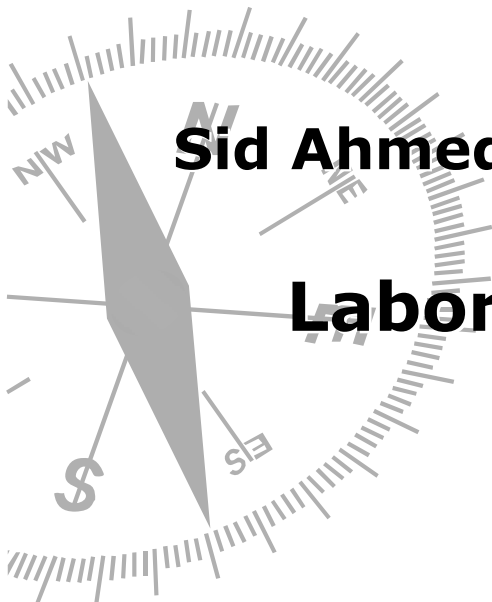


Analysis and Lyapunov-based hybrid control design for stabilization of the ABB Benchmark

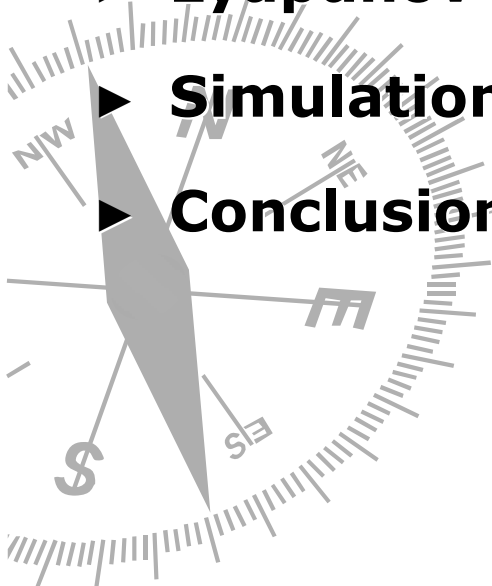
Sid Ahmed Attia, Carlos Canudas de Wit, Mazen Alamir

Laboratoire d'Automatique de Grenoble

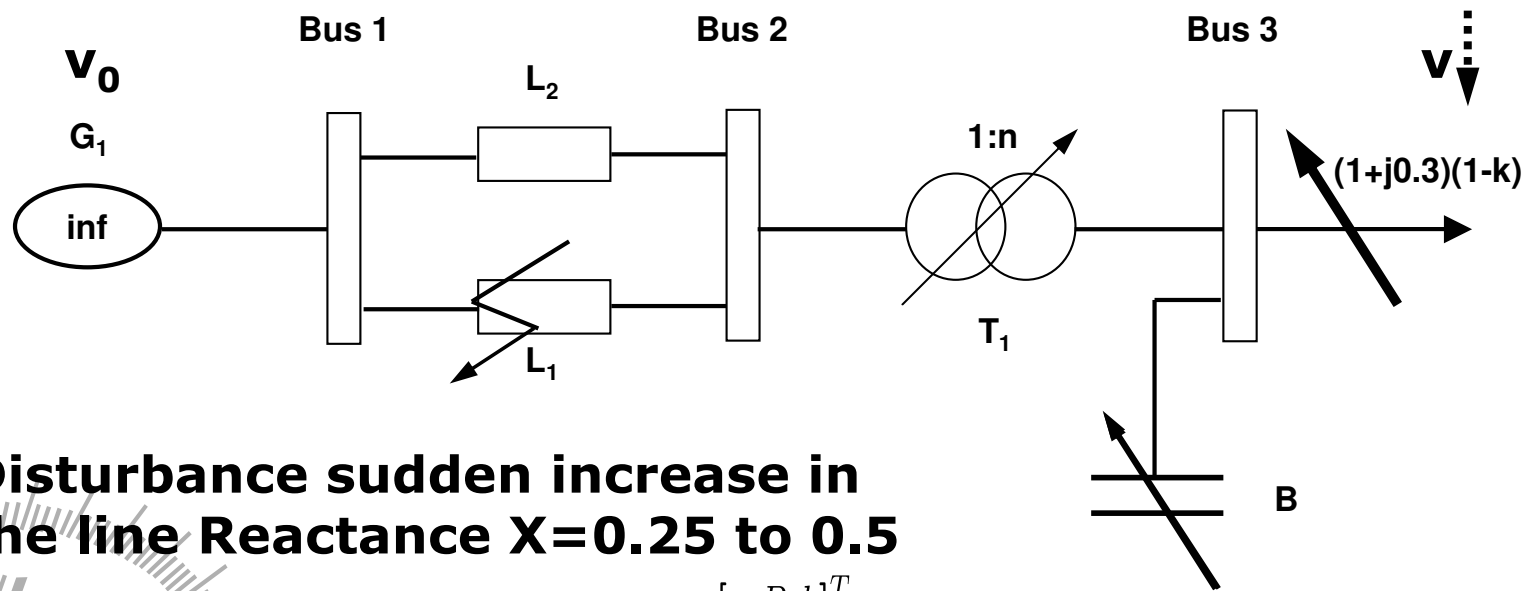


Outline

- ▶ **Simplified benchmark presentation**
- ▶ **Model properties**
- ▶ **Lyapunov-based hybrid control design**
- ▶ **Simulation results**
- ▶ **Conclusions and future work**



System Presentation



**Disturbance sudden increase in
the line Reactance $X=0.25$ to 0.5**

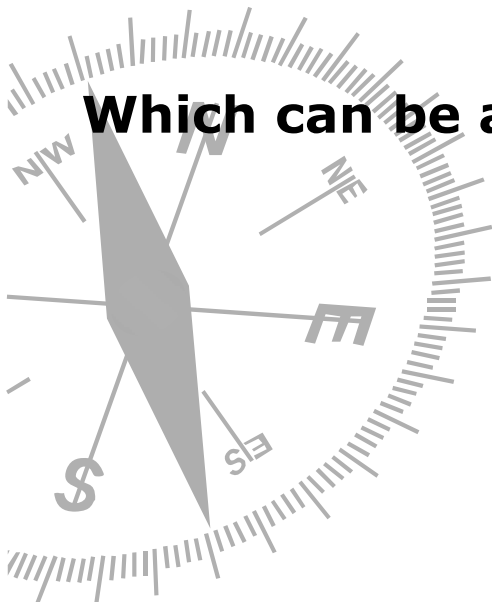
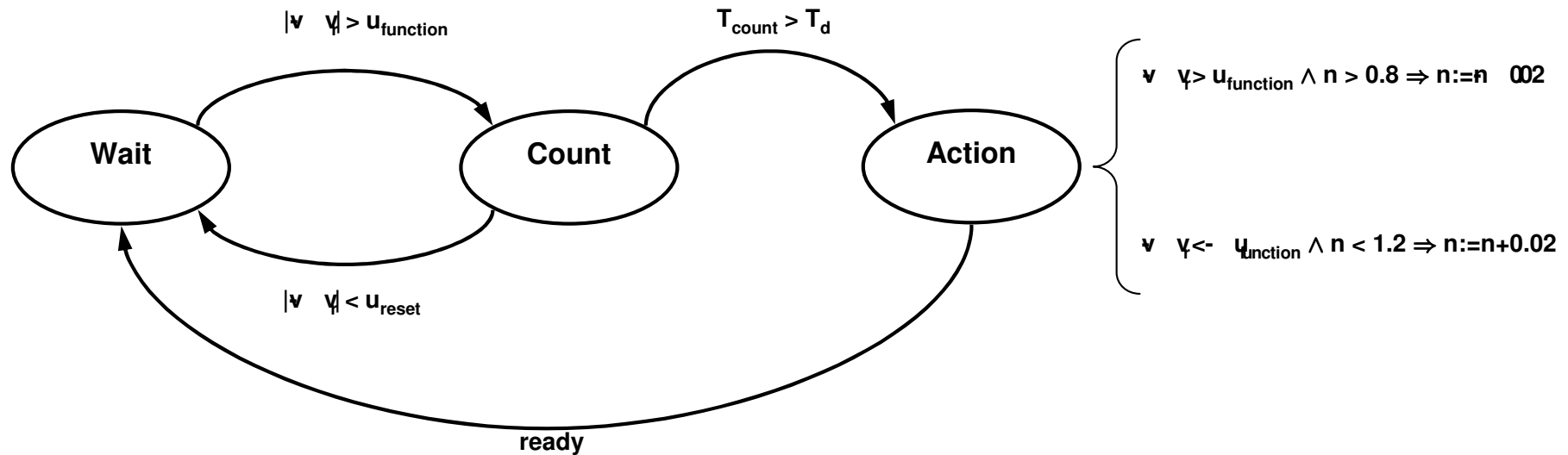
$$u = [n \ B \ k]^T$$

$n \in [n_m \ n_M] \quad n_m = 0.8, \ n_M = 1.2 \quad \text{With steps of } 0.02$

$B \in \{0, 0.1, 0.2, 0.3\}$

$k \in \{0, 0.05, 0.1, 0.15\}$

OLTC dynamics



Which can be approximated by

$$\frac{dn}{dt} = \Gamma \text{sgn}(\nu, n, \Delta)$$

Model properties

The dynamic load model

The system is described by the following

$$\dot{x}_p = -\frac{x_p}{T_p} + P_0 (v^{\alpha_s} - v^{\alpha_t})$$

$$\dot{x}_q = -\frac{x_q}{T_q} + Q_0 (v^{\beta_s} - v^{\beta_t})$$

$$P = (1 - k) \left(\frac{x_p}{T_p} + P_0 v^{\alpha_t} \right)$$

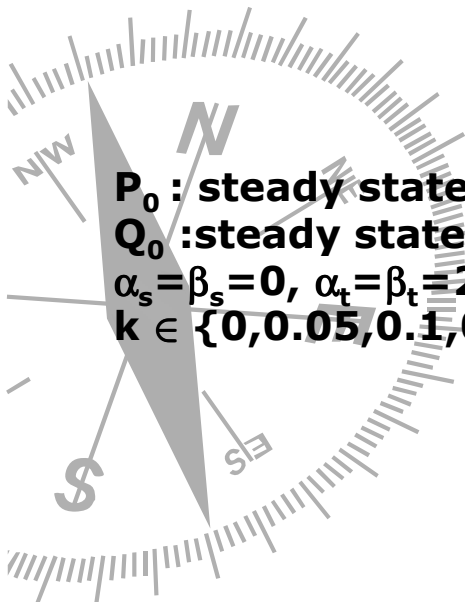
$$Q = (1 - k) \left(\frac{x_q}{T_q} + Q_0 v^{\beta_t} \right)$$

P_0 : steady state value of the active power demand

Q_0 : steady state of the reactive power demand

$\alpha_s = \beta_s = 0, \alpha_t = \beta_t = 2, T_p = T_q = 60 \text{ sec}$

$k \in \{0, 0.05, 0.1, 0.15\}$ is the load shedding control input



The DAE nature of the problem

The load flow equations add the following constraints to the dynamics

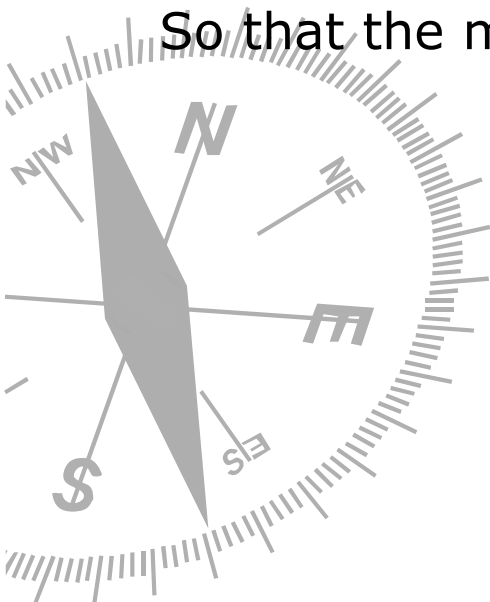
$$(1 - k) \left(\frac{x_p}{T_p} + P_0 v^2 \right) = -\frac{v_0 v}{nX} \sin(\delta)$$

$$(1 - k) \left(\frac{x_q}{T_q} + Q_0 v^2 \right) = \frac{v_0 v}{nX} \cos(\delta) - \frac{v^2}{n^2 X} + Bv^2$$

So that the model can be written in the DAE framework

$$\dot{x} = f(x, v)$$

$$0 = g(x, u, v, \theta)$$



An invariance property

It can be shown that

$$F(x_p, x_q) = P_0 x_q - Q_0 x_p = 0$$

is an invariant manifold

So that under $x(0) \in F$

The system reduces to

$$\dot{x} = -\frac{x}{T_p} + P[1 - y(x, u)]$$

$$0 = y^2 + 2\alpha_1(x, u)y + \alpha_0(x, u)$$

$$y = v^2$$

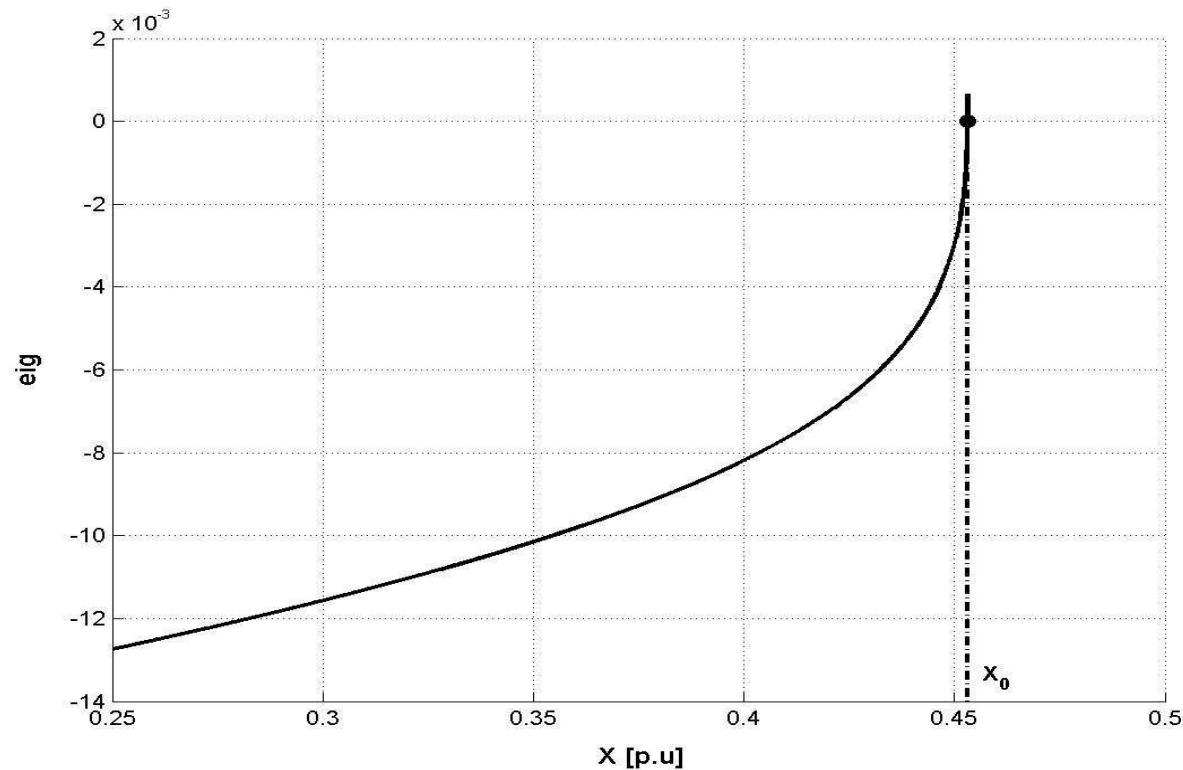
where



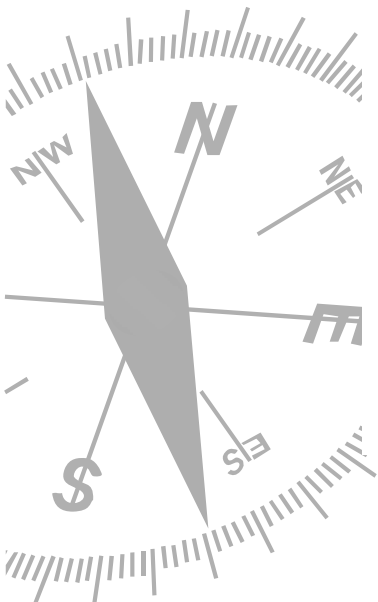
Equilibrium points

Linearization of the system around an equilibria x^*

$$\dot{\tilde{x}} = c\tilde{x}$$

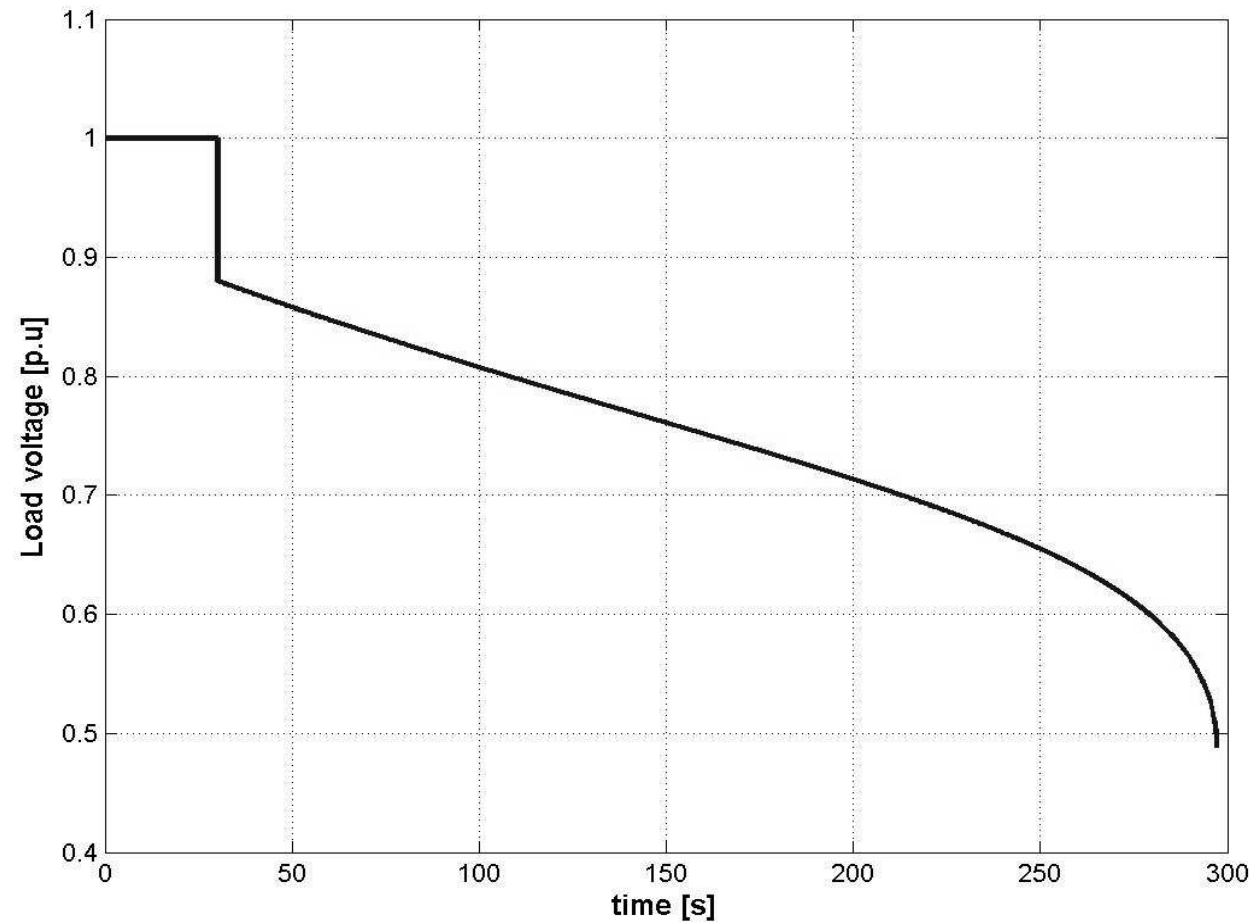


**c as a function of the line reactance X
for $n_n=1$, $B_n=0$, $k_n=0$**



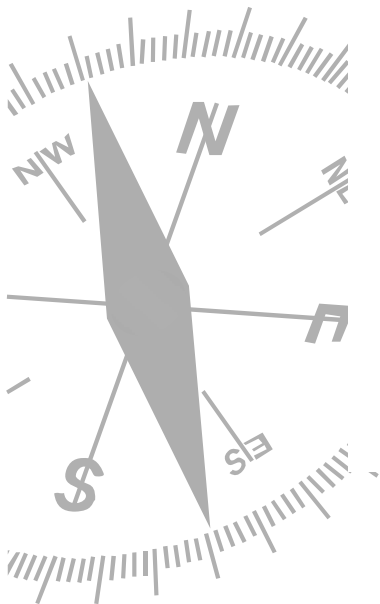
Voltage collapse dynamics

Time domain simulation



Voltage profile for X=0.5

CC project meeting, Amsterdam 16-17 june
2003



Voltage collapse dynamics

State space interpretation

From equation

$$\frac{dg}{dt} = \frac{\partial g}{\partial y} \dot{y} + \frac{\partial g}{\partial x} \dot{x} = 0$$

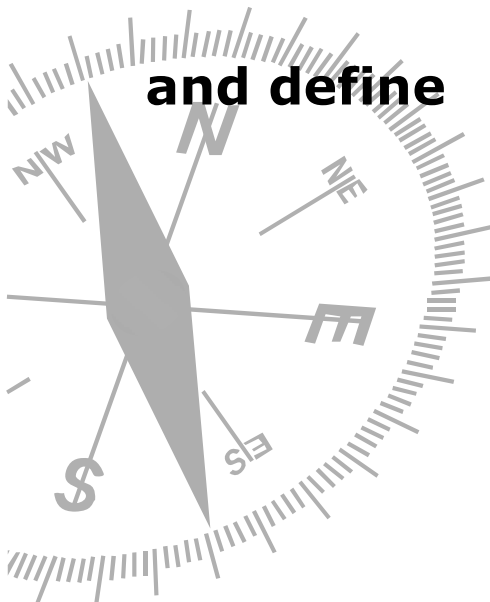
One can writes

$$\dot{y} = - \left(\frac{\partial g}{\partial y} \right)^{-1} \frac{\partial g}{\partial x} \dot{x}$$

and define

$$S = \{(x, y) \in L : \Delta(x, y) = \det \frac{\partial g}{\partial y} = 0\}$$

as the singular surface

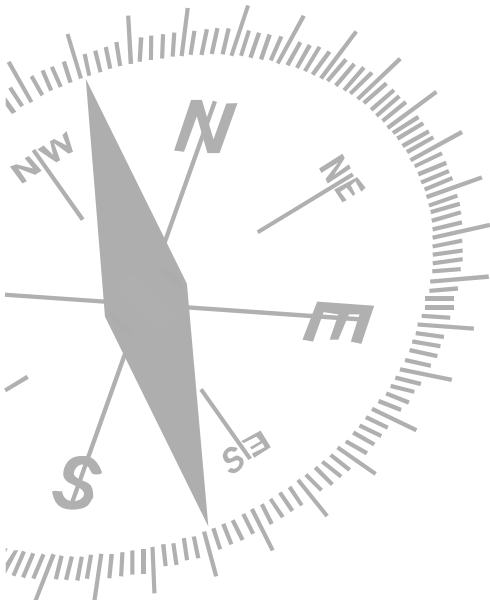


Voltage collapse dynamics

State space interpretation

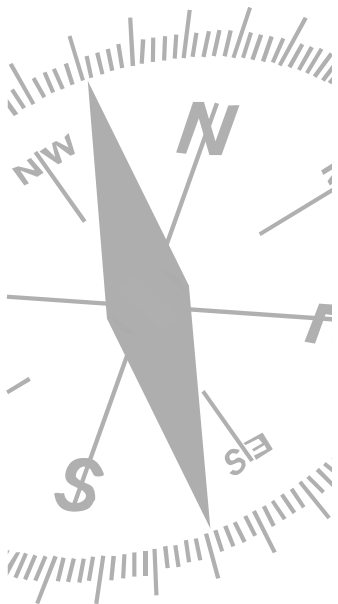
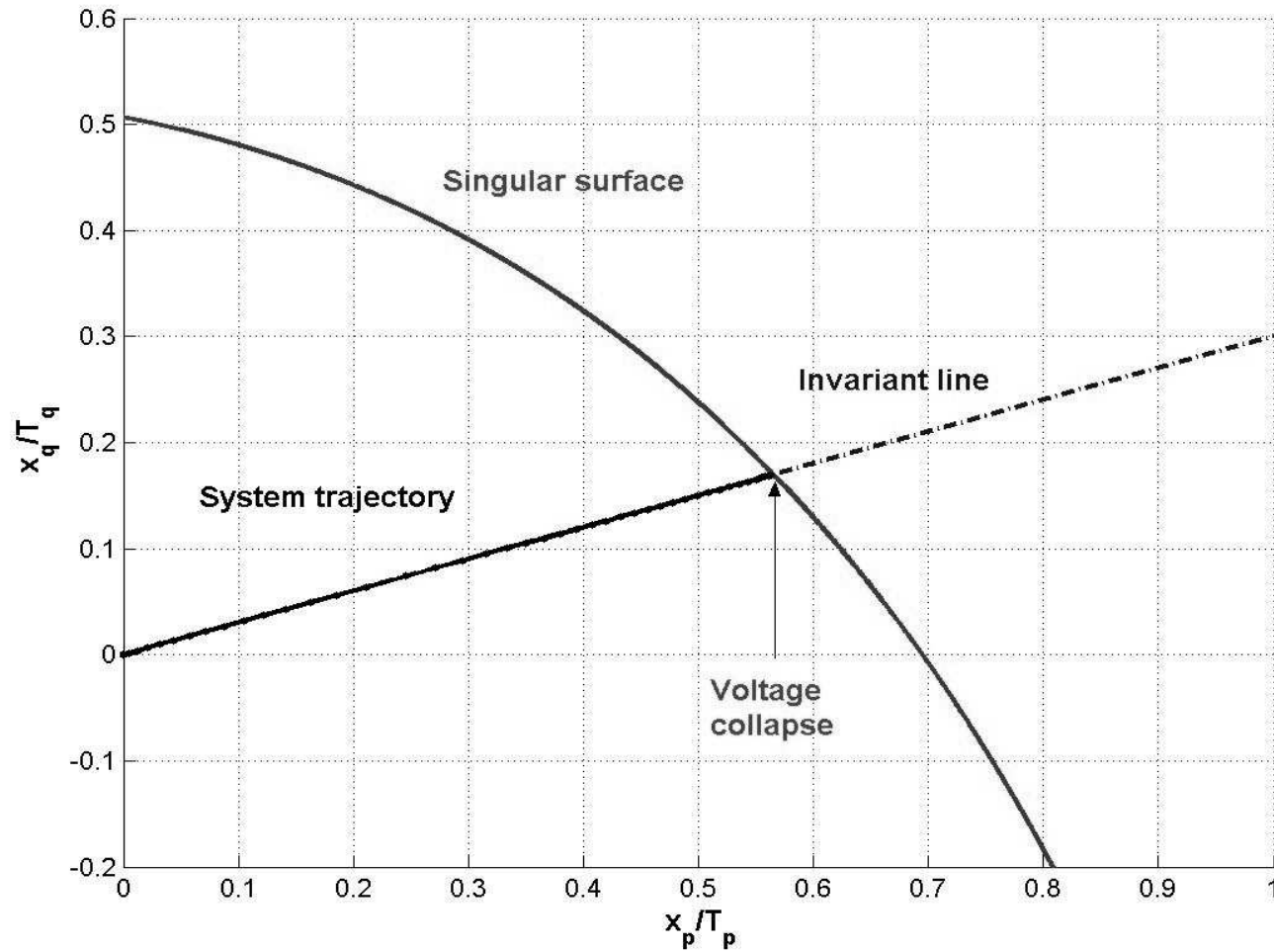
Which gives explicitly

$$\dot{y} = \frac{1}{2(y + \alpha_1(x, u))} \left[2 \frac{\partial \alpha_1}{\partial x}(x, u) y + \frac{\partial \alpha_0}{\partial x}(x, u) \right] \dot{x}$$

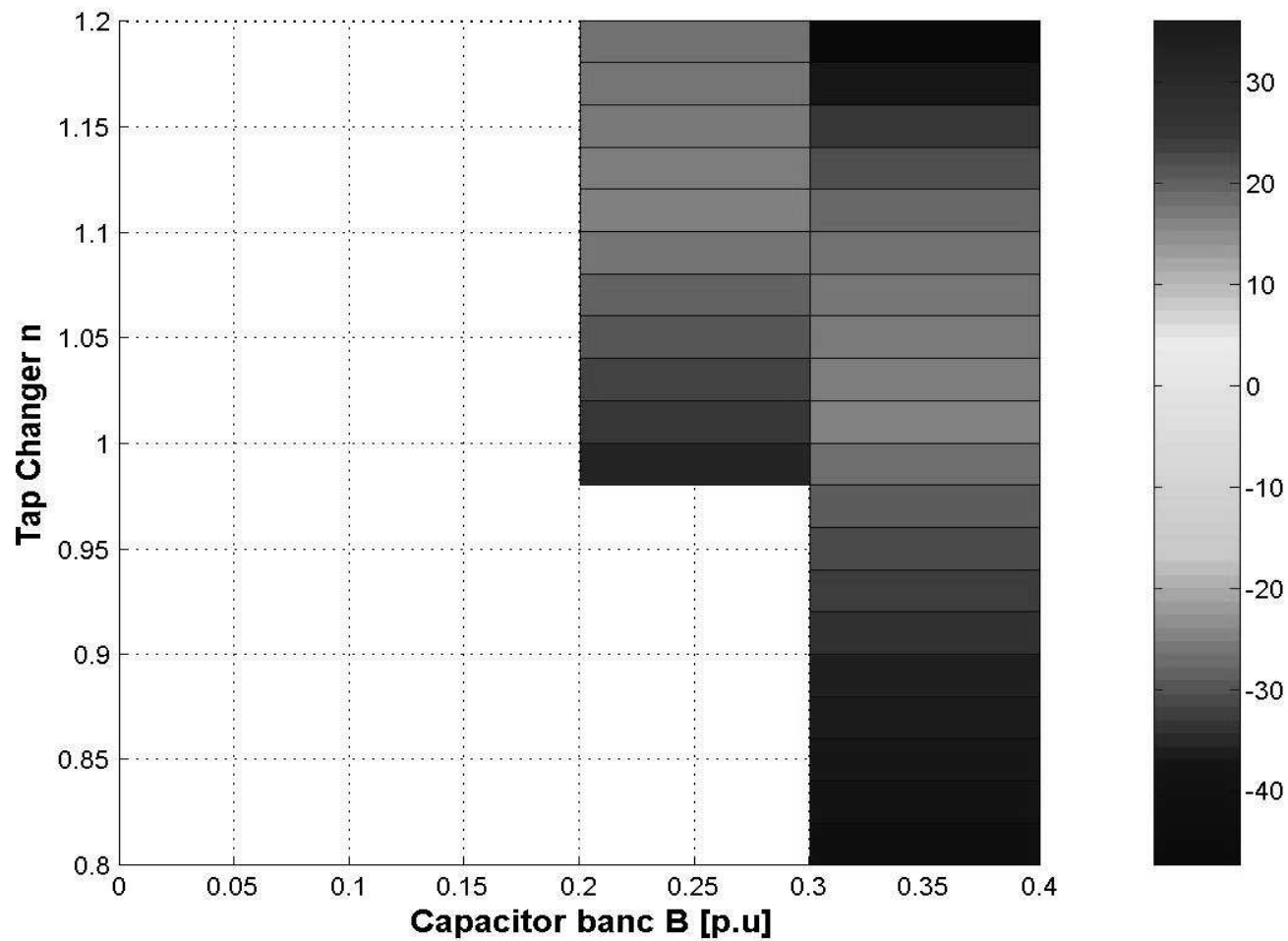


Voltage collapse dynamics

State space interpretation

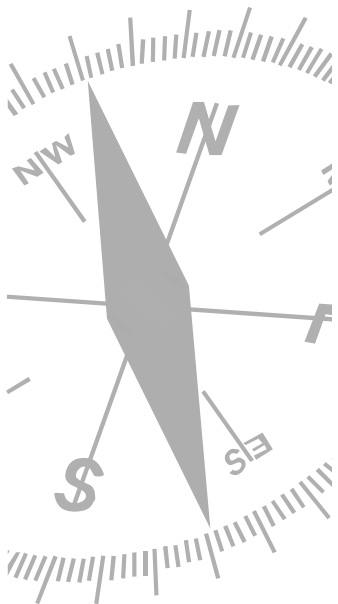


Stationary Points

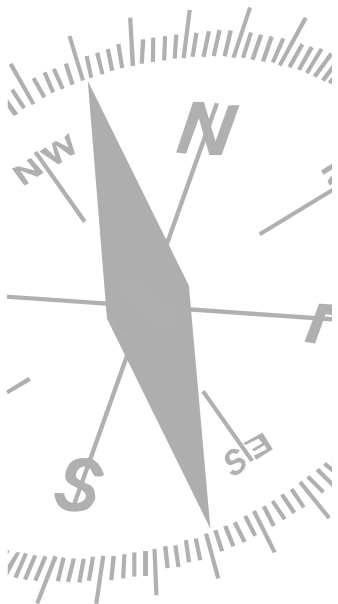
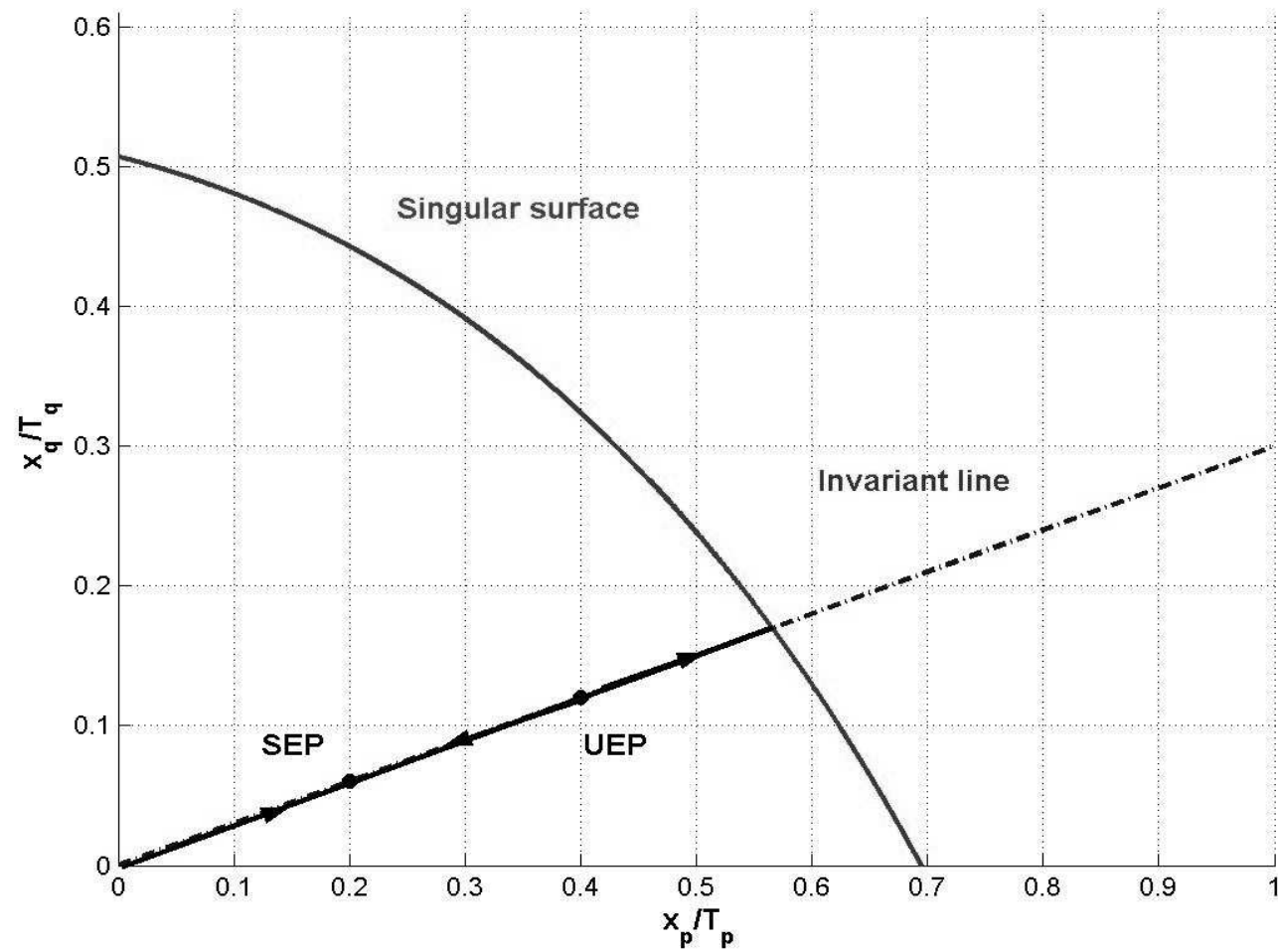


Stationary points for $X=0.5$, $k=0$

CC project meeting, Amsterdam 16-17 june
2003

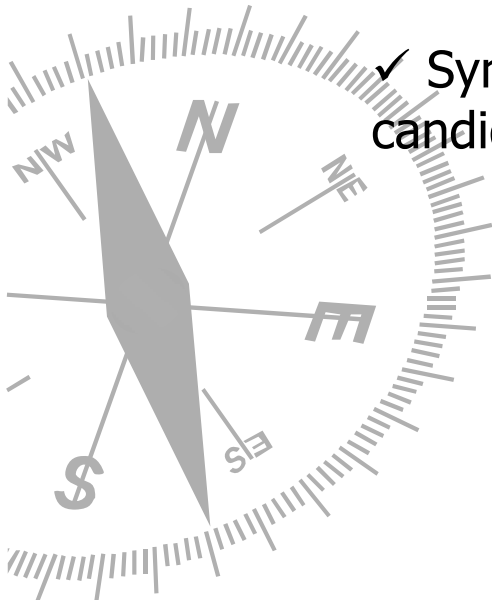


A possible strategy



Lyapunov-based hybrid control

- The general idea is based on the control Lyapunov functions
 - ✓ Take a Lyapunov function candidate
 - ✓ Synthesize a feedback law such that the Lyapunov candidate is decreased



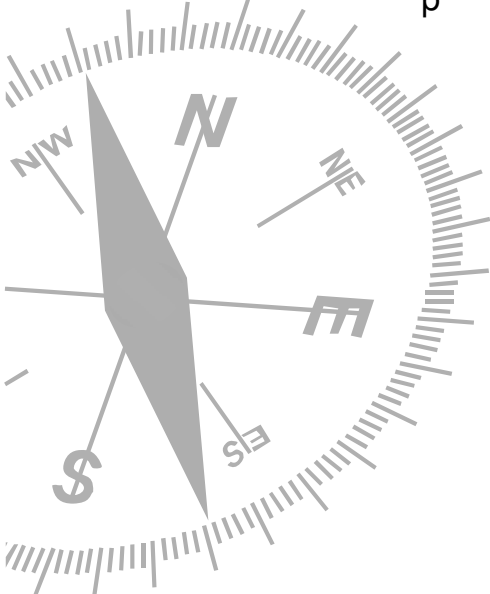
Algorithm formulation

For convenience the model is written

$$\begin{aligned}\dot{x} &= -ax + b[1 - y(x, u)] \\ 0 &= y^2 + 2\alpha_1(x, u)y + \alpha_0(x, u)\end{aligned}$$

With $a=1/T_p$ and $b=P_0$

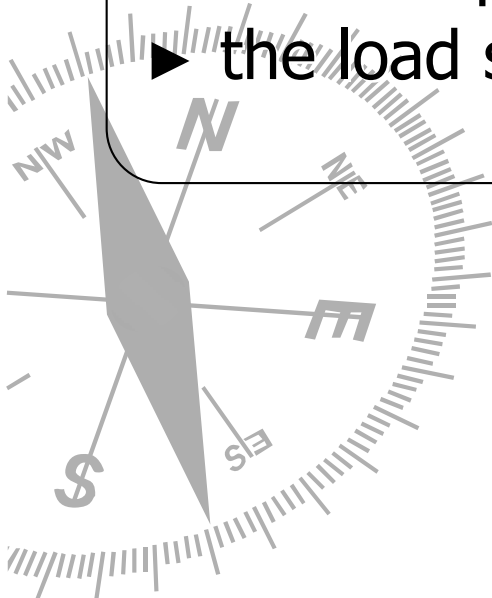
$$u^T = [n, B, k]$$



Algorithm formulation

For convenience the control constraints are repeated

- ▶ the tap ratio $n \in [0.8, 1.2]$ with steps of **0.02 p.u.**
- ▶ the capacitor bank $B \in [0, 0.3]$ with steps of **0.1 p.u.** of reactive power compensation.
- ▶ the load shedding $k \in [0, 0.15]$ with steps of **0.05 p.u.**



Algorithm formulation

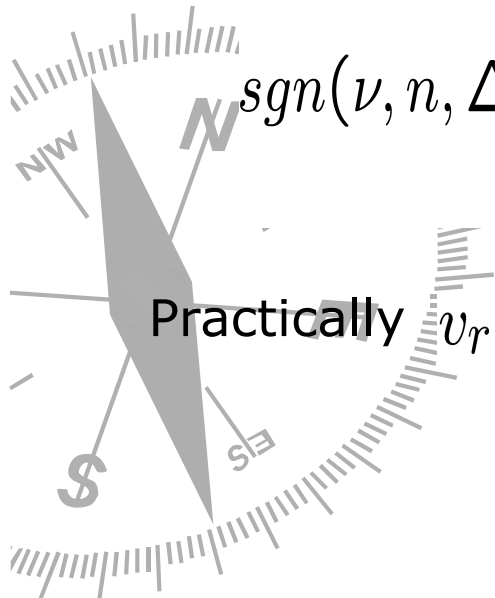
For synthesis purposes the following model

$$\frac{dn}{dt} = \Gamma \operatorname{sgn}(\nu, n, \Delta)$$

Where $\operatorname{sgn}(\cdot, \cdot, \cdot)$ is defined as

$$\operatorname{sgn}(\nu, n, \Delta) = \begin{cases} 1 & \text{if } (\nu > \Delta) \wedge (n < n_M) \\ -1 & \text{if } (\nu < -\Delta) \wedge (n > n_m) \\ 0 & \text{if } (|\nu| \leq \Delta) \vee (n = n_m) \vee (n = n_M) \end{cases}$$

Practically $v_r = \nu + v$ is used to drive the OLTC

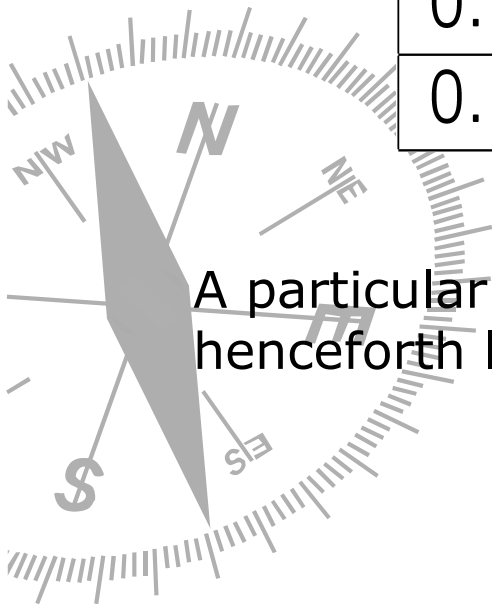


Algorithm formulation

The other control inputs **B** and **k** are casted in a strictly ordered set

k/B	0.0	0.1	0.2	0.3
0.00	$j = 1$	$j = 2$	$j = 3$	$j = 4$
0.05	$j = 5$	$j = 6$	$j = 7$	$j = 8$
0.10	$j = 9$	$j = 10$	$j = 11$	$j = 12$
0.15	$j = 13$	$j = 14$	$j = 15$	$j = 16$

A particular selection of a pair (B, k) will be denoted u_j and is henceforth labelled by an integer j .



Algorithm formulation

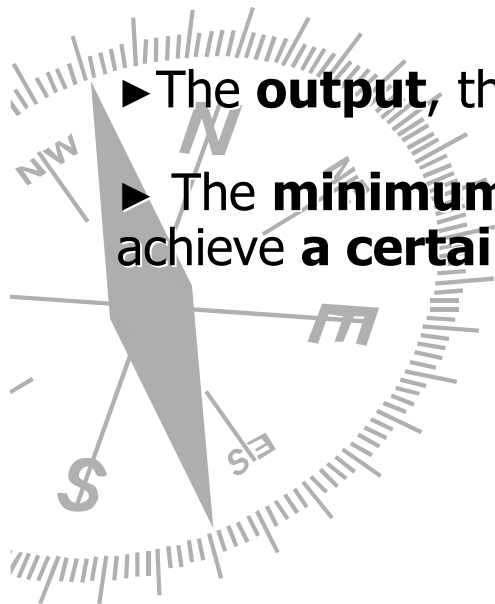
Suppose when the system is in fault

$$\exists (x^*, u^*) \quad \text{s.t.}$$

$$0 = -ax^* + b[1 - y^*]$$

$$0 = y^{*2} + 2\alpha_1^* y^* + \alpha_0^*$$

The set of possible (x^*, u^*) is in general not a singleton



► The **output**, the voltage in this case is **above** or **between certain limits**;

► The **minimum** value of the **capacitor bank** and **load shedding** is used to achieve **a certain closed loop characteristics**.

Algorithm formulation

The dynamic equations can be written in the error coordinates

$$\tilde{x} = x - x^*, \quad \tilde{u} = u - u^*$$

as

$$\dot{\tilde{x}} = -a\tilde{x} + b[y^* - y^+(x, u)]$$

$$\dot{\tilde{n}} = \Gamma \operatorname{sgn}(\nu, n, \Delta)$$

$$0 = y^2 + 2\alpha_1(x, u)y + \alpha_0(x, u)$$



Control design

Consider the following Lyapunov function candidate

$$V(z) = \frac{z^T z}{2}$$

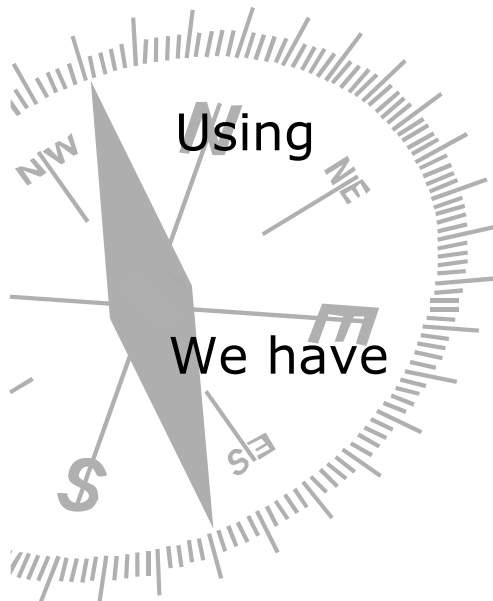
$$z^T = [\ddot{\tilde{x}} \quad \tilde{x} \quad \tilde{n}]$$

So that

$$\begin{aligned} \dot{V} &= \frac{\partial V}{\partial z} \dot{z} \\ &= \ddot{\tilde{x}} \ddot{\tilde{x}} + \tilde{x} \ddot{\tilde{x}} + \tilde{n} \ddot{\tilde{n}} \end{aligned}$$

$$\ddot{\tilde{x}} = -a\tilde{x} + b[y^* - y^+(x, u)]$$

$$\dot{V} = -\left(a + b \frac{\partial \tilde{y}}{\partial x}\right) \dot{\tilde{x}}^2 - \left(b \frac{\partial \tilde{y}}{\partial n} - \tilde{n}\right) \dot{\tilde{n}} + \tilde{x} \ddot{\tilde{x}} - b \dot{\tilde{x}} \left(\frac{\partial \tilde{y}}{\partial u_j}\right)^T \dot{u}_j$$



Control design

Taking into account the update law

$$\dot{\tilde{n}} = \Gamma \operatorname{sgn}(\nu, n, \Delta)$$

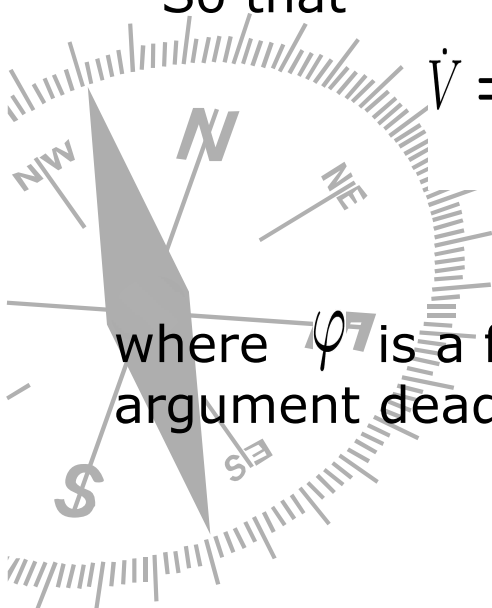
The signal ν is taken as

$$\nu = \left(b \frac{\partial \tilde{y}}{\partial n} - \tilde{n} \right)$$

So that

$$\dot{V} = - \left(a + b \frac{\partial \tilde{y}}{\partial x} \right) \dot{\tilde{x}}^2 - \varphi \left(b \frac{\partial \tilde{y}}{\partial n} - \tilde{n}, \Delta \right) + \tilde{x} \dot{\tilde{x}} - b \dot{\tilde{x}} \left(\frac{\partial \tilde{y}}{\partial u_j} \right)^T \dot{u}_j$$

where φ is a function returning the absolute value of its first argument dead zoned with the parameter Δ .



Control design

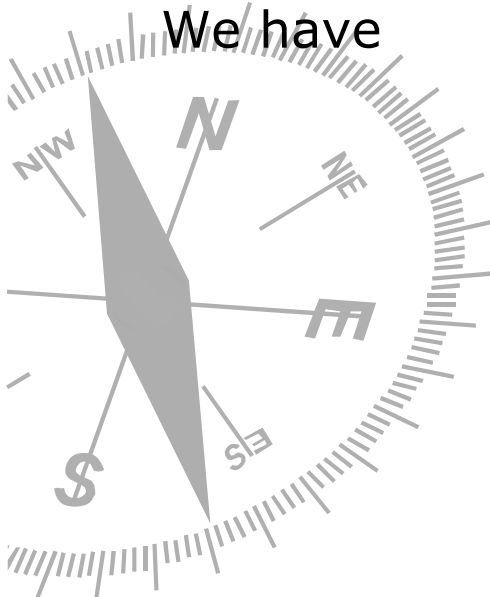
Defining

$$f_{u_j}^T = \tilde{x} b \left(\frac{\partial \tilde{y}}{\partial u_j} \right)^T, \quad \delta u_j = \dot{u}_j$$

And

$$W = \left(a + b \frac{\partial \tilde{y}}{\partial x} \right) \dot{\tilde{x}}^2 + \varphi \left(b \frac{\partial \tilde{y}}{\partial n} \dot{\tilde{x}} + \tilde{n} \right) - \tilde{x} \dot{\tilde{x}}$$

We have



$$\dot{V} = -W(x, n, u_j) - f_{u_j}^T \delta u_j$$

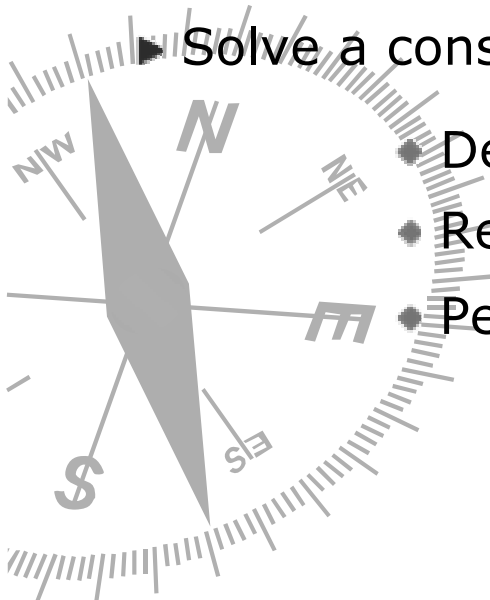
Lyapunov Based Logic Switching Strategy LBLSS

► Make use of the concept of supervisory control at the higher level

- The Lyapunov function is used as a monitoring signal
- Hysteresis switching

► Solve a constrained optimization problem at the lower level

- Decayedness of the Lyapunov function
- Respect a safety margin
- Penalty on switching



LBLSS

1. Let $t:=0$; $J = \{1,2, \dots, N\}$; $R \in \mathbb{R}_+$; $[\varepsilon_1 \ \varepsilon_2 \ \varepsilon_3]^T \in \mathbb{R}_+^3$; $\varepsilon_2 > \varepsilon_1$; $d_0 \in \mathbb{R}_+$;
2. $t:=t+1$;
3. $C_k := \sigma(V, \varepsilon_1, \varepsilon_2)$;
4. If C_k

- a) Among the admissible values for j_k find j_k^* from

$$j_k^* = \arg \min_{j_k \in \mathcal{J}} \{j_k^2 + R (j_k - j_{k-1}^*)^2 \}$$

under

- i. $W=W_{j_k^*} \geq \varepsilon_3$

- ii. $d \triangleq x_{s2}(j_k^*) - x \geq d_0$

5. else

- a) Among the admissible values for j_k find j_k^* from

$$j_k^* = \min \{j_k : j_k \in \mathcal{J}\}$$

for which

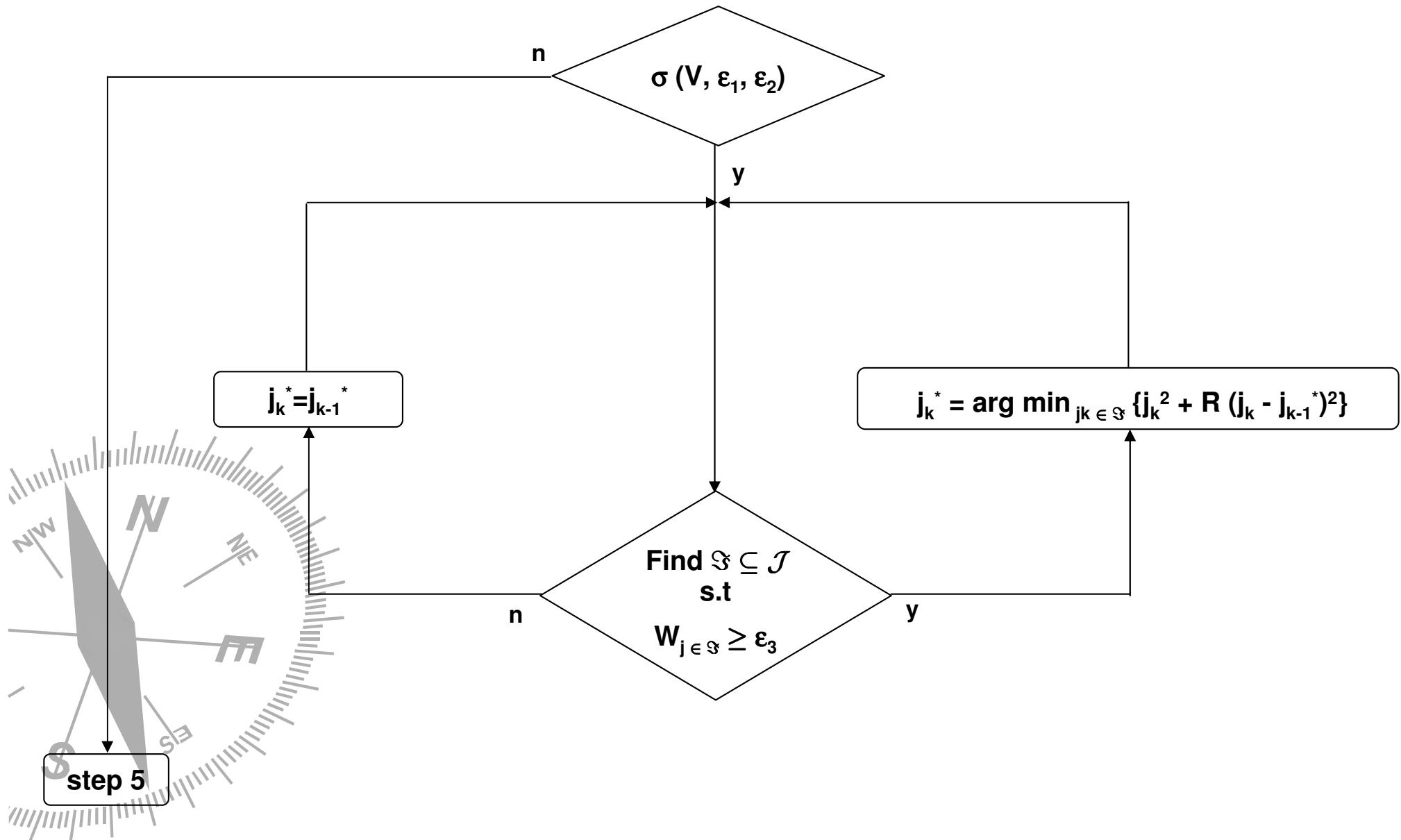
$$0=f(x^*, y^*, j_k^*)$$

$$0=y^{*2} + \alpha_{1,j_k^*} y^* + \alpha_{0,j_k^*}$$

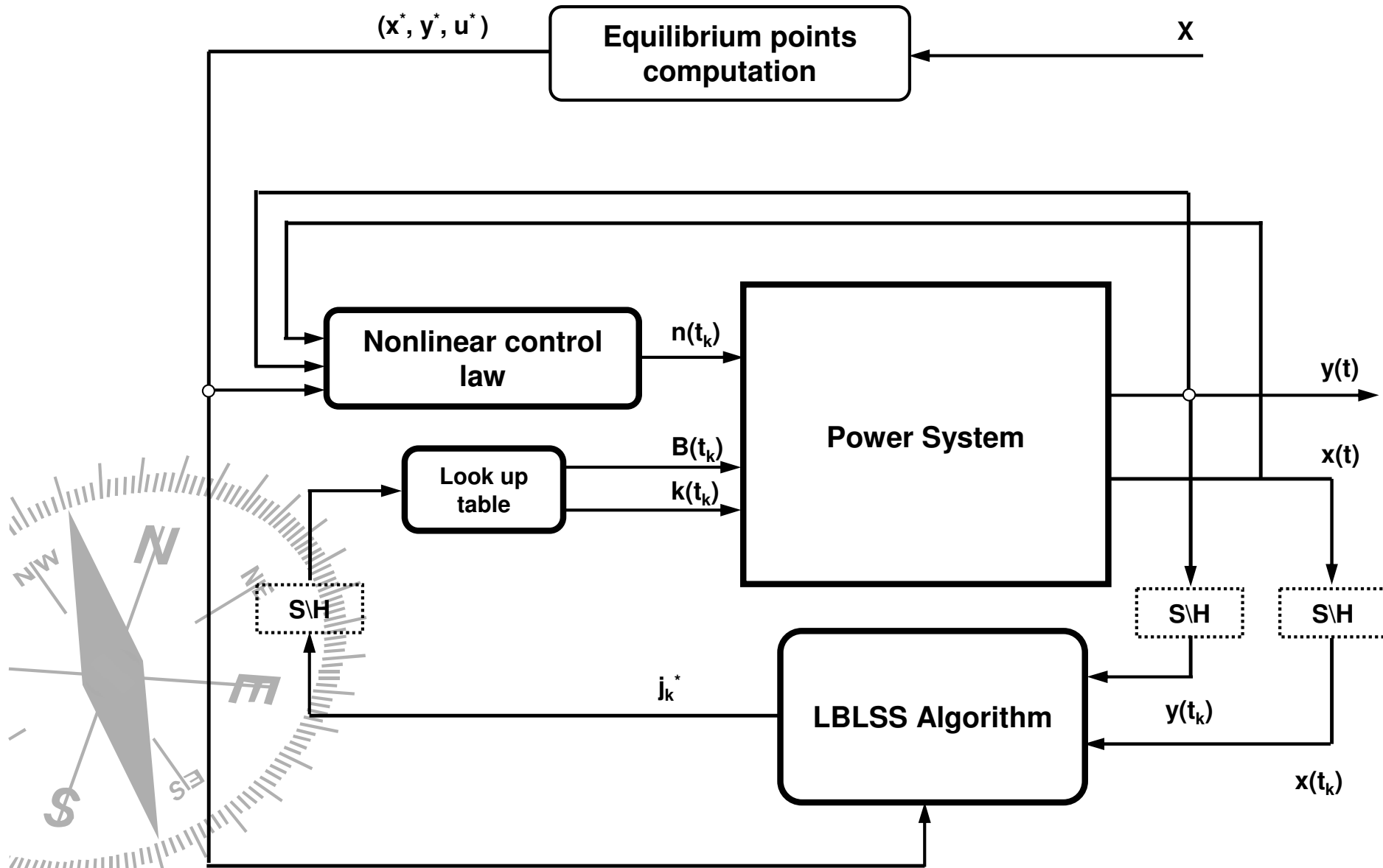
6. Goto 2;



Computer diagram

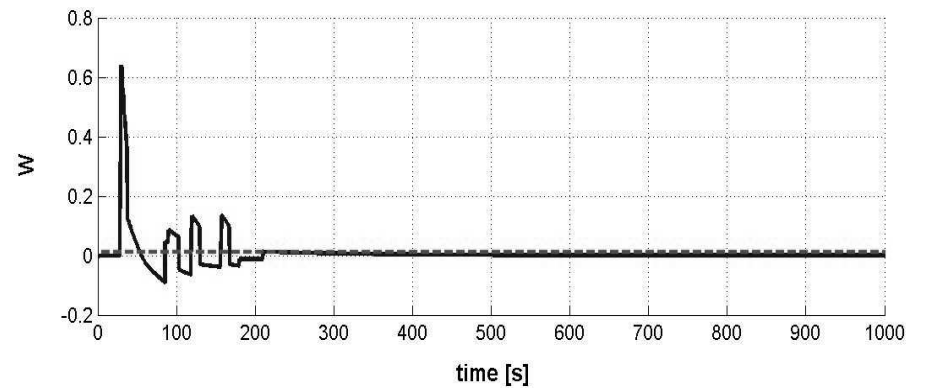
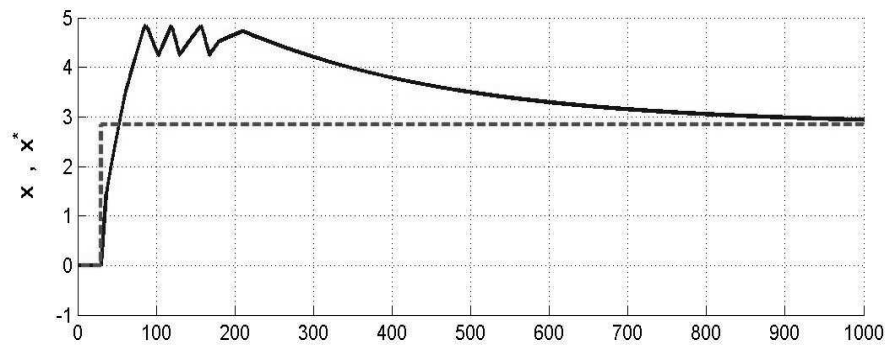
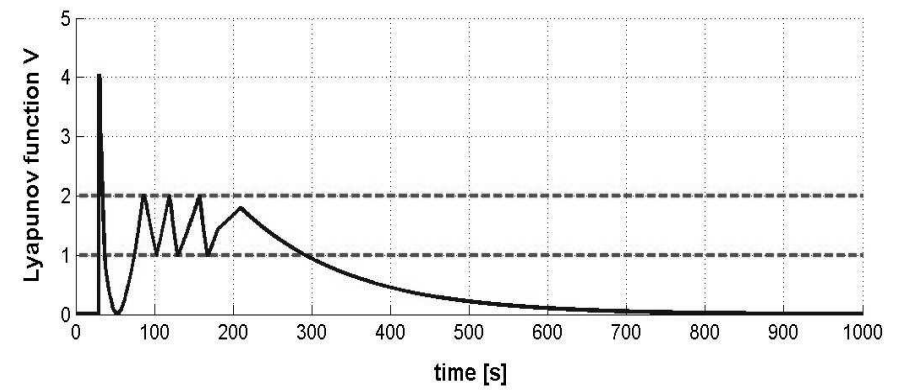
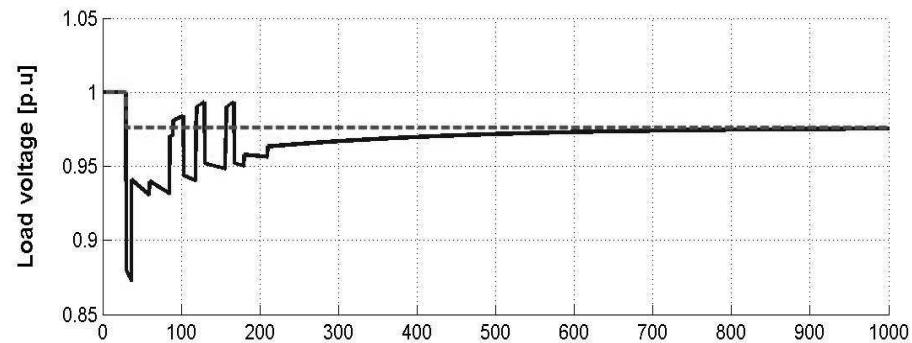


Closed loop system

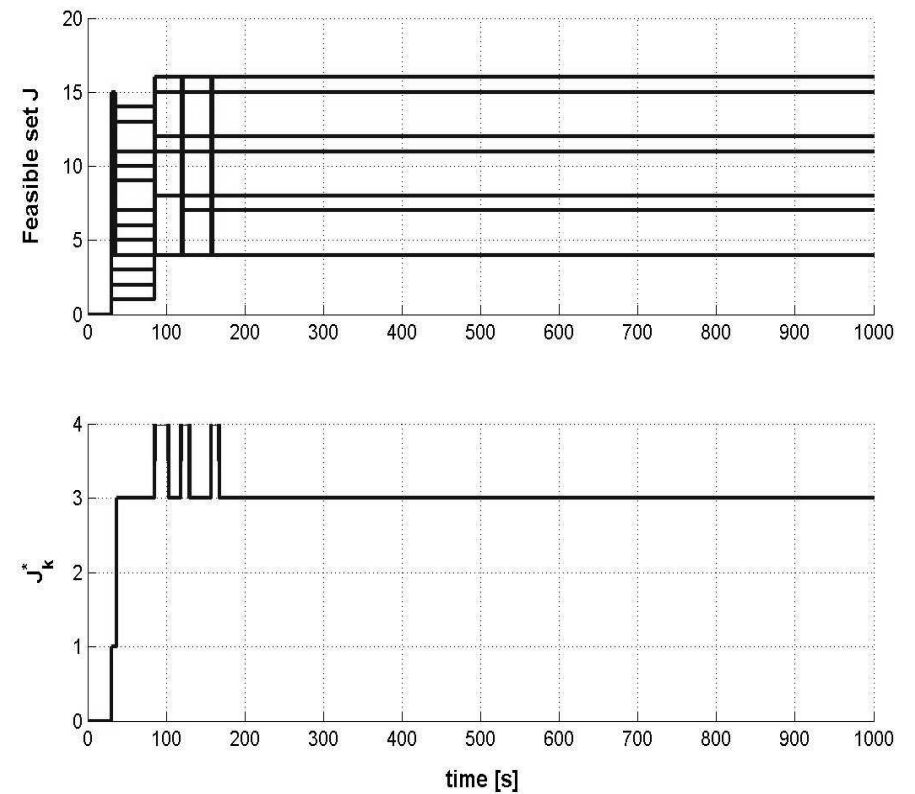
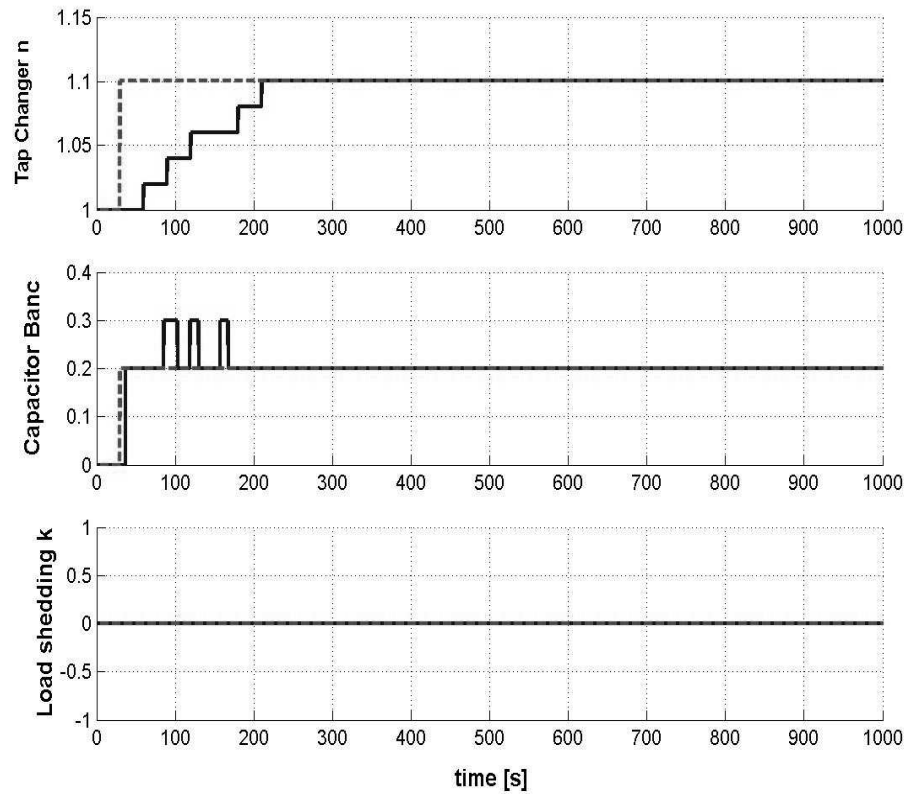


Simulation results

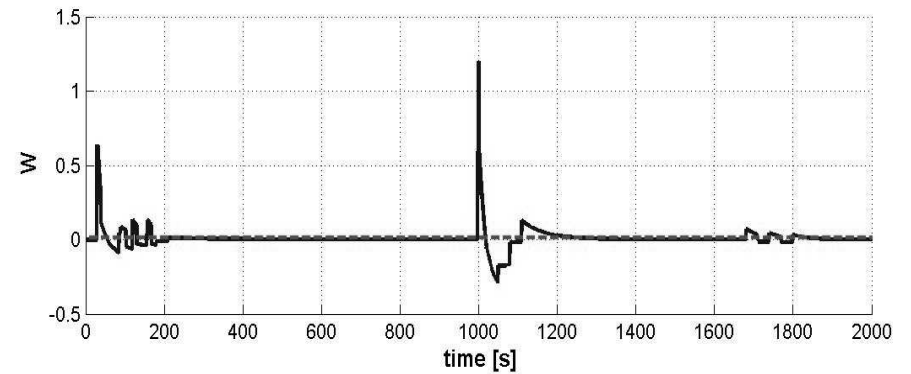
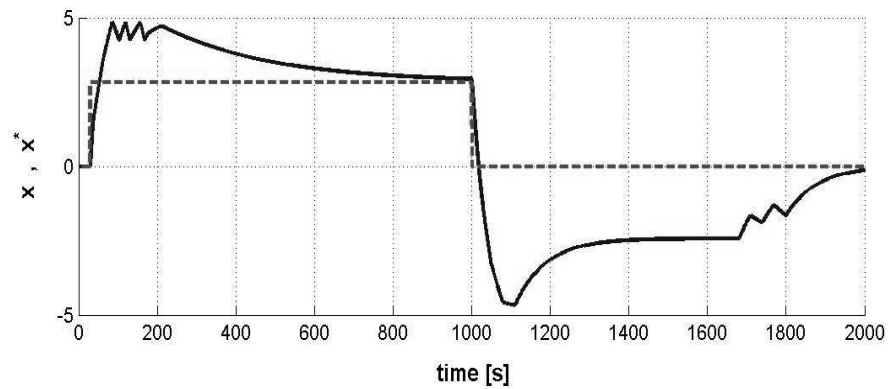
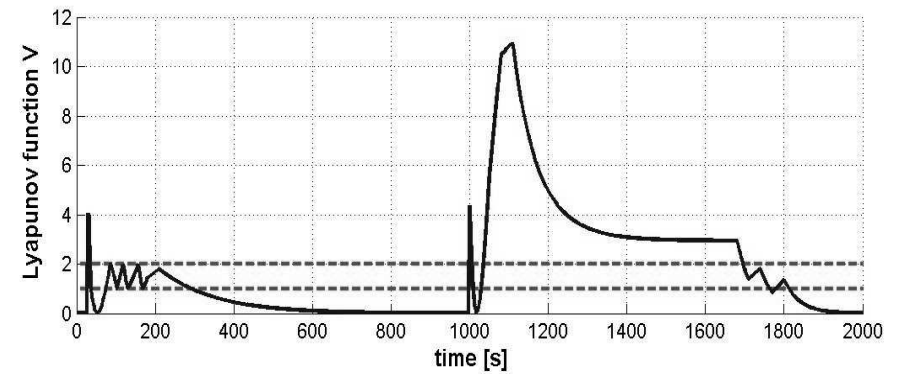
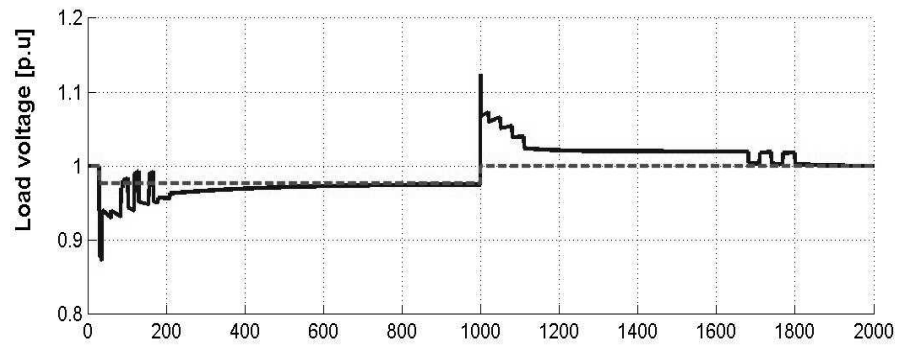
Nominal Case



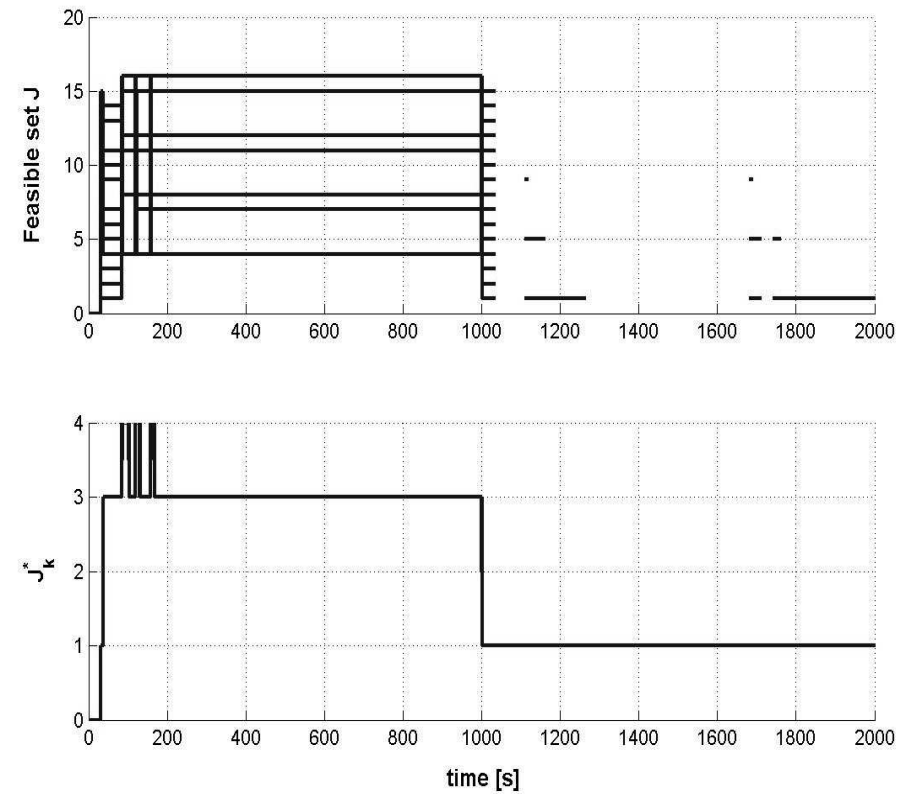
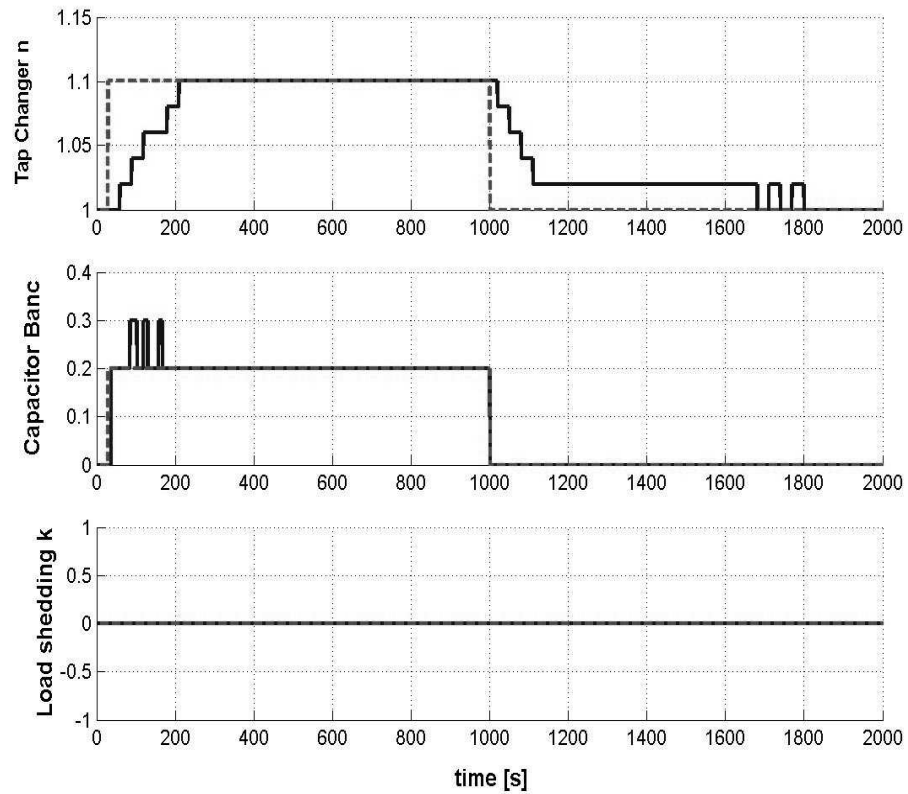
Nominal Case



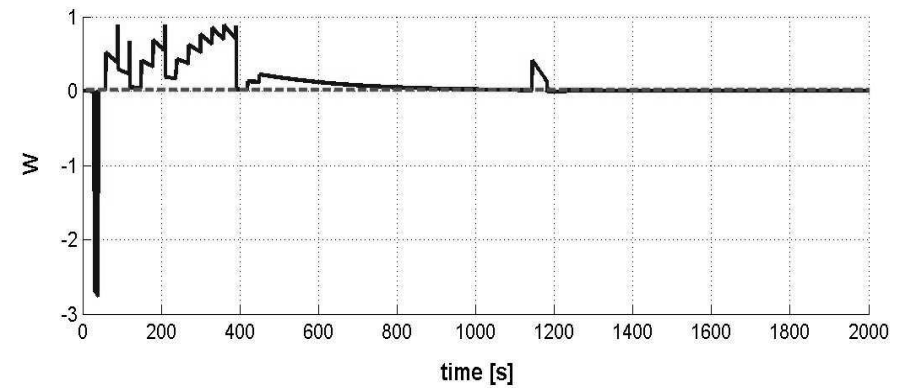
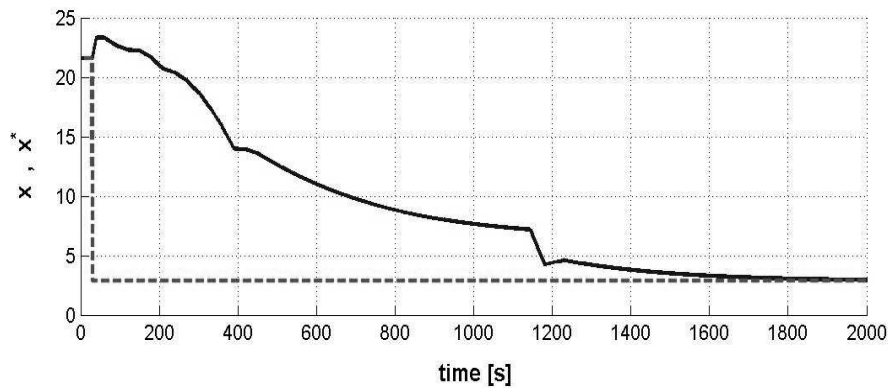
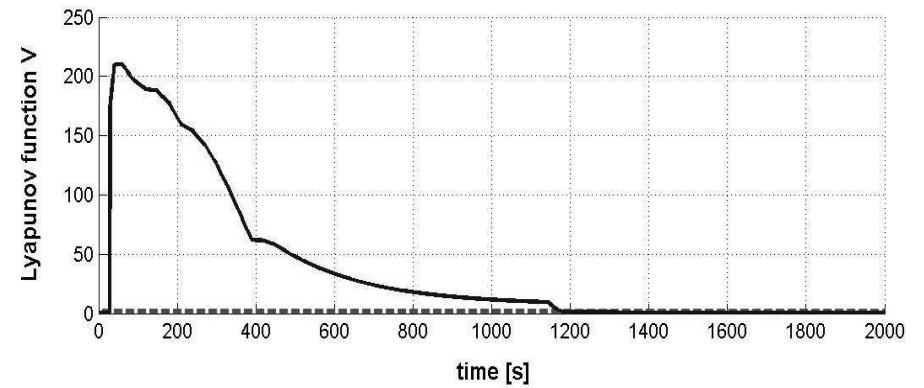
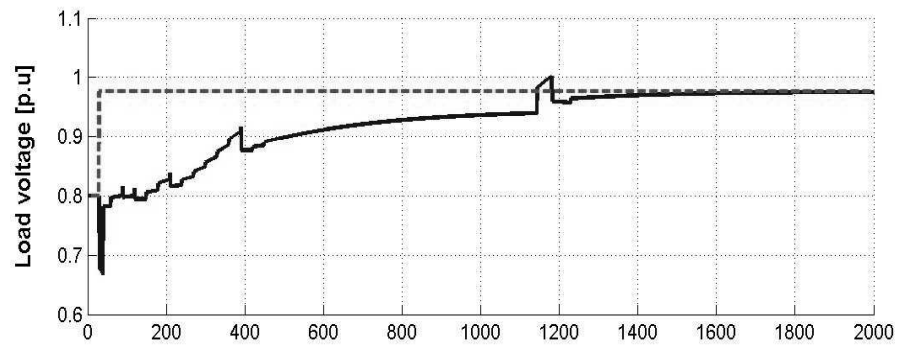
Fault clearance



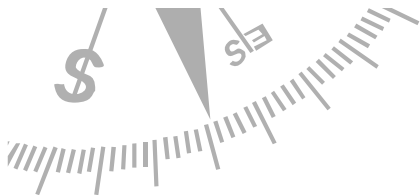
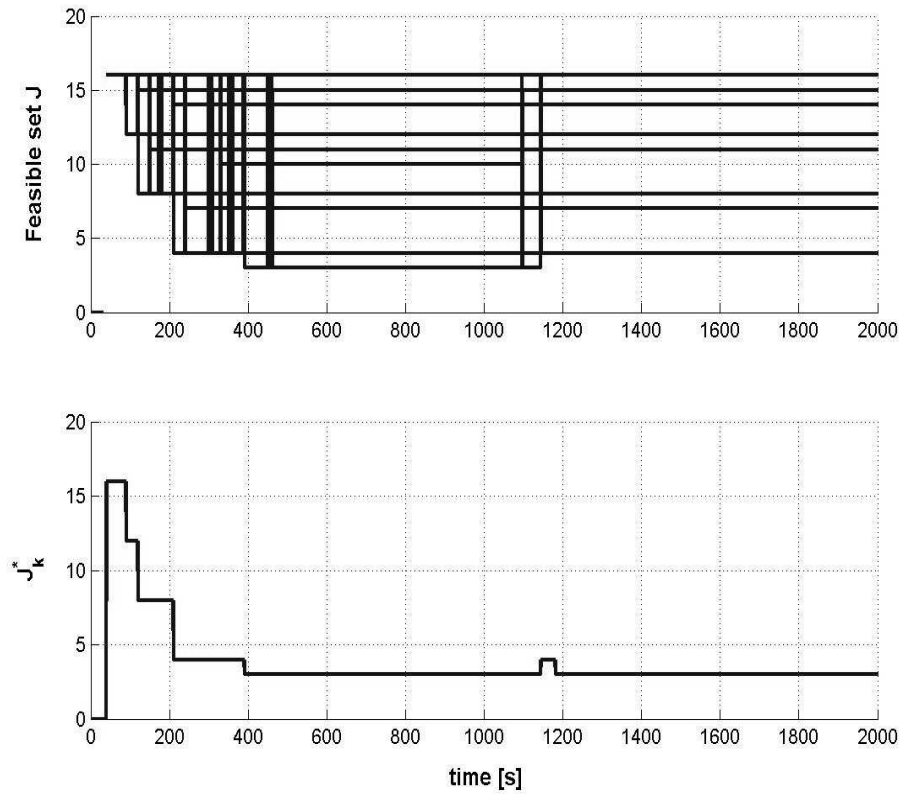
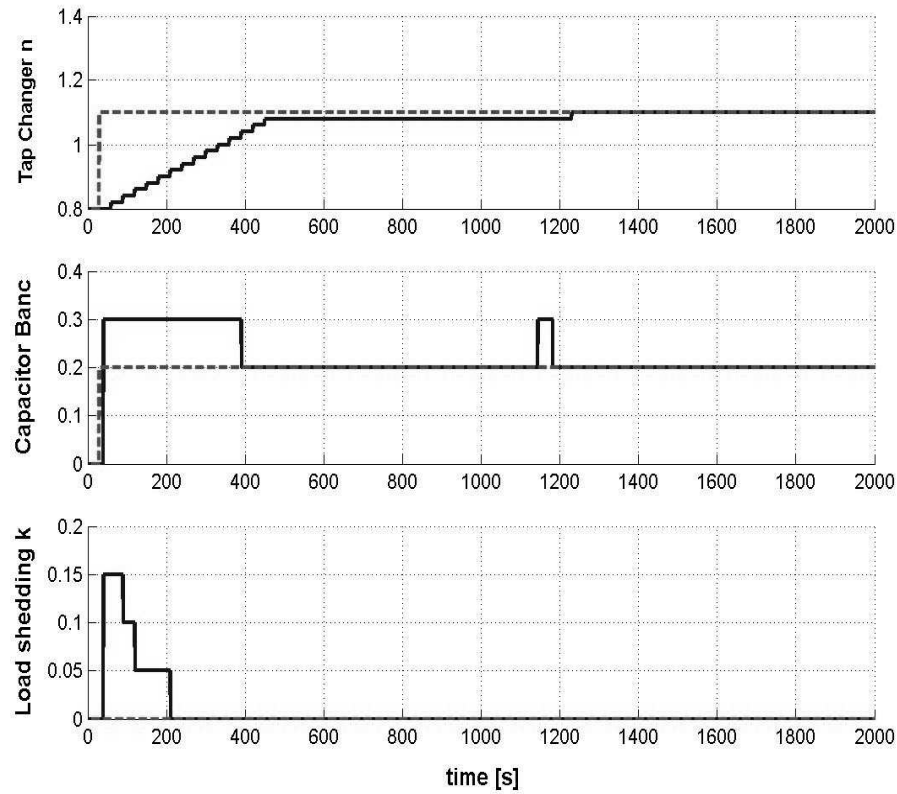
Fault clearance



Worst case

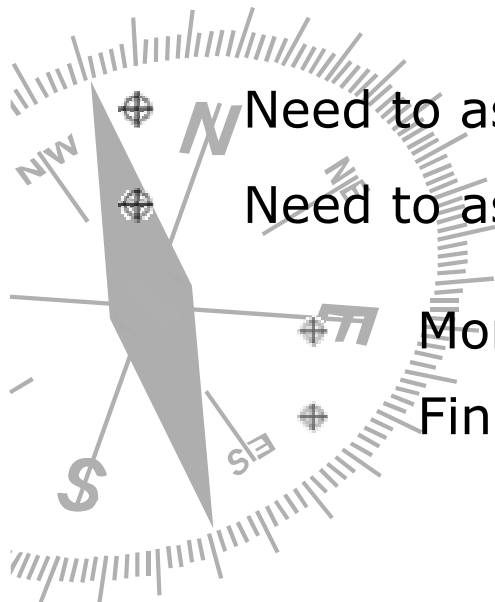


Worst case



Conclusions and future work

- ✗ Stabilization proved possible by simulating different cases
- ✗ The approach proposed is tractable from the computational view point



Need to assert stability

Need to assert that switching is finite

More systematic way to tune the parameters

Find more suitable monitoring signals