimental Apparatus



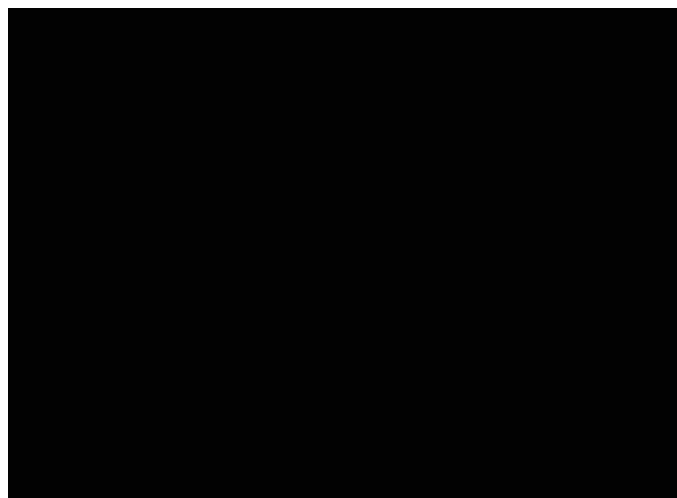
Ford Motor Company

Experimental Results



Ford Motor Company

Experiment



- >500 regions
- 20ms sampling time
- Pentium 266Mhz + Labview

Ford Motor Company

Hybrid Control Example: Cruise Control System

Hybrid Control Problem



Renault Clio 1.9 DTI RXE

GOAL:

command gear ratio, gas pedal,
and brakes to **track** a desired
speed and minimize consumption



Hybrid Model



- Vehicle dynamics

$$m\ddot{x} = F_e - F_b - \beta\dot{x}$$

\dot{x} = vehicle speed

F_e = traction force

F_b = brake force

→ discretized with sampling time $T_s = 0.5$ s

- Transmission kinematics

$$\omega = \frac{R_g(i)}{k_s}\dot{x}$$

ω = engine speed

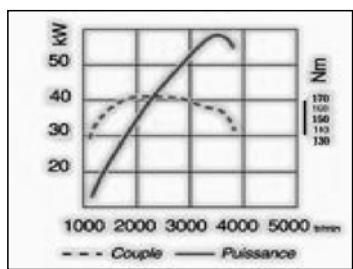
M = engine torque

i = gear

Hybrid Model



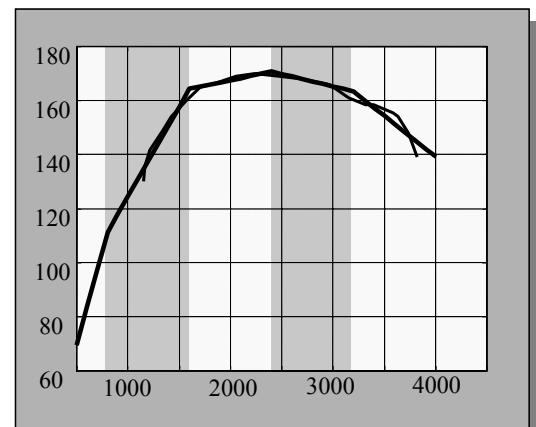
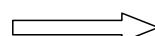
- Engine torque $-C_e^-(\omega) \leq M \leq C_e^+(\omega)$



<http://www.renault.fr>

Piecewise-linearization:
(PWL Toolbox, Julián, 1999)

requires: 4 binary aux variables
4 continuous aux variables



- Min engine torque $C_e^-(\omega) = \alpha_1 + \beta_1\omega$

Hybrid Model



- Gear selection: for each gear # i ,
define a binary input $g_i \in \{0, 1\}$

- Gear selection (traction force):

$$F_e = \frac{R_g(i)}{k_s} M \quad \text{depends on gear } \#i$$

define auxiliary continuous variables:

$$\text{IF } g_i = 1 \text{ THEN } F_{ei} = \frac{R_g(i)}{k_s} M \text{ ELSE } 0$$



$$\longrightarrow F_e = F_{eR} + F_{e1} + F_{e2} + F_{e3} + F_{e4} + F_{e5}$$

- Gear selection (engine/vehicle speed):

$$\omega = \frac{R_g(i)}{k_s} \dot{x} \quad \text{similarly, also requires 6 auxiliary continuous variables}$$

Hybrid Model



- MLD model

$$\begin{aligned} x(t+1) &= Ax(t) + B_1 u(t) + B_2 \delta(t) + B_3 z(t) \\ y(t) &= Cx(t) + D_1 u(t) + D_2 \delta(t) + D_3 z(t) \\ E_2 \delta(t) + E_3 z(t) &\leq E_4 x(t) + E_1 u(t) + E_5 \end{aligned}$$

- 2 continuous states: x, v (vehicle position and speed)
- 2 continuous inputs: M, F_b (engine torque, brake force)
- 6 binary inputs: $g_R, g_1, g_2, g_3, g_4, g_5$ (gears)
- 1 continuous output: v (vehicle speed)
- 16 auxiliary continuous vars: (6 traction force, 6 engine speed, 4 PWL max engine torque)
- 4 auxiliary binary vars: (PWL max engine torque breakpoints)
- 96 mixed-integer inequalities

Hysdel Model

```

SYSTEM car {
  INTERFACE {
    STATE { REAL position, speed; }
    INPUT { REAL torque, F_brake;
            BOOL gear1, gear2, gear3, gear4, gear5, gearR; }

    PARAMETER{
      REAL mass = 1020; /* kg */
      REAL beta_friction = 25; /* N/m*s */
      REAL Rgear1 = 3.7271; REAL Rgear2 = 2.048;
      REAL Rgear3 = 1.321; REAL Rgear4 = 0.971;
      REAL Rgear5 = 0.756; REAL RgearR = -3.545;
      REAL wheel_rim = 14; /* in */
    }
  }

  IMPLEMENTATION {
    AUX {REAL Fel, Fe2, Fe3, Fe4, Fe5, Fel0
         REAL w1, w2, w3, w4, w5, wr;
         BOOL dPWL1, dPWL2, dPWL3, dPWL4;
         REAL DCel, DCe2, DCe3, DCe4; }

    AD { dPWL1 = wPWL1-(w1+w2+w3+w4+w5+wr) <=0;
        dPWL2 = wPWL2-(w1+w2+w3+w4+w5+wr) <=0;
        dPWL3 = wPWL3-(w1+w2+w3+w4+w5+wr) <=0;
        dPWL4 = wPWL4-(w1+w2+w3+w4+w5+wr) <=0; }

    DA { Fel = {IF gear1 THEN torque/speed_factor*Rgear1;
                Fe2 = {IF gear2 THEN torque/speed_factor*Rgear2;
                Fe3 = {IF gear3 THEN torque/speed_factor*Rgear3;
                Fe4 = {IF gear4 THEN torque/speed_factor*Rgear4;
                Fe5 = {IF gear5 THEN torque/speed_factor*Rgear5;
                Fel = {IF gearR THEN torque/speed_factor*RgearR;

    w1 = {IF gear1 THEN speed/speed_factor*Rgear1;
    w2 = {IF gear2 THEN speed/speed_factor*Rgear2;
    w3 = {IF gear3 THEN speed/speed_factor*Rgear3;
    w4 = {IF gear4 THEN speed/speed_factor*Rgear4;
    w5 = {IF gear5 THEN speed/speed_factor*Rgear5;
    wr = {IF gearR THEN speed/speed_factor*RgearR;
  }
}

CONTINUOUS { position = position+Ts*speed;
             speed = speed+Ts/mass*(Fel+Fe2+Fe3+Fe4+Fe5+Fbrake-
             F_brake-beta_friction*speed);
             wmin := w1+w2+w3+w4+w5+wr;
             wlim2:=w1+w2+w3+w4+w5+wr; /* wmax; */
             -F_brake <=0; /* brakes cannot accelerate ! */
             F_brake <= max_brake_force;
             -torque-(alpha1+beta1*(w1+w2+w3+w4+w5+wr)) <=0;
             torque-(aPWL1+bPWL1*(w1+w2+w3+w4+w5+wr)+DCel+DCe2+DCe3+DCe4)-1<=0;
             -(gear1+gear2+gear3+gear4+gear5+gearR) <=1;
             (gear1+gear2+gear3+gear4+gear5+gearR)<=1;
             Fel+Fe2+Fe3+Fe4+Fe5+Fbrake <= max_force;
             -Fel-Fe2-Fe3-Fe4-Fe5-Fbrake <= -max_force;
             dPWL4 => dPWL3; dPWL4 => dPWL2;
             dPWL4 => dPWL1; dPWL3 => dPWL2;
             dPWL3 => dPWL1; dPWL2 => dPWL1; }
}

```



<http://control.ethz.ch/~hybrid/hysdel>

Hybrid Controller

- Max-speed controller

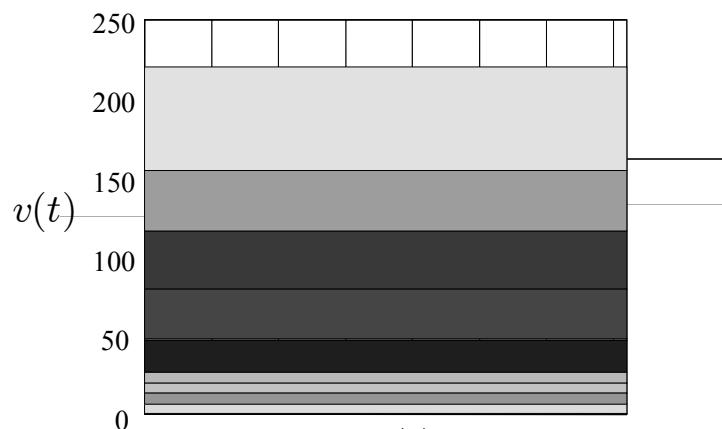
$$\max_{u_t} J(u_t, x(t)) \triangleq v(t+1|t)$$

subj. to $\begin{cases} \text{MLD model} \\ x(t|t) = x(t) \end{cases}$



MILP optimization problem

Linear constraints	96
Continuous variables	18
Binary variables	10
Parameters	1
Time to solve mp-MILP (Sun Ultra 10)	45 s
Number of regions	11

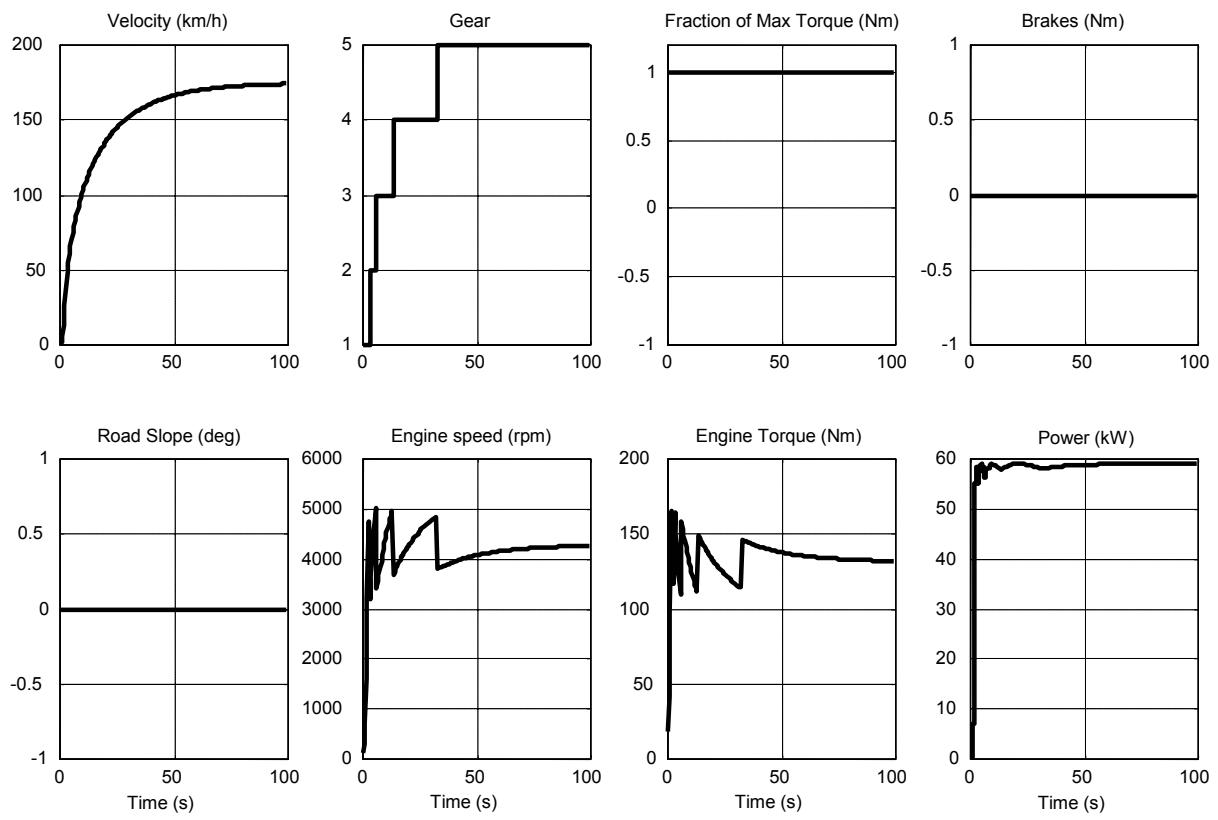


$(x(t) \text{ is irrelevant})$

Hybrid Controller



- Max-speed controller



Hybrid Controller



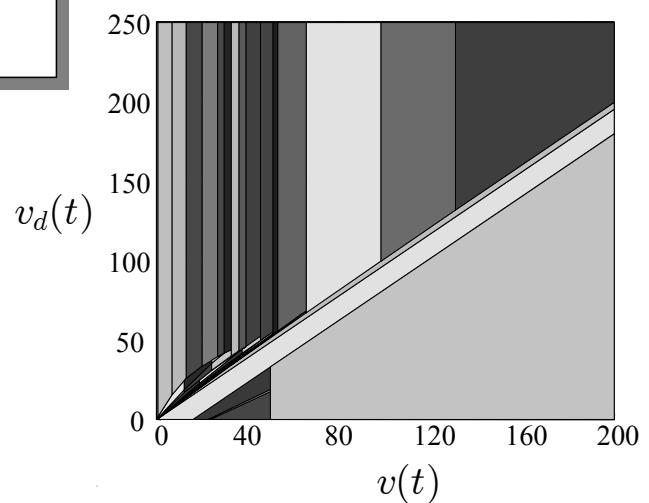
- Tracking controller

$$\min_{u_t} J(u_t, x(t)) \triangleq |v(t+1|t) - v_d(t)| + \rho |\omega|$$

subj. to $\begin{cases} \text{MLD model} \\ x(t|t) = x(t) \end{cases}$

MILP optimization problem

Linear constraints	98
Continuous variables	19
Binary variables	10
Parameters	2
Time to solve mp-MILP (Sun Ultra 10)	27 m
Number of regions	49

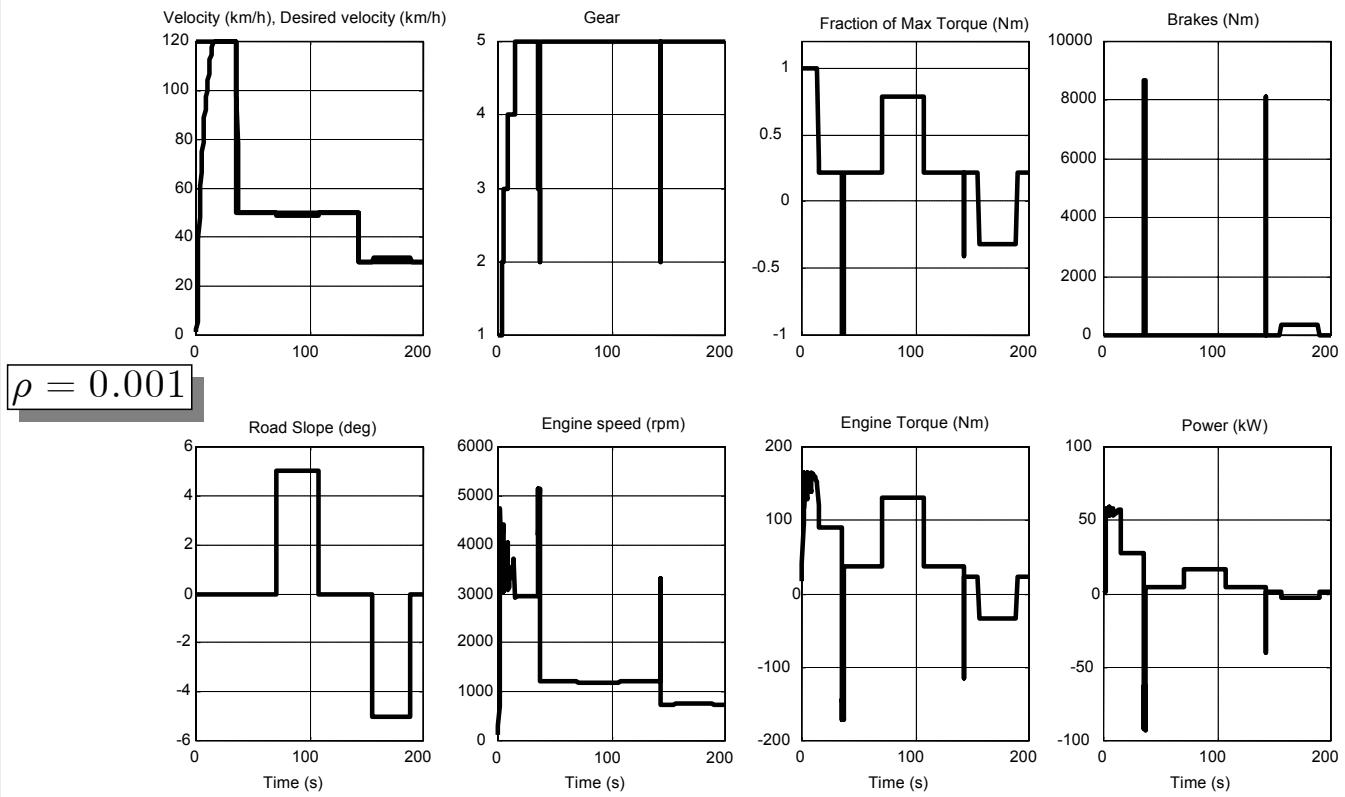


Hybrid Controller



- Tracking controller

$$\min_{u_t} J(u_t, x(t)) \triangleq |v(t+1|t) - v_d(t)| + \rho |\omega|$$



Hybrid Controller



- Smoother tracking controller

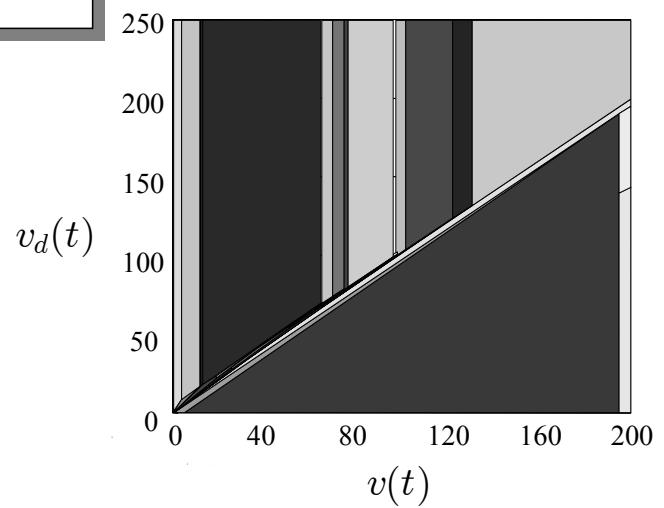
$$\min_{u_t} J(u_t, x(t)) \triangleq |v(t+1|t) - v_d(t)| + \rho |\omega|$$

subj. to

$$\begin{cases} |v(t+1|t) - v(t)| < T_s a_{\max} \\ \text{MLD model} \\ x(t|t) = x(t) \end{cases}$$

MILP optimization problem

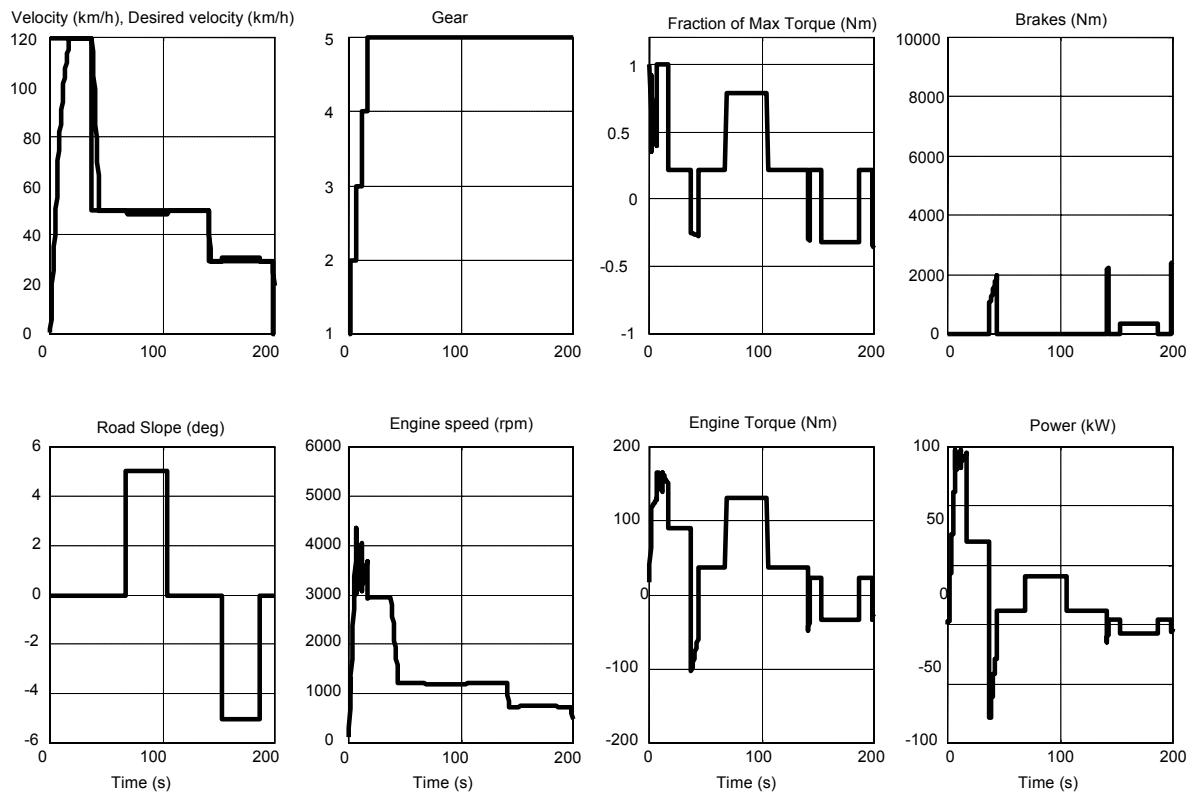
Linear constraints	100
Continuous variables	19
Binary variables	10
Parameters	2
Time to solve mp-MILP (Sun Ultra 10)	28 m
Number of regions	54



Hybrid Controller



- Smoother tracking controller



Reachability Analysis (Verification)

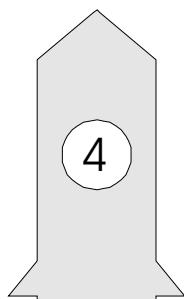
Verification

- **GIVEN:** an embedded system (continuous dynamical system + logic controller)
- **CERTIFY** that such combination behaves as desired
 - for ALL initial conditions within a given set
 - for ALL disturbances within a given class
- or **PROVIDE** a counterexample.

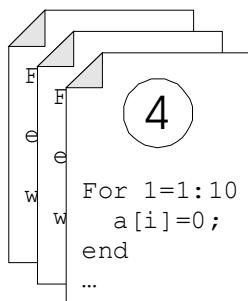
Simulation: provides a partial answer (not all possibilities can be tested!)

Reachability Analysis: provides the answer.

Ariane 5 - 1996



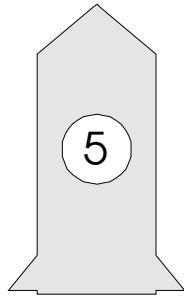
+



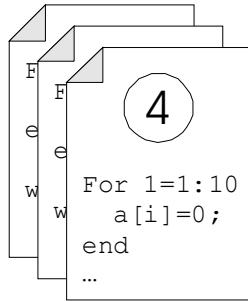
• Success

Physical behaviour

Computer code



+

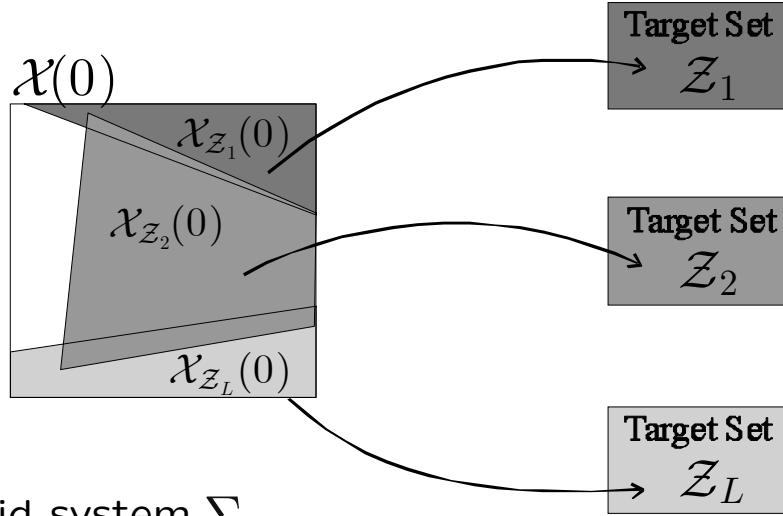


• Failure

Need for verification

Reachability Analysis/Verification

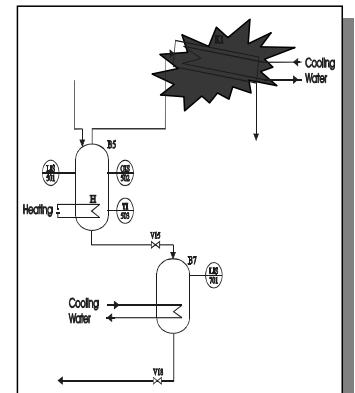
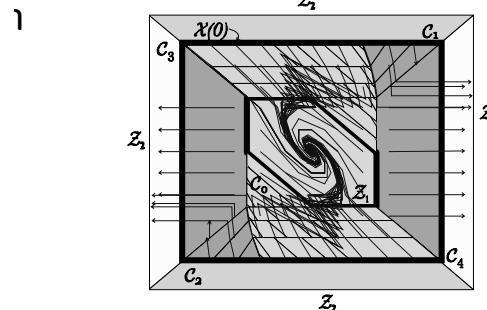
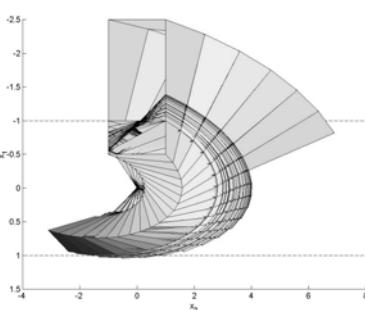
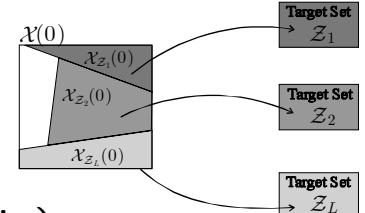
(Bemporad, Torrisi, Morari, 2000)



- Given:
 - A hybrid system Σ
 - A set of initial conditions $\mathcal{X}(0)$
 - Target sets $\mathcal{Z}_1, \mathcal{Z}_2, \dots, \mathcal{Z}_L$ (disjoint)
 - Time horizon $t \leq T_{\max}$
- Problem:
 - Is \mathcal{Z}_i reachable from $\mathcal{X}(0)$ in t steps ?
 - If yes, from which subset $\mathcal{X}_{\mathcal{Z}_i}(0)$ of $\mathcal{X}(0)$?
 - Disturbance/input sequences driving $\mathcal{X}_{\mathcal{Z}_i}(0)$ to \mathcal{Z}_i .

Applications

- Safety ($\mathcal{Z}_1, \mathcal{Z}_2, \dots, \mathcal{Z}_L = \text{unsafe sets}$)
- Stability ($\mathcal{Z}_1 = \text{invariant set around the origin}$)
- Optimal control ($u(0), \dots, u(T_{\max}) = \text{optimal strategy}$,
 $\mathcal{Z}_1 = \text{reference set}$) (Bemporad, Giovanardi, Torrisi, CDC 2000)
- (practical) Liveness ($\mathcal{Z}_1 = \text{set to be reached within a finite time}$)



Verification Algorithm via MILP

- Simple solution: Solve $\forall T > 0$ the mixed-integer linear feasibility test

$$\left\{ \begin{array}{l} x(0) \in \mathcal{X}(0) \\ x(T) \in \mathcal{Z}_i \\ u(t) \in \mathcal{U}, 0 \leq t \leq T \\ x(t+1) = Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) \\ E_2\delta(t) + E_3z(t) \leq E_1u(t) + E_4x(t) + E_5 \end{array} \right.$$

with respect to $x(0), \{u(t), \delta(t), z(t)\}_{t=0}^T$

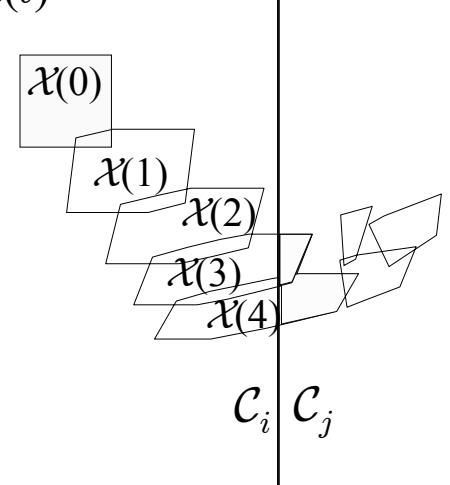
- Only practical for small problems ! because number of free integer variables $\delta(0), \delta(1), \dots, \delta(T)$ grows with T
- Efficient Solution: Exploit the special structure of the problem.
 (Bemporad, Torrisi, Morari, 2000)
 (Torrisi, Bemporad, Giovanardi, 2003)

Reach-Set Evolution

$T_{\max} < \infty$, discrete-time \Rightarrow Decidable but \mathcal{NP} -hard

Reachability analysis algorithm:

- Compute the polyhedral reach set $\mathcal{X}(t)$
- Detect switching
- Describe new intersections $\mathcal{X}(t) \cap \mathcal{C}_j$
- Stopping criteria for a single exploration
- Organize the search



Reach Set Computation

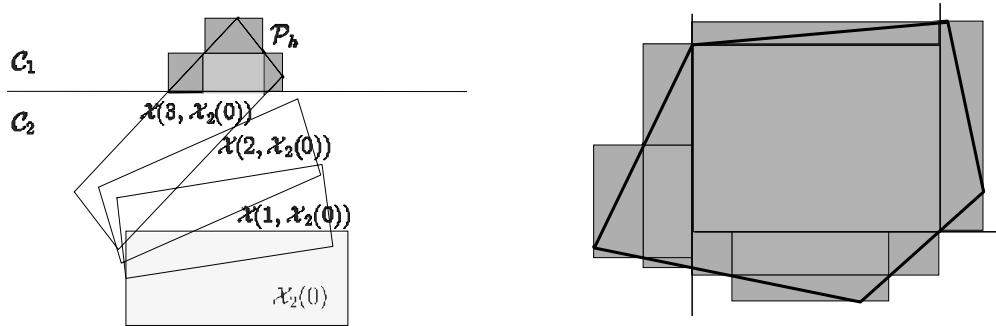
Reach set implicitly defined by linear inequalities

$$\left\{ \begin{array}{l} x \in \mathcal{X}(0) \\ K_i^x \left(A_i^k x + \sum_{k=0}^{t-1} A_i^j [B_i u(t-1-k) + f_i] \right) + \\ + K_i^u u(t) \leq H, \quad k = 1, \dots, t \\ u_{\min} \leq u(k) \leq u_{\max}, \quad k = 0, \dots, t-1 \end{array} \right.$$

where $\mathcal{X}_i = \{(x, u) : K_i^x x + K_i^u u \leq H\}$ is the current region

- Simple to compute
- Number of constraints grows linearly with time
- Explicit form also possible via projection methods (e.g. CDD, Fukuda 1997)

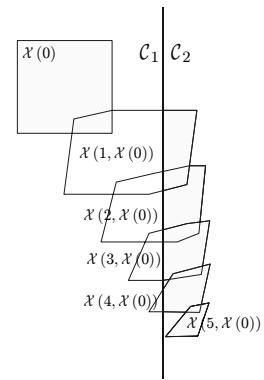
Approximation of Intersections



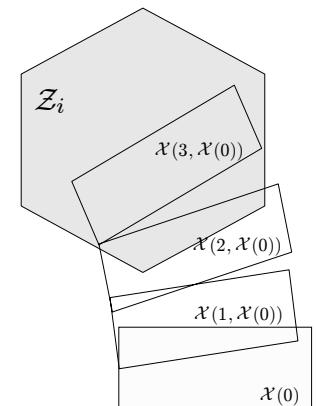
- ✓ Simple to compute via Linear Programming (LP)
- ✓ Can approximate with arbitrary precision
- ✓ Trade off between conservativeness and complexity
- ✓ Both inner and outer approximations in one shot
- ✓ Approximate computation of projections

Stopping Criteria

- Reach set has left the current region

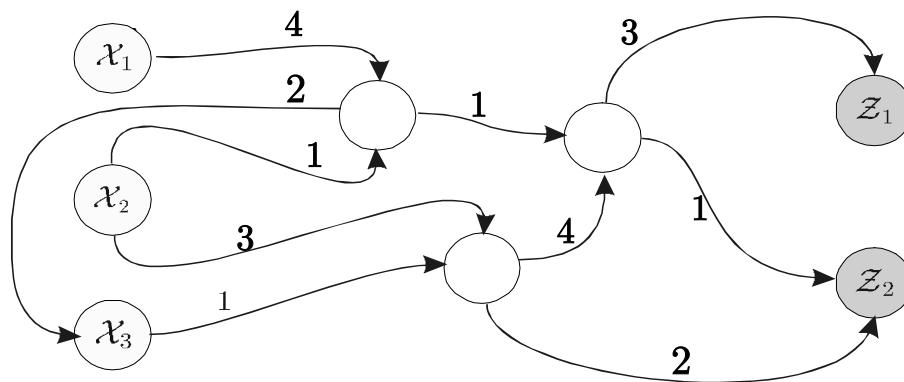


- Reach set is all inside a target \mathcal{Z}_i



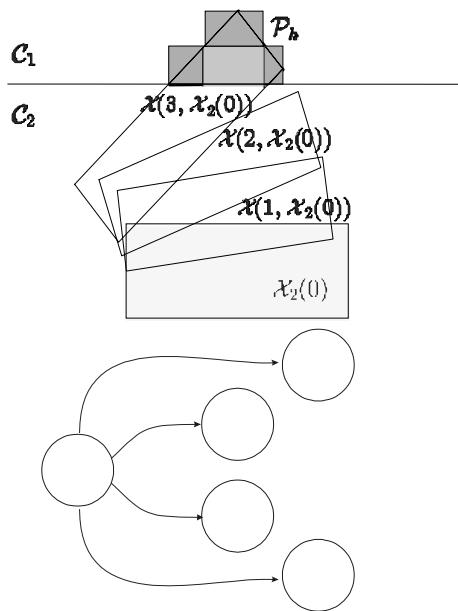
- $t > T_{\max}$ (to guarantee termination)

Graph of Evolution

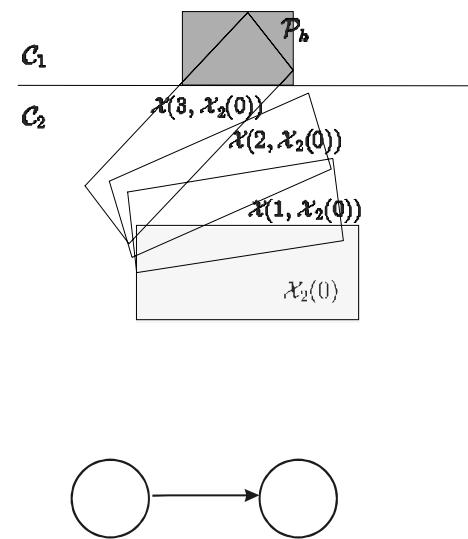


- The graph is initialized with all the initial subsets $\mathcal{X}_i = \mathcal{X}(0) \cap \mathcal{C}_i$ and all the target sets
- A node is added for each initial region of a new exploration
- If a set can reach another set, an oriented arc is drawn
- The time needed to reach the set is a weight associated with the arc

Conservativeness vs Graph Complexity



- Less conservative
- Graph more complicate

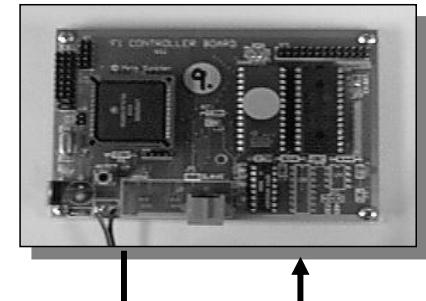


- Graph less complicate
- More conservative

LP determines if a switching sequence is feasible

Verification Example: Cruise Control System

Cruise Control System



GOAL:

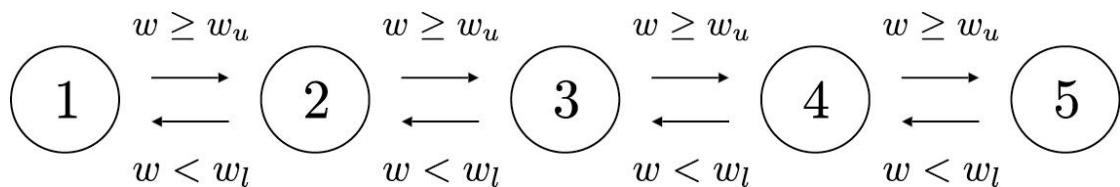
Verify if a given switching controller satisfies certain specifications

(Torrisi, Bemporad, 2001)

Cruise Control System



Gear selector:

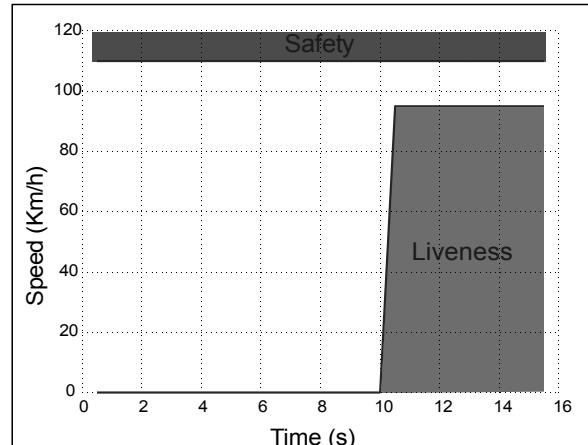


Speed controller:

$$\begin{aligned} e(t+1) &= e(t) + T_s(v_r(t) - v(t)) \quad (+ \text{ saturation}) \\ u_e(t) &= \begin{cases} k_e(v_e(t) - v(t)) + i_{ee}(t) & \text{if } v(t) < v_r + 1 \\ 0 & \text{otherwise} \end{cases} \\ u_b(t) &= \begin{cases} k_b(v_e(t) - v(t)) & \text{if } v(t) \geq v_r + 1 \\ 0 & \text{otherwise} \end{cases} \end{aligned}$$

Verification Problem

Question: Will the cruise control reach the desired speed reference within 10 s without exceeding the speed limit?



Safety

$$\mathcal{Z}_1 = \{v : v > v_r + r_{\text{toll}}\}$$

Liveness

$$\mathcal{Z}_2 = \{v, t : v < v_r - 2r_{\text{toll}}, t > 10/T_s\}$$

$$r_{\text{toll}} = 5 \text{ km/h}$$

Hysdel Model (HYbrid Systems DEscription Language)

```

SYSTEM car {
INTERFACE {
    STATE { REAL speed, err, vr; BOOL gear1, gear2, gear3, gear4, gear5; }
    PARAMETER {...}
IMPLEMENTATION {
    AUX {
        REAL Fe1, Fe2, Fe3, Fe4, Fe5, w1, w2, w3, w4, w5, DCe1, DCe2, DCe3, DCe4, zut, zub, ierr, torque, F_brake;
        BOOL dPWLL1, dPWLL2, dPWLL3, dPWLL4, sd, su, verr, sat_torque, sat_F_brake, no_sat;
    LOGIC { no_sat = ~(sat_torque | sat_F_brake) & verr; }
    AD {dPWLL1 = wPWLL1 - (w1 + w2 + w3 + w4 + w5) <= 0; dPWLL2 = wPWLL2 - (w1 + w2 + w3 + w4 + w5) <= 0;
        dPWLL3 = wPWLL3 - (w1 + w2 + w3 + w4 + w5) <= 0; dPWLL4 = wPWLL4 - (w1 + w2 + w3 + w4 + w5) <= 0;
        sd = (w1 + w2 + w3 + w4 + w5) - wl <= 0; su = wu - (w1 + w2 + w3 + w4 + w5) <= 0; verr = speed - vr - 2 <= 0;
        sat_torque = - zut + (DCe1 + DCe2 + DCe3 + DCe4) + 1 <= 0; sat_F_brake = - zub + max_brake_force <= 0; }
    DA {Fe1 = {IF gear1 THEN torque / speed_factor * Rgear1}; Fe2 = {IF gear2 THEN torque / speed_factor * Rgear2};
        Fe3 = {IF gear3 THEN torque / speed_factor * Rgear3}; Fe4 = {IF gear4 THEN torque / speed_factor * Rgear4};
        Fe5 = {IF gear5 THEN torque / speed_factor * Rgear5};

        w1 = {IF gear1 THEN speed / speed_factor * Rgear1}; w2 = {IF gear2 THEN speed / speed_factor * Rgear2};
        w3 = {IF gear3 THEN speed / speed_factor * Rgear3}; w4 = {IF gear4 THEN speed / speed_factor * Rgear4};
        w5 = {IF gear5 THEN speed / speed_factor * Rgear5};

        DCe1 = {IF dPWLL1 THEN (aPWLL2) + (bPWLL2) * (w1 + w2 + w3 + w4 + w5) ELSE (aPWLL1) + (bPWLL1) * (w1 + w2 + w3 + w4 + w5)};
        DCe2 = {IF dPWLL2 THEN (aPWLL3 - aPWLL2) + (bPWLL3 - bPWLL2) * (w1 + w2 + w3 + w4 + w5)};
        DCe3 = {IF dPWLL3 THEN (aPWLL4 - aPWLL3) + (bPWLL4 - bPWLL3) * (w1 + w2 + w3 + w4 + w5)};
        DCe4 = {IF dPWLL4 THEN (aPWLL5 - aPWLL4) + (bPWLL5 - bPWLL4) * (w1 + w2 + w3 + w4 + w5)};

        zut = {IF verr THEN kt * (vr - speed) + it * err}; zub = {IF ~verr THEN - kb * (vr - speed) - ib * err};
        torque = {IF sat_torque THEN (DCe1 + DCe2 + DCe3 + DCe4) + 1 ELSE zut}; F_brake = {IF sat_F_brake THEN max_brake_force ELSE zub};
        ierr = {IF no_sat THEN err + Ts * (vr - speed)}; }

    CONTINUOUS {
        speed = speed + Ts / mass * (Fe1 + Fe2 + Fe3 + Fe4 + Fe5 - F_brake - beta_friction * speed); err = ierr; vr = vr; }

    AUTOMATA {
        gear1 = (gear2 & sd) | (gear1 & ~su); gear2 = (gear1 & su) | (gear2 & sd) | (gear2 & ~sd & ~su);
        gear3 = (gear2 & su) | (gear4 & sd) | (gear3 & ~sd & ~su); gear4 = (gear3 & su) | (gear5 & sd) | (gear4 & ~sd & ~su);
        gear5 = (gear4 & su) | (gear5 & ~sd ); }

    MUST {
        -wl <= -wemin; wl <= wemax; -w2 <= -wemin; w2 <= wemax; -w3 <= -wemin; w3 <= wemax; -w4 <= -wemin; w4 <= wemax; -w5 <= -wemin;
        w5 <= wemax; -F_brake <= 0; F_brake <= max_brake_force; torque - (DCe1 + DCe2 + DCe3 + DCe4) - 1 <= 0;
        -(REAL gear1) + (REAL gear2) + (REAL gear3) + (REAL gear4) + (REAL gear5) <= -0.9999;
        (REAL gear1) + (REAL gear2) + (REAL gear3) + (REAL gear4) + (REAL gear5) <= 1.0001;
        dPWLL4 -> dPWLL3; dPWLL4 -> dPWLL2; dPWLL4 -> dPWLL1; dPWLL3 -> dPWLL2; dPWLL3 -> dPWLL1; dPWLL2 -> dPWLL1;}}
}

```



Hybrid Model



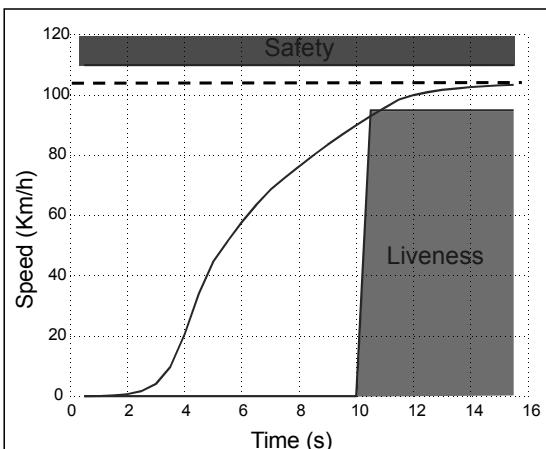
- MLD model

$$\begin{aligned}
 x'(k) &= Ax(k) + B_1u(k) + B_2\delta(k) + B_3z(k) + B_5 \\
 y(k) &= Cx(k) + D_1u(k) + D_2\delta(k) + D_3z(k) + D_5 \\
 E_2\delta(k) + E_3z(k) &\leq E_1u(k) + E_4x(k) + E_5
 \end{aligned}$$

- 3 continuous states: v, v_r, e (speed, reference and tracking error)
- 5 binary states: g_1, g_2, g_3, g_4, g_5 (gears)
- 19 auxiliary continuous vars: (5 traction force, 5 engine speed, 5 reset/saturation, 4 PWL max engine torque)
- 15 auxiliary binary vars: (4 PWL max torque breakpoints, 4 saturations, 5 logic updates, 2 gear switching conditions)
- 173 mixed-integer inequalities

Verification Results

- For all $v_r \in [30, 70] \text{ km/h}$ the controller satisfies both liveness & safety properties
- CPU time: $\sim 2.5\text{h}$ (Matlab 5.3, PC650MHz)



- For $v_r \in [30, 120] \text{ km/h}$ the verification algorithm finds the first counterexample after $\sim 7\text{m}$

Conclusions

- Hybrid systems as a framework for new applications, where both logic and continuous dynamics are relevant
- Mixed Logical Dynamical (MLD) systems as discrete-time, *computation-oriented* models for hybrid systems

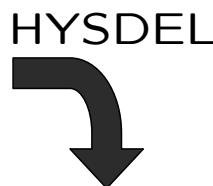
```
/* HEAT EXCHANGER model - (C) 2001 by A. Bemporad, Zurich, March 13, 2001 */
SYSTEM furnaces {
INTERFACE {
    STATE {
        REAL t1,t2,u0; }
    INPUT {
        BOOL heat1,heat2; }
    PARAMETER {
        REAL Ts=0.08; /* sampling time, seconds */
        REAL Bd1=0.07688365361396; /* discretization of B matrix, heat 1 active */
        REAL Bd2=0.07932810551699; /* discretization of B matrix, heat 2 active */
        REAL A11=-9231; /* discretization of A matrix */
        REAL A22=8521; /* discretization of A matrix */

        REAL u0max=10; /* upperbound on u0 */
        REAL e = 1e-6; /* precision for strict inequalities */
    }
IMPLEMENTATION {
    AXIOM {REAL z1,z2; }

    DA {
        z1 = (IF heat1 THEN Bd1*u0 [Bd1*u0max,0,e]);
        z2 = (IF heat2 THEN Bd2*u0 [Bd2*u0max,0,e]);
    }

    CONTINUOUS {
        t1 = A11*t1+z1;
        t2 = A22*t2+z2;
        u0 = u0;
    }

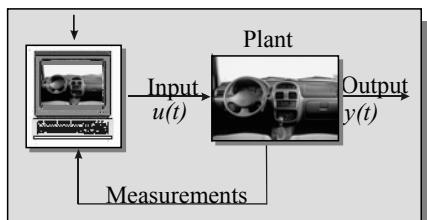
    HUST {
        -heat1 | -heat2; /* heat1 and heat2 cannot be both active */
    }
}
}
```



$$\begin{aligned} x(t+1) &= Ax(t) + B_1u(t) + B_2\delta(t) + B_3z(t) \\ y(t) &= Cx(t) + D_1u(t) + D_2\delta(t) + D_3z(t) \\ E_2\delta(t) + E_3z(t) &\leq E_4x(t) + E_1u(t) + E_5 \end{aligned}$$

Conclusions

- Hybrid systems as a framework for new applications, where both logic and continuous dynamics are relevant
- Mixed Logical Dynamical (MLD) systems as discrete-time, *computation-oriented* models for hybrid systems
- Supervisory MPC controllers and State Estimation/Fault Detection schemes can be synthesized via on-line mixed-integer programming (MILP/MIQP)
- Piecewise Linear Optimal Controllers can be synthesized via off-line multiparametric programming for fast-sampling applications



$$u(x) = \begin{cases} F_1x + G_1 & \text{if } H_1x \leq K_1 \\ \vdots & \vdots \\ F_Nx + G_N & \text{if } H_Nx \leq K_N \end{cases}$$

Conclusions

- **Hybrid systems** as a framework for new applications, where both logic and continuous dynamics are relevant
- **Mixed Logical Dynamical (MLD) systems** as discrete-time, *computation-oriented* models for hybrid systems
- **Supervisory MPC controllers** and **State Estimation/Fault Detection** schemes can be synthesized via on-line mixed-integer programming (MILP/MIQP)
- **Piecewise Linear Optimal Controllers** can be synthesized via off-line multiparametric programming for fast-sampling applications
- **Safety Analysis** properties can be formally verified

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- [2] A. Bemporad, F. Borrelli, and M. Morari. Piecewise linear optimal controllers for hybrid systems. In *American Control Conference*, pages 1190–1194, Chicago, IL, June 2000.
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- [8] W.P.M.H. Heemels, B. De Schutter, and A. Bemporad. Equivalence of hybrid dynamical models. *Automatica*, 37(7):1085–1091, July 2001.